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**CLOSED-LOOP OPTIMIZATION PROGRAM  
FOR THE M60A1 TANK GUN STABILIZATION  
SYSTEM**

**W. Binroth, et al**

**Bendix Research Laboratories**

**Prepared for:**

**Rock Island Arsenal**

**February 1975**

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20. ABSTRACT (Continue on reverse side if necessary and identify by block number) THE OBJECTIVE OF THIS STUDY WAS TO PROVIDE A GUN STABILIZATION SYSTEM FOR THE M60A1 TANK GUN IN WHICH PERFORMANCE WAS OPTIMIZED. A MATHEMATICAL MODEL OF THE VEHICLE AND THE APPROPRIATE CONTROL SYSTEM WAS DERIVED AND PROGRAMMED ON A HYBRID COMPUTER. PERFORMANCE REQUIREMENTS WERE ESTABLISHED WITH RESPECT TO GUN POINTING ACCURACY AND CONTROL SYSTEM STABILITY. HARDWARE CHARACTERISTICS REQUIRED TO MEET ACQUISITION RATE REQUIREMENTS WERE ALSO DEFINED. A COMPLETE ANALYTICAL SYSTEMS STUDY WAS PERFORMED TO YIELD A BASIS FOR THE HYBRID COMPUTER ANALYSIS. IN THE COMPUTER ANALYSIS, SYSTEM PERFORMANCE		

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WAS INVESTIGATED IN DETAIL AND FIVE PROSPECTIVE SENSORS WERE STUDIED WITH  
RESPECT TO SYSTEM PERFORMANCE CRITERIA.

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## FOREWORD

This final report describes the study conducted under Army Contract DAAA09-74-C-2068, "Closed-Loop Optimization Program for the M60A1 Tank Gun Stabilization System." The work was administered under the direction of the Thomas J. Rodman Laboratories. Mr. Jack Connors and Mr. Verlin Baumgarth were the Contracting Officer's Representative and Alternate, respectively. The work was conducted during the period 1 July 1974 to 31 December 1974.

This report was prepared by the Information Processing Department, Bendix Research Laboratories, Southfield, Michigan. Mr. W. Binroth was project supervisor for this program, and responsible for the analytical effort, along with Mr. G. A. Cornell and Mr. R. W. Presley. The computer simulation studies were supported by Mr. T. A. Somer and Ms. J. A. Lindsay.

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# LIST OF SYMBOLS AND ABBREVIATIONS

<u>Symbol</u>	<u>Definition</u>	<u>Units</u>
$A_p$	Piston area	$\text{cm}^2$
$A_v$	Servo valve flow area	$\text{cm}^2$
$A_{v0}$	Effective servo valve flow area	$\text{cm}^2$
$A_{v \max}$	Maximum servo valve flow area	$\text{cm}^2$
$a_1, a_2, a_3, a_4$	Dummy constants in stiffness transfer function	—
$B$	Bulk modulus	$\text{kg/cm}^2$
$C$	Flow coefficient	$\text{cm} / \left( \text{s} - \sqrt{\text{kg/cm}^2} \right)$
$C_h$	Hull suspension constant	$\text{kg-m/rad}$
$D_a$	Denominator of the actuator and gun transfer function	—
$D_c$	Coulomb friction coefficient	$\text{kg-m}$
$D_g$	Denominator of gyro transfer function	—
$D_{gh}$	Gun trunnion friction coefficient	$\text{kg-m}$
$D_m$	Motor displacement	$\text{cm}^3/\text{rad}$
$D_s$	Stiction friction coefficient	$\text{kg-m}$
$D_{sa}$	Hull suspension damping-azimuth	$\text{kg-m-s}$
$D_{se}$	Hull suspension damping-elevation	$\text{kg-m-s}$
$D_{th}$	Turret friction coefficient	$\text{kg-m-s}$
$D_v$	Viscous friction coefficient	$\text{kg-m-s}$
$G_A$	Servo valve transfer function	—

<u>Symbol</u>	<u>Definition</u>	<u>Units</u>
$G_c$	Cubic denominator of linearized system transfer function	—
$G_p$	Pressure feedback transfer function	—
$G_1$	Controller transfer function	—
$G_2$	Actuator transfer function	—
$H_x$	Angular momentum about x-axis	—
$J_g$	Gun inertia	$\text{kg-m-s}^2$
$J_{gx}$	Gun inertia about the roll axis	$\text{kg-m-s}^2$
$J_{gy}$	Gun inertia about the yaw axis	$\text{kg-m-s}^2$
$J_{ha}$	Hull inertia-azimuth	$\text{kg-m-s}^2$
$J_{he}$	Hull inertia-elevation	$\text{kg-m-s}^2$
$J_m$	Motor inertia	$\text{kg-m-s}^2$
$J_t$	Turret inertia	$\text{kg-m-s}^2$
$J_{tx}$	Turret inertia about the roll axis	$\text{kg-m-s}^2$
$J_{tz}$	Turret inertia about the pitch axis	$\text{kg-m-s}^2$
$J_x$	Inertia about x axis	$\text{kg-m-s}^2$
$K$	Gain of nulling network	—
$k_a$	Servo valve gain	$\text{cm}^2/(\text{rad/s})$
$K_{bg}$	Gun sensor gain	—
$K_{bh}$	Hull sensor gain	—
$K_h$	Control gain of hull rate cancellation term	—
$K_i$	Integral controller gain	$1/\text{s}$
$K_o$	Static stiffness	$\text{kg-m/rad}$
$K_p$	Pressure feedback gain	$\text{rad(s-kg/cm}^2)$

<u>Symbol</u>	<u>Definition</u>	<u>Units</u>
$K_r$	Proportional controller gain	—
$K_s$	Open loop system gain	—
$K_{sa}$	Hull suspension spring constant-azimuth	kg-m/rad
$K_{se}$	Hull suspension spring constant-elevation	kg-m/rad
$k$	Ratio of specific heats	—
$L$	Leakage coefficient	$\text{cm}^3/(\text{s}-\text{kg}/\text{cm}^2)$
$L_g M_g$	Gun mass unbalance	—
$L_t M_t$	Turret mass unbalance	—
$l$	Lever arm of piston actuator	cm
$N_c$	Numerator of feedback compensation transfer function	—
$P_f$	Output of pressure feedback network	rad/s
$P_m$	Motor differential pressure	$\text{kg}/\text{cm}^2$
$P_o$	Effective motor pressure	$\text{kg}/\text{cm}^2$
$P_s$	Supply pressure	$\text{kg}/\text{cm}^2$
$P_{so}$	Initial supply pressure	$\text{kg}/\text{cm}^2$
$Q_l$	Leakage flow	$\text{cm}^3/\text{s}$
$Q_m$	Motor flow	$\text{cm}^3/\text{s}$
$Q_o$	Dummy variable used in linear analysis	—
$Q_p$	Pump flow	$\text{cm}^3/\text{s}$
$Q_v$	Servo valve flow	$\text{cm}^3/\text{s}$
$R$	Transmission ratio	—

<u>Symbol</u>	<u>Definition</u>	<u>Units</u>
$R_h, R_p$	Distance between the hull centers of rotation and the trunnion	cm
$s$	Laplace operator	s
$T_A$	Actuator torque	kg-m
$T_c$	Coulomb friction	kg-m
$T_d$	Disturbance torque	kg-m
$TE$	Tracking error	rad
$T_f$	Total friction	kg-m
$T_{fg}$	Trunnion friction torque	kg-m
$T_{gc}$	Gun gyroscopic torque	kg-m
$T_{gf}$	Gun firing torque	kg-m
$T_{g1}$	Coupling torque due to turret angle	kg-m
$T_{g2}$	Coupling torque due to hull acceleration	kg-m
$T_s$	Stiction friction	kg-m
$T_t$	Coupling torque due to mass unbalance	kg-m
$T_{tc}$	Turret gyroscopic torque	kg-m
$T_v$	Viscous friction	kg-m
$T_z$	Gyroscopic torque about z axis	kg-m
$V$	Actuator volume	cm <sup>3</sup>
$V_a$	Accumulator gas volume	cm <sup>3</sup>
$V_o$	Steady state gas volume in accumulator	cm <sup>3</sup>
$Y(s)$	Sensor transfer function	-

<u>Symbol</u>	<u>Definition</u>	<u>Units</u>
$\alpha, \beta$	Dummy variables used in linear analysis	—
$e$	Controller error	rad/s
$e_d$	Integrator drift	rad/s
$e_h$	Hull sensor offset error	rad/s
$e_g$	Gun sensor offset error	rad/s
$e_s$	Input to system open loop transfer function	rad/s
$e_t$	Sensed tracking error	mils
$\zeta$	Sensor damping ratio	—
$\Delta A_v$	Change in servovalve area	cm <sup>2</sup> /s
$\Delta P_m$	Change in motor pressure	(kg/cm <sup>2</sup> )/s
$\Delta T$	Breakaway torque in friction model	kg-m
$\dot{\theta}$	Threshold speed in friction model	rad/s
$\Delta \theta_g$	Change in gun angle	rad
$\Delta \theta_h$	Change in hull angle	rad
$\theta_g$	Gun angle	rad
$\dot{\theta}_{gs}$	Gun sensor output	rad/s
$\theta_h$	Hull pitch angle	rad
$\dot{\theta}_{hs}$	Hull sensor output	rad/s
$\dot{\theta}_{ht}$	Terrain pitch rate	rad/s
$\dot{\theta}_{hv}$	Relative hull pitch rate	rad/s
$\theta_s$	Position command	rad
$\dot{\theta}_s$	Rate command	rad/s
$\tau_1, \tau_2$	Time constants in forward compensation network	s



<u>Symbol</u>	<u>Definition</u>	<u>Units</u>
$\tau_3, \tau_4$	Time constants in feedback compensation network	s
$\tau_a$	Integrator gain in the integrating accelerometer rate computation network	s
$\tau_b$	Time constant of the integrating accelerometer transfer function	s
$\tau_c$	Compressibility time constant	s
$\tau_d$	Time delay of laminar vortex sensor	s
$\tau_f$	Stiction friction time constant	s
$\tau_i$	Time constant of the pseudo-differentiator in the input compensation network	s
$\tau_n$	Time constant of servovalve dynamics	s
$\tau_p$	Pressure feedback time constant	s
$\tau_s$	Time constant of laminar vortex sensor	s
$\phi_h$	Hull roll angle	rad
$\dot{\psi}_{ht}$	Terrain yaw rate	rad/s
$\dot{\psi}_{hv}$	Relative hull yaw rate	rad/s
$\psi_t$	Turret angle	rad
$\omega_b$	Bandwidth of integrating accelerometer	rad/s
$\omega_n$	Natural frequency of sensor	rad/s
$\omega_{NS}$	Dummy variable used in linear analysis	—
$\omega_x, \omega_y$	Angular rates about x and y axes, respectively	rad/s

<u>Symbol</u>	<u>Definition</u>	<u>Units</u>
$\frac{\partial T}{\partial n}$	Slope of torque-speed curve	kg-m-s
$\frac{\partial T}{\partial e}$	Slope of torque-error signal curve	kg-m/rad

#### ABBREVIATIONS

cc	cubic centimeter
cm	centimeter
gpm	gallons per minute
Hz	Hertz
in	inch(es)
kg-m	kilogram-meter
kg/cm <sup>2</sup>	kilogram/square centimeter
mil	milliradian
p-p	peak to peak
psi	pounds/square inch
rad	radian(s)
s	second(s)

## SECTION 1

### INTRODUCTION AND SUMMARY

#### 1.1 INTRODUCTION

The "Closed-Loop Optimization Program for the M60A1 Tank Gun Stabilization System" study project was undertaken by the Bendix Research Laboratories under Contract DAAA09-74-C-2068 for the Rock Island Arsenal. The objective was to develop a gun stabilization system in which performance and reliability are optimized. It had been concluded prior to the initiation of this contract that fluidic approaches appeared attractive from the reliability and cost standpoint. The purpose of this contract is to provide the Contracting Officers with conclusive evidence and recommendations regarding the applicability of the fluidic approach to tank weapons systems, while supplying a mathematical model of the entire vehicle/control system. In order to provide this information, fluidic devices were to be considered without forcing incorporation of a particular existing fluidic device.

The results of Phase I, discussed in this report, consist of a requirement analysis, a model definition, and a system stability and performance analysis. A detailed performance study of the different applicable sensors consisting of electronic rate gyros, a hydraulic rate sensor, an integrating accelerometer, a laminar vortex sensor, and a pneumatic accelerometer were also included for future trade studies. A complete error analysis for two of the sensors is recommended in order to augment a decision on the applicability of fluidic approaches in this area.

Control system design was performed in accordance with specifications which call for a pointing accuracy of 1/2 mil rms diameter circle without a gunner operating the controls. A 3 dB bandwidth of at least 15 Hz is also desired. It is required that the system hardware be out of saturation 98 to 99 percent of the time. The duration of an engagement was specified to be 15 to 20 s.

#### 1.2 SCOPE

For the purpose of deriving a realistic mathematical model of the stabilization system, a study was made of existing literature on the M60A1 tank, and of previous system studies on the vehicle. Upon simplifying models of the vehicle in the literature, where appropriate, and independently developing realistic models of the servovalve, actuators and sensors, a complete control system model was derived. Selection of the appropriate actuator system hardware to be used was carried out in close cooperation with the Contracting Officer's Representatives. Models were defined separately for the azimuth (turret) and elevation

(gun) axes. In addition to the vehicle and gun models, the mathematical model included the controller and compensation, the effects of nonlinear valve flow, and static, coulomb, and viscous friction effects. The effect of hull dynamics on the system was also evaluated. A mathematical model of coupling between axes due to roll motion was investigated analytically with respect to the approximate effects on system performance and pointing accuracy. Control laws were derived analytically in accordance with the performance specifications set forth prior to the start of the contract. A control law was specified for both rate and position command input concepts.

From the model definition, an analog computer program was defined and programmed for each control concept. Diagrams and descriptions of all programs used in this study are contained in Appendices A and B. The Rock Island Arsenal digital program "HITPRO" was adapted to the Bendix Research Laboratories computer, and the analog programs were, at that point, modified to hybrid programs in order to accept inputs from HITPRO. These inputs consisted of vehicle rates corresponding to those experienced by an M60 tank while traversing the Aberdeen Proving Ground terrain. Additional hybrid functions of these programs consisted of a time delay property of the laminar vortex sensor.

Prior to a complete hybrid computer analysis of the system, an analytical systems study was conducted in order to develop a fuller comprehension of the overall system behavior. The study also aided in the development of digital check solutions for all of the hybrid computer analyses. In addition, the analytical study includes a sensor error analysis and an evaluation of coupling of axes on pointing errors.

The hybrid computer simulation results consist of an evaluation of the effects of nonlinear valve flow and hull dynamics, a comparison of the rate and position control concepts, and a detailed evaluation of each of the five sensors studied. The electric rate gyros, as they are presently used in the field, were evaluated in both the elevation and azimuth axes. The hydraulic rate sensor, the integrating accelerometer, and the laminar vortex sensor were evaluated in the elevation axis, and the pneumatic accelerometer was evaluated in the azimuth axis.

The conclusions reached on the basis of the computer analysis are contained in Section 7. Recommendations as to future studies and stabilization approaches are contained in Section 8.

### 1.3 SUMMARY

A mathematical model of a suitable stabilization system for the M60A1 tank main gun was formulated and programmed. The model was formulated so as to include most of the significant nonlinearities such as nonlinear valve flow and hull dynamics due to gun motion. A hybrid computer analysis was performed to determine the operating characteristics of the stabilization system, to evaluate prospective sensors for sensing gun and/or hull rate, and to determine whether a rate or a

position command control concept is preferable with respect to specified performance criteria.

The analytical study revealed that both the rate and the position control concepts required a proportional plus integral control law in order to minimize the gun tracking error. It was also shown that the rate and the position concepts are equivalent in terms of nulling out the effects of hull motions, and thus in terms of stabilizing the gun after the target is in the sight. A computer analysis which followed verified this equivalence. In addition, it was possible to show that the effect of hull motions on the system can be minimized by either a high control loop gain along with a lead-lag compensation network or by using a hull sensor signal in the control law.

The extensive computer simulation analysis revealed several significant conclusions in the areas of stabilization control philosophy and sensor applicability. In the process of arriving at a full computer model of the system for sensor evaluation, it was found that the effect of hull dynamics on the gun was negligible. Nonlinear valve flow, however, was found to have a significant influence on system performance. A linearized flow model was not sufficiently accurate for use in this study.

It was found that all five of the sensors studied meet the performance criteria set forth by the Contracting Officer's Representative. In addition, this study indicates that these criteria can be met by using only a gun sensor. If verified by further studies, the need for a corresponding hull sensor may be eliminated.

The detailed sensor study revealed that automatic offset and integrator drift nulling circuits are required when using an acceleration sensor. A method which can be used for this purpose is described in Section 5.5. In addition, it was determined that increasing the gain of the acceleration sensor will decrease the sensor offset effects and, hence, the drift rate. More generally, in the sensor study it was found that sensor gain errors have a small effect on the tracking error. Also, a combination of sensor deadband and gun or turret friction will cause the system to limit cycle.

In order to compensate for sensor phase lag, feedback compensation was found to be required. Forward path compensation is desirable for obtaining stability with higher loop gains for this system.

## SECTION 2

### MATHEMATICAL MODEL DESCRIPTION

#### 2.1 GENERAL

This section contains the mathematical model of the gun stabilization system as programmed on the hybrid computer. This model consists of the vehicle model, servovalve and motor, sensor models, and the controller. The azimuth and elevation systems are presented separately. The effects of coupling are not included. The individual system components are described in Sections 2.2 through 2.4. Also presented are the controller models for the following configurations:

- (1) Rate control with two rate sensors
- (2) Rate control with one acceleration sensor in gun axis
- (3) Position control with two rate sensors
- (4) Position control with one acceleration sensor in gun axis

A simplified block diagram of a rate control system with two rate sensors is shown in Figure 2-1.

#### 2.2 VEHICLE MODELS

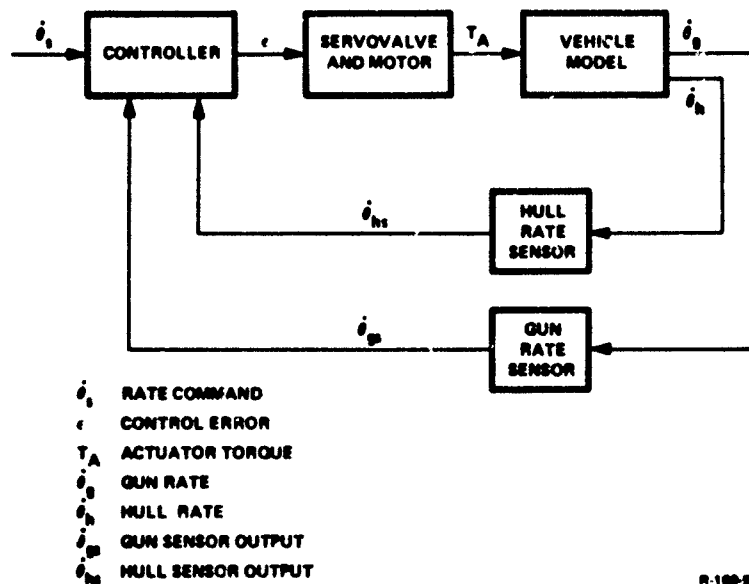
The vehicle models consist of an elevation (gun) axis model and an azimuth (turret) axis model. They include, respectively, the gun and turret inertias, effects of friction, and hull suspension effects.

##### 2.2.1 Elevation Axis Vehicle Model

The elevation axis vehicle model shown in Figure 2-2 consists of the combined gun and hull dynamics. For the gun dynamics, the torque applied to the gun equals the actuator torque  $T_A$ , plus the disturbance torque  $T_d$  acting on the gun, minus the trunnion friction torque  $T_{fg}$ . It follows that the gun angular acceleration with respect to inertial space is given by:

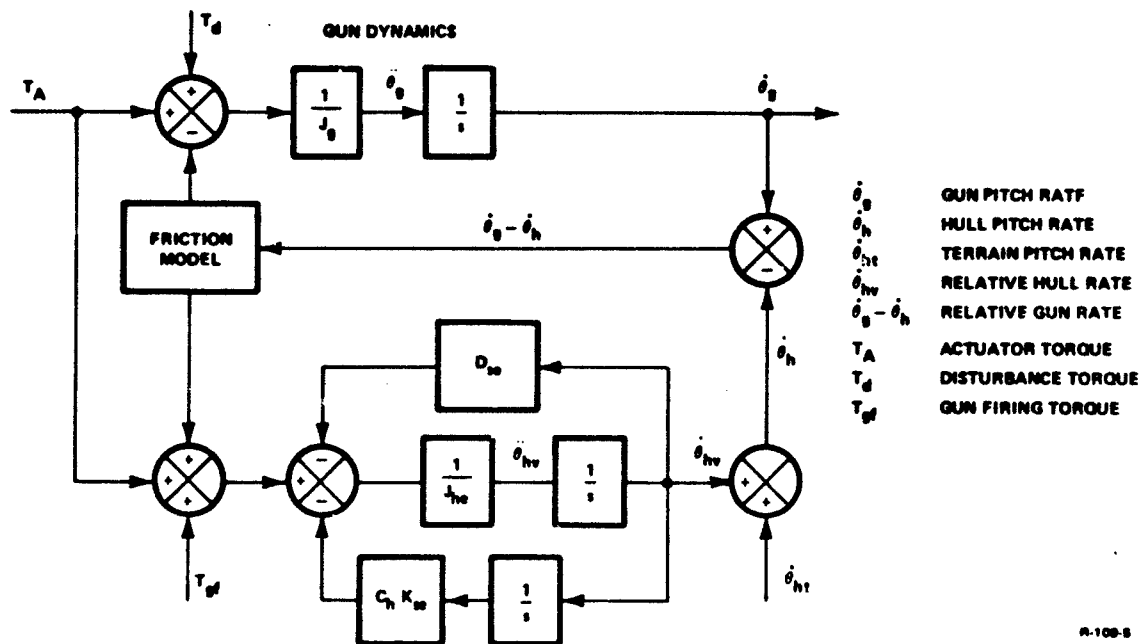
$$\ddot{\theta}_g = \frac{1}{J_g} (T_A + T_d - T_{fg})$$

The friction model is discussed in Section 2.2.3. The disturbance torque  $T_d$  is generated, for example, by a gun firing. The hull dynamics generate a hull angular acceleration  $\ddot{\theta}_{hv}$  with respect to the tracks. This



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Figure 2-1. Simplified Block Diagram of the Rate Command Stabilization System



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Figure 2-2. Elevation Axis Vehicle Model

acceleration results from the reaction of the actuator torque and the gun firing torque  $T_{gf}$ . The acceleration is given by:

$$\ddot{\theta}_{hv} = \frac{1}{J_{he}} [T_A + T_{gf} - D_{se} \dot{\theta}_{hv} - (K_{se} + C_h) \theta_{hv}]$$

where the last two terms represent the hull suspension.

The terrain pitch rate  $\dot{\theta}_{ht}$  added to the relative hull rate  $\dot{\theta}_{hv}$  gives the total hull pitch rate  $\dot{\theta}_h$  as shown below.

$$\dot{\theta}_h = \dot{\theta}_{hv} + \dot{\theta}_{ht}$$

The angular rate of the gun with respect to the hull (relative gun rate) is given by:

$$(\dot{\theta}_g - \dot{\theta}_h).$$

A list of the vehicle elevation parameters is contained in Table 2-1.

Table 2-1. Elevation Vehicle Parameters

Parameter	Value	Units
$D_{se}$	$7.8 \times 10^4$	kg-m-s
$D_{gh}$	166	kg-m-s
$J_g$	527	kg-m-s <sup>2</sup>
$J_{he}$	$1.73 \times 10^4$	kg-m-s <sup>2</sup>
$K_{se}$	$9.85 \times 10^6$	kg-m/rad
$C_h$	40,500	kg-m/rad
$\tau_f$	0.01	s

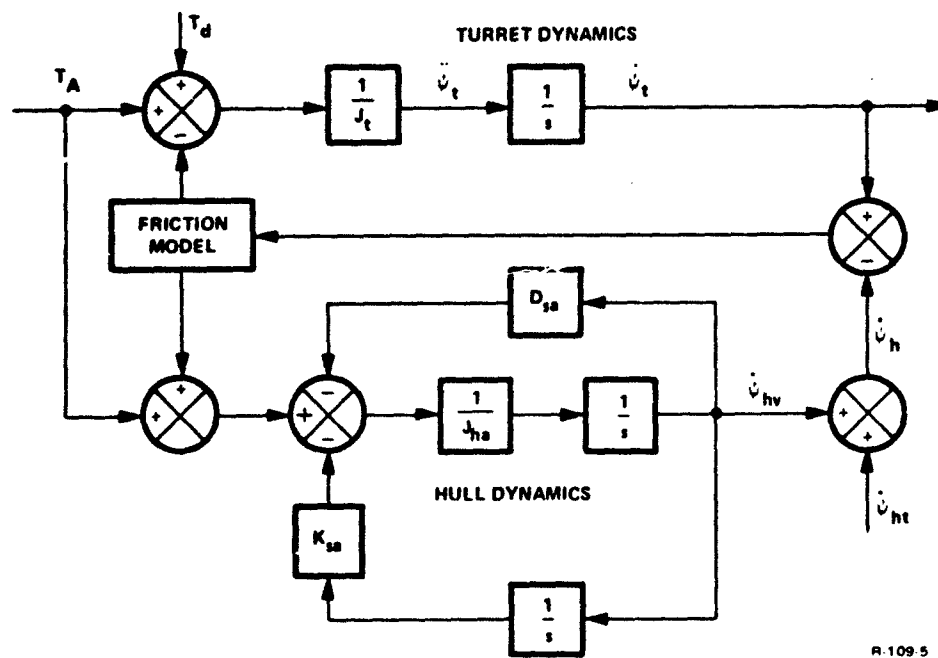


### 2.2.2 Azimuth Axis Vehicle Model

The azimuth axis vehicle model, shown in Figure 2-3, is the same as the elevation model, except for the values of the moments of inertia and friction. Angular position is denoted by  $\psi$  rather than  $\theta$ , as in the elevation model. A list of the vehicle azimuth parameters is contained in Table 2-2.

Table 2-2. Azimuth Vehicle Parameters

Parameter	Value	Units
$D_{sa}$	$8.3 \times 10^4$	kg-m-s
$D_{th}$	7.75	kg-m-s
$J_t$	3140	kg-m-s <sup>2</sup>
$J_{ha}$	$1.84 \times 10^4$	kg-m-s <sup>2</sup>
$K_{sa}$	$9.4 \times 10^6$	kg-m/rad



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Figure 2-3. Azimuth Axis Vehicle Model

### 2.2.3 Friction

Three types of friction, viscous, coulomb, and stiction, are considered here. Actuators generally have significant quantities of all three types. In addition, the turret and the gun trunnion may have one or more of the friction types. The three types of friction are described below.

#### Viscous Friction

Viscous friction is proportional to the speed of the gun relative to the hull.

$$T_v = D_v (\dot{\theta}_g - \dot{\theta}_h) \quad (2-1)$$

#### Coulomb Friction

Coulomb friction is constant in magnitude. The sign, or direction, changes when the direction of motor rotation changes.

$$T_c = D_c \text{ sign } (\dot{\theta}_g - \dot{\theta}_h) \quad (2-2)$$

#### Stiction Friction

Stiction occurs each time the motor starts or reverses direction. It rapidly decays to zero. The sign changes when the direction of rotation reverses. Stiction friction is modeled by the following equation:

$$T_s = D_s \frac{\tau_f s}{1 + \tau_f s} \text{ sign } (\dot{\theta}_g - \dot{\theta}_h) \quad (2-3)$$

Total friction is the sum of the three types of friction for both axes:

$$T_f = T_v + T_c + T_s \quad (2-4)$$

When compliance is added to the model, the actuator and trunnion frictions have different equations. When compliance is not used, the equations are the same, and the actuator and trunnion friction can be combined as is done here. The friction values given in equations (2-1) through (2-4) are the combined friction of the actuator and

trunnion, for elevation. For the azimuth axis, equation (2-4) represents the combined friction of the actuator and the turret.

The block diagram used for simulating friction is shown in Figure 2-4.

An ideal circuit for friction simulation is shown in Figure 2-5. In this diagram, the breakaway torque  $\Delta T$  is the minimum actuator torque that will result in actuator movement. When the magnitude of  $T_A$  is less than  $\Delta T$  and the gun is at rest, the friction torque exactly equals the actuator torque. At this point, no torque is applied to the gun. For this region of operation, the friction is called static friction.

$$T_f = T_A \quad (T_f = \text{static friction})$$

When the actuator torque exceeds the breakaway torque, the circuit switches to running friction, and  $T_f$  is given by equation (2-4). A torque is then applied to the gun, and gun motion starts. As long as the gun speed exceeds the threshold value  $\Delta \theta$ , the circuit stays in the running friction mode regardless of the value of  $T_A$ . The parameters in the friction model must be adjusted so that at the instant of switching from static to running friction, the two friction values are equal.

### 2.3 SERVOVALVE AND MOTOR MODELS

The servovalve and motor for both the elevation and azimuth systems are modeled as shown in Figure 2-6. This model includes a pressure feedback servovalve with nonlinear flow dynamics. The voltage applied to the servovalve is the sum of  $\epsilon$  and  $P_f$ . The quantity  $\epsilon$  is the output of the control law as described in Section 3. The quantity  $P_f$  is the output of the pressure feedback network. The servovalve dynamics  $G_a$  are represented by a first order lag, i.e.,

$$G_a = \frac{1}{\tau_n s + 1}$$

The servovalve gain is  $K_a$ . In the simulation, the deadband was assumed zero while the valve area  $A_v$  was limited to  $A_v \text{ max}$ . The servovalve flow rate  $Q_v$  is given by:

$$Q_v = CA_v \sqrt{\frac{P_s - P_m}{2}}, \quad \text{for } A_v > 0$$

$$Q_v = CA_v \sqrt{\frac{P_s + P_m}{2}}, \quad \text{for } A_v < 0$$

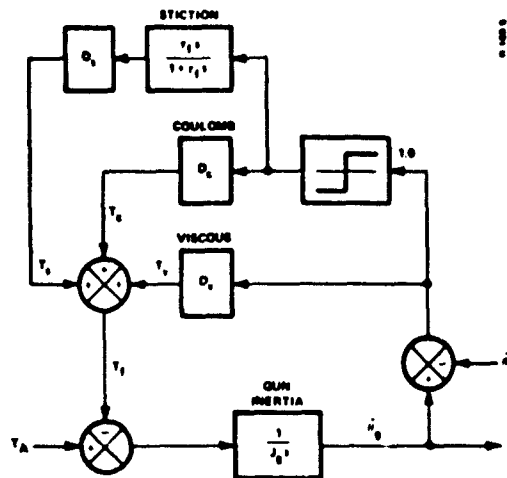


Figure 2-4. Block Diagram for Simulating Friction

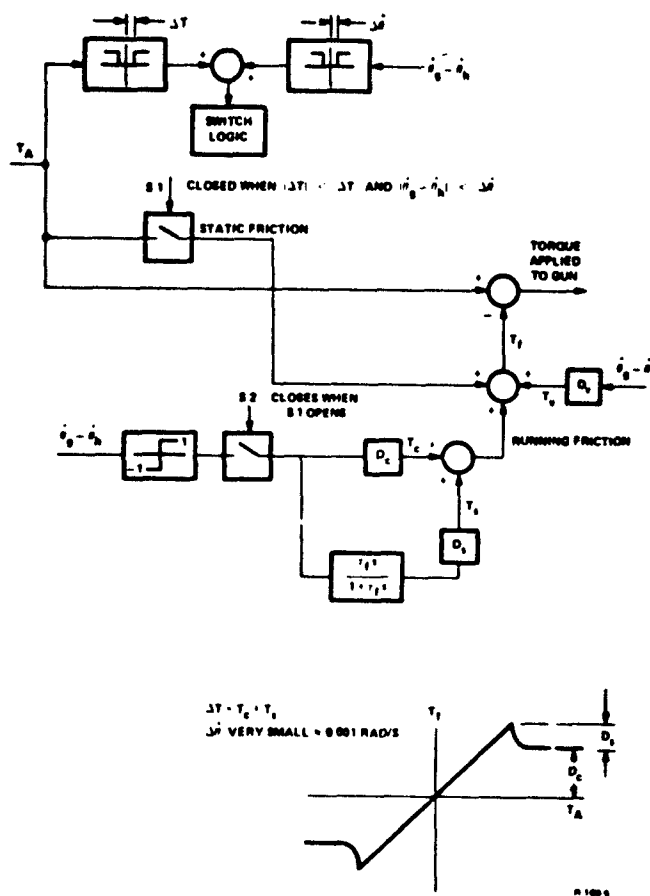


Figure 2-5. Idealized Block Diagram for Simulating Friction

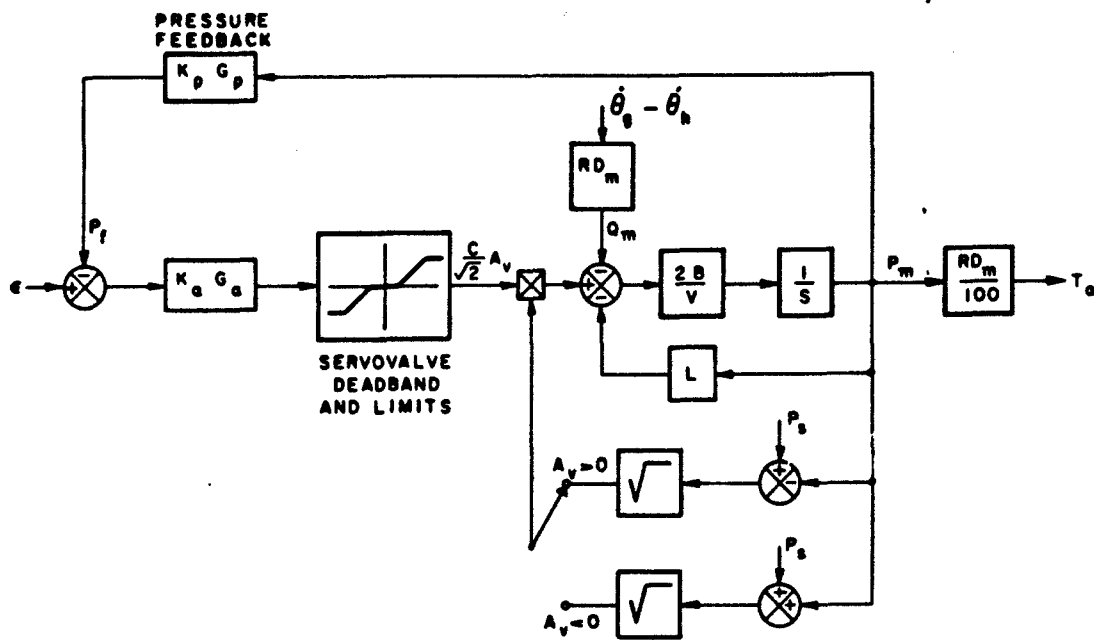


Figure 2-6. Servovalve and Motor Simulation Block Diagram

where

$C$  = flow coefficient

$P_s$  = supply pressure

$P_m$  = motor pressure

The rate of change of motor pressure  $P_m$  is given by:

$$\dot{P}_m = \frac{2B}{V} (Q_v - Q_m - Q_l)$$

where

$B$  = bulk modulus

$V$  = volume

$Q_m$  = displacement flow rate, i.e.

$$Q_m = RD_m (\dot{\theta}_g - \dot{\theta}_h)$$

and where  $RD_m$  is the motor displacement reflected to the gun. For a piston actuator,  $RD_m$  is replaced by  $2A_p$ . The model is otherwise unchanged. Leakage flow  $Q_l$  is assumed proportional to the motor pressure and is thus given by:

$$Q_l = L P_m$$

The resultant actuator torque  $T_A$  is then given by:

$$T_A = \frac{RD_m}{100} P_m$$

The dynamic pressure feedback network  $K_p G_p$  is utilized for stabilizing and improving the servovalve/motor response. The compensator  $G_p$  is used to wash out the pressure feedback at low frequency or at steady state. Pressure feedback without the washout would reduce the static stiffness. The function  $G_p$  can either be provided by a servovalve designed to have dynamic pressure feedback, or by an electrical network. For each axis, the gain  $K_p$  is a system parameter that was varied to obtain the desired system response. The pressure feedback transfer function is given by:

$$K_p G_p = K_p \frac{\tau_p s}{1 + \tau_p s}$$

Table 2-3 contains the parameters of the servovalve and motor for both the elevation and azimuth systems.

Table 2-3. Hardware Parameters

Parameter	Value	Units
B	10,550	kg/cm <sup>2</sup>
C	955	cm/s $\sqrt{\text{kg/cm}^2}$
D <sub>c</sub>	Variable	kg-m
D <sub>s</sub>	Variable	kg-m
D <sub>v</sub>	Variable	kg-m-s
J <sub>m</sub>	Negligible	kg-m-s <sup>2</sup>
L	0.116	cm/s-kg/cm <sup>2</sup>
P <sub>s</sub>	210 (nominal)	kg/cm <sup>2</sup>
$\tau_p$	0.07	s
K <sub>a</sub>	1	cm <sup>2</sup> /(rad/s)
$\tau_n$	0.008	s
<u>Elevation</u>		
A <sub>v max</sub>	0.22	cm <sup>2</sup>
RD <sub>m</sub>	2970 (initial studies)	cm <sup>3</sup> /rad
	1311 (optimum value)	cm <sup>3</sup> /rad
V	855	cm <sup>3</sup>
<u>Azimuth</u>		
A <sub>v max</sub>	0.45	cm <sup>2</sup>
RD <sub>m</sub>	1639	cm <sup>3</sup> /rad
V	492	cm <sup>3</sup>
<u>Linear Model</u>		
$\partial T / \partial n$	$2.08 \times 10^4$	kg-m-s
$\partial T / \partial \epsilon$	$6.25 \times 10^4$	kg-m/rad
$\tau_c$	0.016	s

## 2.4 SENSOR MODEL DESCRIPTIONS

### 2.4.1 General

A set of five different sensors was selected by the Contracting Officers for study. All five sensors were likely candidates for this application, and it was desired to establish the performance characteristics of each sensor on a relative and absolute basis, in an appropriate realistic environment. Properties of the electric rate gyros, the General Electric hydraulic rate sensors, and the Honeywell laminar vortex sensor were provided to Bendix by the Contracting Officers. The model of the Airsearch pneumatic accelerometer was provided directly to Bendix by Airsearch. The model of the Bendix integrating accelerometer was obtained from Bendix personnel. For each of the three rate sensors, a configuration was used in which a sensor was placed at the gun and in the hull. For the two accelerometers, a sensor was placed only at the gun.

The models for the sensors are presented in terms of transfer functions. All sensor models have provisions for adding threshold and varying the sensor gains. All nominal sensor gains were considered unity. The sensor gain has no effect on response to command inputs or terrain disturbances. Gain distribution does, however, affect the response to noise, and drift due to offsets. Noise and drift were not studied in detail in this phase. When these effects are studied, the actual sensor gains must be used.

### 2.4.2 Electric Rate Gyro Model

The electric rate gyros which measure the angular gun rate and hull rate can be represented by a second order transfer function as given by:

$$\dot{\theta}_{gs} = \frac{1}{1 + \frac{2\zeta}{\omega_n} s + \frac{1}{\omega_n^2} s^2} \dot{\theta}_g$$

$$\dot{\theta}_{hs} = \frac{1}{1 + \frac{2\zeta}{\omega_n} s + \frac{1}{\omega_n^2} s^2} \dot{\theta}_h$$

where

$$\omega_n = 157 \text{ rad/s}$$

$$\zeta = 0.7$$



#### 2.4.3 Hydraulic Rate Sensor Model

The hydraulic rate sensor utilized in this study is manufactured by the General Electric Corporation. The dynamics of this sensor can be represented by a second order transfer function as follows:

$$\dot{\theta}_{gs} = \frac{1}{1 + \frac{2\zeta}{\omega_n} s + \frac{1}{\omega_n^2} s^2} \dot{\theta}_g$$

$$\dot{\theta}_{hs} = \frac{1}{1 + \frac{2\zeta}{\omega_n} s + \frac{1}{\omega_n^2} s^2} \dot{\theta}_h$$

where

$$\zeta = 0.707$$

and

$$\omega_n = 40 \times 2\pi \text{ rad/s}$$

The following are additional characteristics of this sensor:

Scale factor: 0.13 psi/deg/s

Mass unbalance drift: 2 deg/s/g at 1800 psi

Noise: 0.6 deg/s (peak to peak)

Supply pressure drift: 0.3 deg/s/percent change in supply pressure

Scale factor drift: 0.8 percent/percent change in supply pressure

#### 2.4.4 Hydraulic Integrating Accelerometer Model

The hydraulic integrating accelerometer produced by Bendix measures the gun acceleration  $\ddot{\theta}_g$ . Its transfer function is given by:

$$\ddot{\theta}_{gs} = \frac{\tau_b}{1 + \tau_b s} \ddot{\theta}_g$$

where

$$\tau_b = \frac{1}{\omega_b}$$

and

$$\omega_b = 2\pi \times 0.1 \text{ rad/s}$$

The sensed acceleration is converted to a rate signal  $\dot{\theta}'_g$  by the rate computation network shown below.

$$\dot{\theta}'_{gs} = \left(1 + \frac{1}{\tau_a s}\right) \ddot{\theta}_g$$

where  $\tau_a$  is set equal to  $\tau_b$  which is the time constant of the integrating accelerometer.

If the sensor break frequency varies due to temperature changes or other factors, the sensed rate signal becomes

$$\dot{\theta}'_{gs} = \frac{\tau_b s}{1 + \tau_b s} \left(1 + \frac{1}{\tau_a s}\right) \ddot{\theta}_g = \frac{\tau_b}{\tau_a} \left(\frac{1 + \tau_a s}{1 + \tau_b s}\right) \ddot{\theta}_g$$

Above the break frequency, the signal is not affected by the change in break frequency. At lower frequencies, there is a gain change proportional to the change in break frequency.

#### 2.4.5 Laminar Vortex Rate Sensor Model

The transfer function for the laminar vortex sensor produced by Honeywell is a first order lag with transport delay, i.e.,

$$\dot{\theta}_{gs} = \frac{e^{-\tau_d s}}{1 + \tau_s s} \dot{\theta}_g$$

$$\dot{\theta}_{hs} = \frac{e^{-\tau_d s}}{1 + \tau_s s} \dot{\theta}_h$$

where

$$\tau_d = 0.01 \text{ s (transport delay)}$$

$$\tau_g = 0.002 \text{ s (time constant)}$$

#### 2.4.6 Pneumatic Accelerometer Model

The basic pneumatic accelerometer produced by Garrett Airsearch has a second order transfer function, with lead-lag compensation, in a closed loop configuration. The block diagram of the sensor is shown in Figure 2-7 where

$$\omega_n = 14.8 \text{ rad/s}$$

$$\zeta = 0.6$$

The accelerometer saturates at  $0.7 \text{ rad/s}^2$ .

The rate computation network which transforms the sensed acceleration into a rate is a simple integrator.

#### 2.5 CONTROLLER MODELS

The four controller model variations developed for this study are described in this section, and are listed below.

- (1) Rate command with two rate sensors
- (2) Rate command with single acceleration sensor
- (3) Position command with two rate sensors
- (4) Position command with single acceleration sensor

Two of the variations considered are rate command systems where the gunner commands angular rates. One of the rate systems utilizes two rate sensors (gun and hull) while the other utilizes a single acceleration sensor (gun). Two of the models are position control where the gunner commands an angular position. Again, one of these systems utilizes two rate sensors

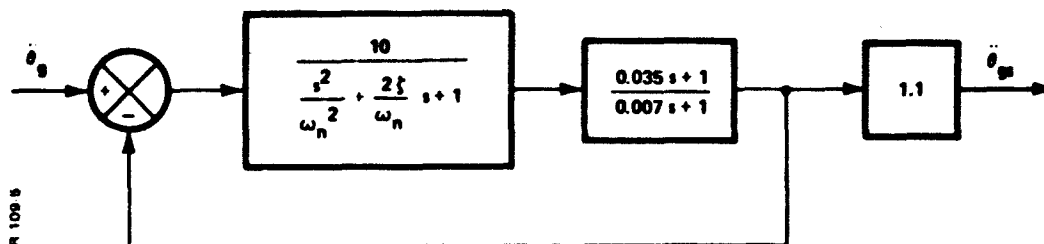
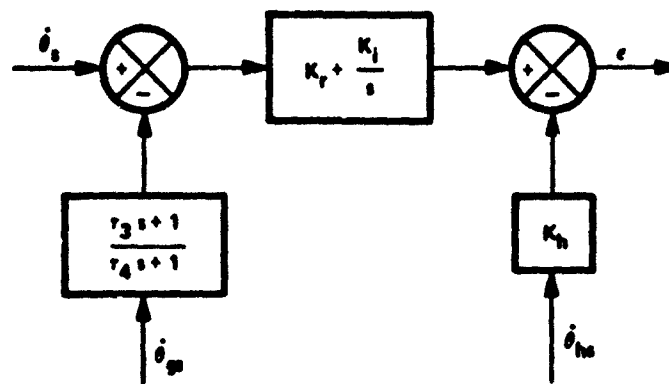


Figure 2-7. Block Diagram of the Pneumatic Accelerometer

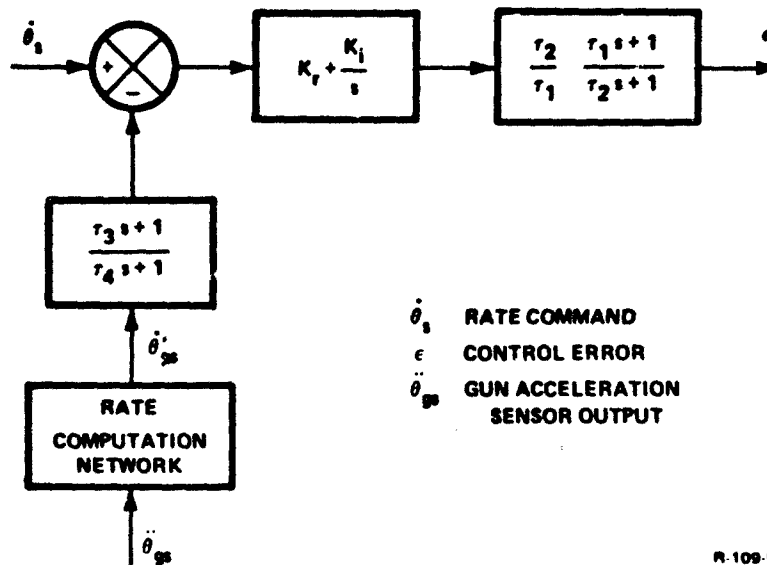
while the other utilizes a single acceleration sensor. Block diagrams of each of these controllers are shown in Figures 2-8 through 2-11. Proportional and integral control is utilized along with compensation networks in the forward and feedback paths and on the input in the position command system. The controller utilizing a single acceleration sensor requires a rate computation network to transform the sensed acceleration into a rate. These rate computation networks are defined with the acceleration sensor models in Section 2.4.4 and 2.4.6.



$\dot{\theta}_s$  RATE COMMAND  
 $e$  CONTROL ERROR  
 $\dot{\theta}_{gs}$  GUN RATE SENSOR OUTPUT  
 $\dot{\theta}_{hs}$  HULL RATE SENSOR OUTPUT

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Figure 2-8. Controller for Rate Command with Two Rate Sensors



$\dot{\theta}_s$  RATE COMMAND  
 $e$  CONTROL ERROR  
 $\ddot{\theta}_{gs}$  GUN ACCELERATION  
 SENSOR OUTPUT

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Figure 2-9. Controller for Rate Command with Single Acceleration Sensor



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2

## SECTION 3 CONTROL SYSTEMS APPROACH

### 3.1 GENERAL

General requirements of the M60A1 tank gun stabilization system are as follows:

- (1) To rapidly attain a target by means of gunner commands after the target has been sighted by the gunner.
- (2) To keep the gun on target after the gun has been aimed by the gunner in spite of hull motions.

In order to satisfy criterion (1), it is necessary that the response of the system be sufficiently fast so that the time from sighting the target until firing of the first round be minimum. For satisfying criterion (2), it is necessary to minimize the effects of vehicle hull rates on the tracking error of the stabilization system.

Since the design of a control system for a given application is subject to all performance criteria, the criteria for the M60A1 gun stabilization system are described in Section 3.2. Note that these criteria are given in terms of pointing requirements as well as general performance. Hardware acceleration performance criteria are also given.

The control systems approach utilized in this study for achieving the specified performance criteria is described in detail in Sections 3.3.1 and 3.3.2, for the rate and the position command inputs, respectively. It was specified at the start of the program that contemporary conventional control system techniques would be used to achieve the desired performance. If conventional techniques could not achieve the performance goals, more complex optimal procedures would be utilized. At this point in the development of the complete stabilization system, conventional methods of synthesis appear satisfactory.

### 3.2 PERFORMANCE CRITERIA

The following performance criteria were utilized in the control system design and hardware selection procedure for the stabilization system.

#### (1) Tracking Error Requirements

The requirement for the tracking error is that it be within a circle of 0.5 mil diameter 67 percent of the time. Data to verify conformance with this requirement will not be obtained in Phase I. Therefore, a goal was set for the tracking error to be 0.5 mil peak to peak or less for the HITPRO bump course at 8 mph.

## (2) Frequency Response and Stability Requirements

The frequency response goal was 15 Hz bandpass. Bandpass is defined as the frequency at which the amplitude ratio is -3 dB.

Gain margin and phase margin are two criteria which are frequently used to indicate the system stability. In this study, the following definitions are used to establish stability goals:

- (a) Gain Margin =  $1.0 - (\text{open loop amplitude ratio at } 180 \text{ deg open loop phase lag})$
- (b) Phase Margin =  $180 \text{ deg} - (\text{open loop phase lag at gain crossover})$
- (c) Gain Crossover = frequency at which the open loop amplitude ratio is 1.0 (or zero dB)

A gain or phase margin of zero indicates an unstable system. The greater the gain or phase margin, the less likely it is that instability will occur due to variation in control system components. The goals are a gain margin of 0.5 or greater and a phase margin of 35 deg or greater.

## (3) Acceleration and Speed Capability of Actuators

The goal for the elevation actuator system is to accelerate to a speed of 60 deg/s in 10 deg of gun rotation, starting from rest. The goal for the azimuth actuator system is to accelerate to 90 deg/s in 45 deg of turret rotation, starting from rest.

## (4) Gain Variation

The control system shall be designed to minimize the effects of rate sensor gain variation and friction.

### 3.3 CONTROL LAW DERIVATIONS

The control synthesis approach utilized for this study is outlined in the following two sections, for the rate command input (rate control system) and the position command input (position control system), respectively. For the rate control system, a hull rate cancellation technique and a lead-lag stabilization technique are presented. The position control system is mathematically compared to the rate control system.

#### 3.3.1 Rate Control System

A simplified block diagram of a rate command stabilization system is shown in Figure 3-1. The inputs are the command rate  $\dot{\theta}_s$  and the hull rate  $\dot{\theta}_h$ . The output gun rate  $\dot{\theta}_g$ , which is compared with  $\dot{\theta}_s$  to generate an error, is processed by the controller  $G_1$ . An additional output is the gun attitude angle  $\theta_g$ .





The first of these methods is a hull rate cancellation technique. The hull motion  $\dot{\theta}_h$  is sensed and is then used in the control laws. The control signal output  $\epsilon$  is therefore given by:

$$\epsilon = \left( K_r + \frac{K_i}{s} \right) (\dot{\theta}_s - \dot{\theta}_g) - K_h \dot{\theta}_h$$

where

$K_h = K_3$  = gain of the hull motion effect on the gun rate

The second method uses a single gun sensor and does not use a hull sensor. The tracking error is reduced by increasing the loop gain and introducing a lead-lag network to maintain stability. The control signal output in this case is given by:

$$\epsilon = \frac{\tau_2}{\tau_1} \left( \frac{\tau_1 s + 1}{\tau_2 s + 1} \right) \left( K_r + \frac{K_i}{s} \right) (\dot{\theta}_s - \dot{\theta}_g)$$

where

$$\tau_2/\tau_1 = 10$$

### 3.3.2 Position Control System

The block diagram for position control with proportional plus integral control is shown in Figure 3-2. The gunner generates an angular position command  $\theta_g$  rather than a rate command  $\dot{\theta}_g$ . The position command signal is differentiated and compared with the gun rate signal to obtain the rate error signal. Since perfect differentiation cannot be achieved, a first order lag with time constant  $\tau_1$  is included in the input compensation.

In addition, the gun rate signal is integrated and compared with the input position command to obtain the position error signal. Potentiometer feedback was not used as this would have resulted in increased sensitivity to hull sensor gain errors, and increased drift rate.

The position command system can be shown to be mathematically identical to the rate system except for the first order lag  $\tau_1$  in the input compensation. (This is discussed in more detail in Sections 5.2.2 and 6.3.) On this basis, the tracking error for hull rate inputs is identical for the rate and the position control systems. In order to

perform a meaningful comparison of the two control system concepts, a rate command input was assumed for both. Under these conditions, for ramp position command inputs, the tracking error is also the same for both the rate and the position systems, as long as the input lag  $\tau_1$  is zero. If  $\tau_1$  is large enough to affect the system response, or if input compensation is not used, response of the position control system to rate inputs will be slower. However, one advantage with the position control system is that a position command signal can be used. This results in considerably faster response to the process of aiming the gun at a target for the position control system, even if input compensation is not used.

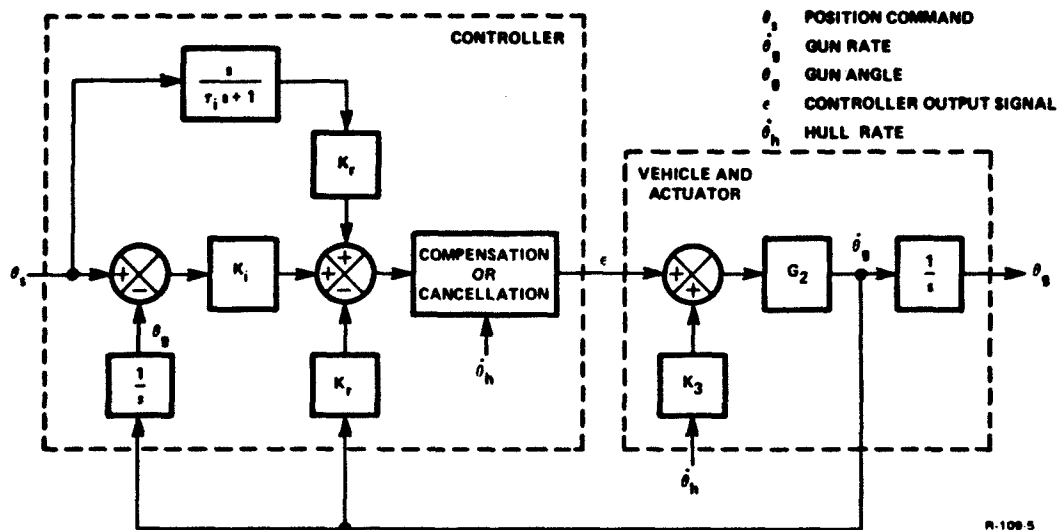


Figure 3-2. Simplified Block Diagram of a Position Command Stabilization System

**SECTION 4**  
**HARDWARE SELECTION PROCEDURE**

**4.1 GENERAL**

For the purpose of this contract, a study was performed for selecting the appropriate actuator hardware in each of the two axes to meet the following set of specifications set forth by the Contracting Officer's Representatives:

Axis	Slew Rate Requirement
Azimuth	90 deg/s in 45 deg
Elevation	60 deg/s in 10 deg

These specifications were not previously required for the M60A1 actuators. A mathematical model of the hardware was thus derived (Section 4.3), and from this model the appropriate servovalve, accumulator volume, transmission gear ratio, and motor displacement were selected. The selection results for the elevation and azimuth axes are given in Sections 4.4 and 4.5, respectively. A total pump flow capacity of 30 gpm was assumed in this study.

Upon completion of the study, the hardware was jointly selected by Bendix and personnel from Rock Island Arsenal, for inclusion in the subsequent computer study.

**4.2 HARDWARE CONSIDERATIONS AND ASSUMPTIONS**

An actuator is considered optimal if it achieves a maximum turret or gun acceleration. If the actuator displacement is too small, the actuator torque will not be great enough to rapidly accelerate the load inertia. If the actuator displacement is too large, the pump will not provide sufficient flow to reach the desired speed. It follows that an optimum actuator displacement exists which is small enough to obtain the desired speed with the available pump flow, and large enough to obtain rapid acceleration. Likewise, the servovalve flow area should be large enough so that significant pressure drop does not occur across the servovalve during acceleration. A model for the actuator system is illustrated in Figure 4-1.

For the purpose of this analysis, it was assumed that a variable delivery constant pressure pump would be used. The schematic for this type of pump is shown in Figure 4-2. The pump servovalve passes fluid

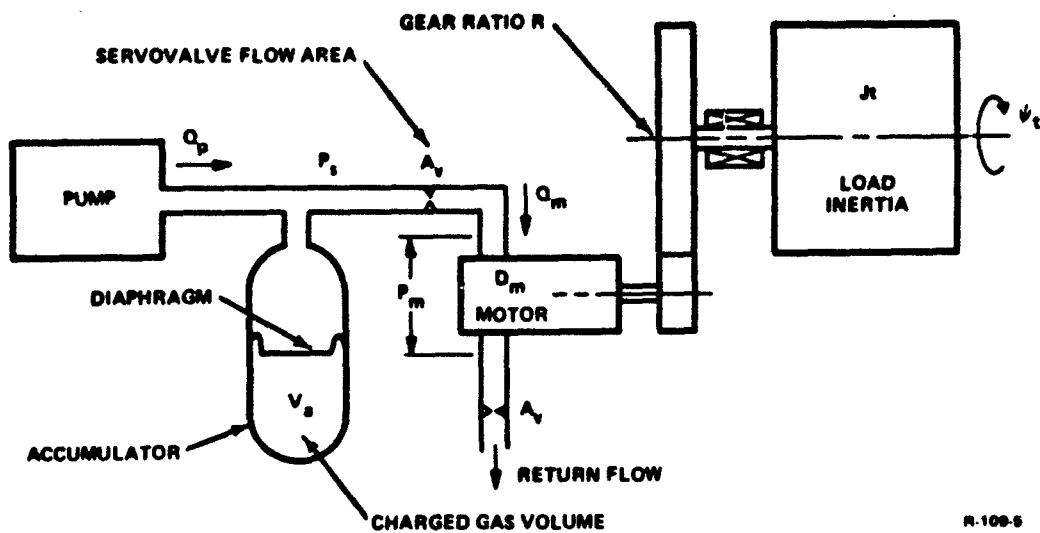


Figure 4-1. Model for Actuator Systems

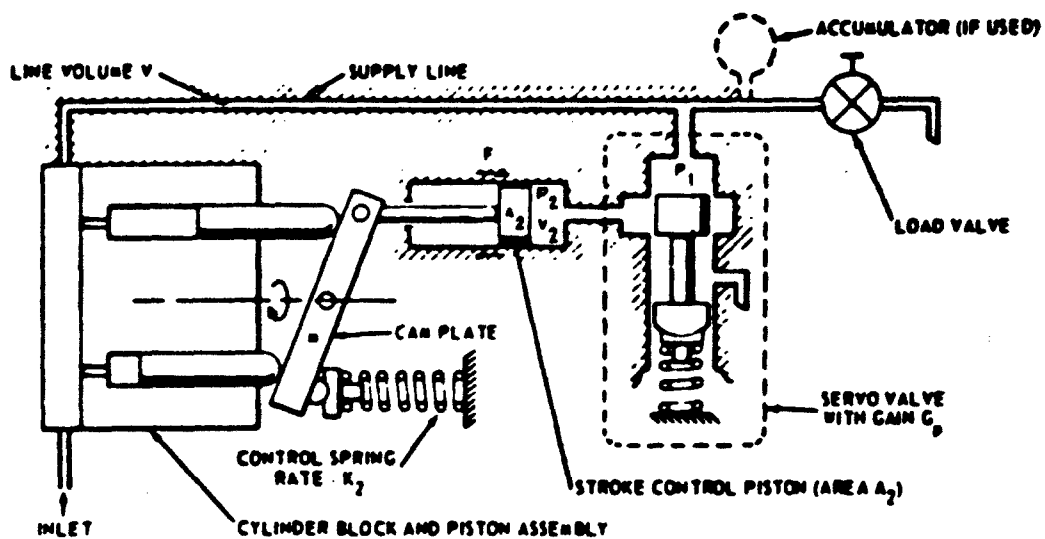


Figure 4-2. Pump Control System Schematic

either to or from the control piston, depending on the sign of the pressure error. If the pressure is higher than the set pressure, the valve passes fluid from the main circuit into the stroking piston cylinder, thus decreasing the delivery of the pump so that the pressure will be reduced. On the other hand, if the pressure is low, the valve passes fluid to the pump case and the control spring forces the pump displacement to increase.

Thus, as the flow demand changes, the pump displacement automatically changes to maintain the supply pressure constant, within the flow range of the pump.

The following hardware and fluid properties were neglected, since they would not affect the choice of the optimum actuator system.

- Pump dynamics
- Fluid leakage
- Fluid compressibility
- Friction

The pump dynamics and the compressibility lag are generally sufficiently fast to have negligible effect on acceleration. Use of an accumulator slows the pump response. This effect was neglected for this study. The addition of pump dynamics would add considerably to the complexity of the simulation.

The turret and gun acceleration achieved in practice will be a little less than the values obtained in the study due to the effects of friction and leakage.

The parameters which must be optimized to obtain the best performance are the product of transmission ratio and motor displacement  $RD_m$ , the maximum servovalve flow area ( $A_{v \max}$ ), and accumulator volume  $V_a$ . The product  $RD_m$  is the motor displacement reflected to the turret or gun, respectively.

#### 4.3 MATHEMATICAL MODEL OF THE ACTUATOR SYSTEM

The pump model assumed here provides that pump flow  $Q_p$  equal to that of the motor  $Q_m$  below the maximum pump flow in accordance with:

$$\begin{aligned} Q_p &= Q_m, & \text{for } Q_m < 20 \text{ gpm} \\ Q_p &= 20 \text{ gpm}, & \text{for } Q_m \geq 20 \text{ gpm} \end{aligned} \tag{4-1}$$

The maximum pump flow for a single axis was taken to be 20 gpm. However, a 30 gpm pump should be used to provide the required flow when both axes are activated simultaneously.

It follows for the azimuth axis that equating the motor flow to the servovalve flow yields:

$$RD_m \dot{\psi}_t = CA_v \sqrt{\frac{P_s - P_m}{2}} \quad (4-2)$$

Squaring equation (4-2) and solving for  $P_m$ :

$$P_m = P_s - \left( \frac{RD_m}{CA_v} \right)^2 \dot{\psi}_t^2 \quad (4-3)$$

The turret acceleration is:

$$\ddot{\psi}_t = \frac{RD_m}{100 J_t} P_m \quad (4-4)$$

The accumulator is shown in Figure 4-1. The volume of gas under the diaphragm is initially charged to the supply pressure. The volume above the diaphragm is filled with hydraulic fluid. When the hydraulic fluid is at the desired supply pressure, approximately one-half of the accumulator volume is filled with gas. As long as the servovalve flow does not exceed the pump flow capability, the pump will maintain the desired supply pressure, and the hydraulic fluid and gas volumes will not change. When the valve flow exceeds the pump flow, hydraulic fluid will flow from the accumulator, the gas volume will expand, and the supply pressure will drop. This process is adiabatic and it follows that:

$$P_s V_a^k = C \quad (4-5)$$

Differentiating with respect to time:

$$k P_s V_a^{k-1} \dot{V}_a + V_a^k \dot{P}_s = 0$$

Solving for  $\dot{P}$ :

$$\dot{P}_s = \frac{kP_s \dot{V}_a}{V_a} \quad (4-6)$$

The rate of change of gas volume equals the difference between the motor flow and pump flow.

$$\begin{aligned} \dot{V}_a &= Q_m - Q_p, \text{ when } Q_m > Q_p \\ \dot{V}_a &= 0, \text{ when } Q_m < Q_p \end{aligned} \quad (4-7)$$

For a hydraulic motor, the displacement flow is given by:

$$Q_m = RD_m \dot{\psi}_t \quad (4-8)$$

The gas volume in the accumulator and the supply pressure are:

$$V_a = V_o + \int \dot{V}_a dt \quad (4-9)$$

where

$V_o$  = steady state gas volume in accumulator at 3000 psi

$$P_s = P_{so} + k \int P_s \frac{\dot{V}_a}{V_a} dt \quad (4-10)$$

Equations (4-1), (4-3), (4-4), and (4-7) through (4-10) were used for the computer study to determine the actuator system sizing. The same set of equations was used for the elevation axis, with only a change in parameters and a substitution of  $\theta_g$  for  $\psi_t$ .

Table 4-1 lists the actuator system parameters.



Table 4-1. Parameters for Hardware Selection Study

Parameter	Value	Units
C	100	$\text{in/s}\sqrt{\text{psi}}$
$J_g$	3800	slugs
$J_t$	22,700	slugs
k	1.4	—
$P_{so}$	3000	psi
$V_o$	one-half accumulator volume	$\text{in}^3$

#### 4.4 ELEVATION AXIS HARDWARE SELECTION

Results of the hardware selection procedure for the elevation axis are summarized in Figure 4-3. Gun elevation rate achieved at an elevation angle displacement of 10 deg is plotted for three different servovalve areas and numerous values of the product of transmission ratio R and motor displacement  $D_m$ . It is evident from Figure 4-3 that with a servovalve area of  $0.034 \text{ in}^2$ , the gun elevation rate exceeds 60 deg/s in 10 deg of displacement for values of  $RD_m$  between 50 and 110. For maximum rate, a value of 80 for  $RD_m$  was selected for elevation.

Since servovalves are generally rated in terms of gpm flow at a pressure drop of 1000 psi instead of effective flow area, an appropriate conversion is achieved by use of Figure 4-4. This figure also indicates which commercially available valves can handle a given application. The flow area of  $0.034 \text{ in}^2$  is thus equivalent to a 20 gpm servovalve, and the largest available MOOG series 30 valve can be utilized for this application. A 20 gpm servovalve was therefore selected. A MOOG Series 30 valve was also used in an actuator system utilized in the past by the Chrysler Corporation for the M60A1 main gun.

A time history plot for the gun elevation angle, the elevation rate, and the accumulator gas volume is contained in Figure 4-5. As indicated by this figure, the accumulator was not used in the first 10 deg of elevation displacement. Therefore, an accumulator is not required for elevation. An elevation accumulator may be of value, however, when both the gun and turret are rotated simultaneously.

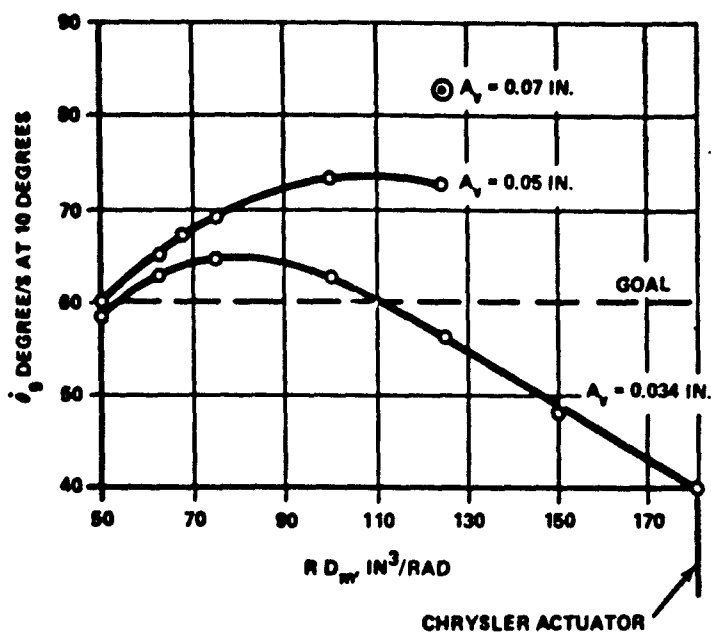


Figure 4-3. Hardware Selection Results for Elevation Axis

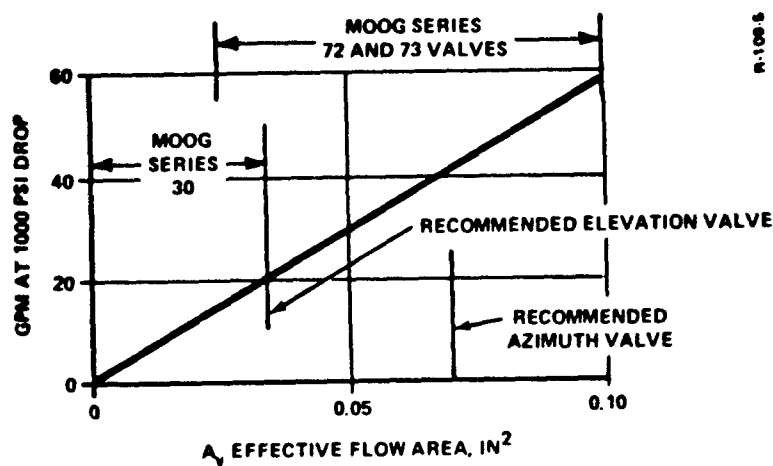


Figure 4-4. Servo valve Conversion Diagram

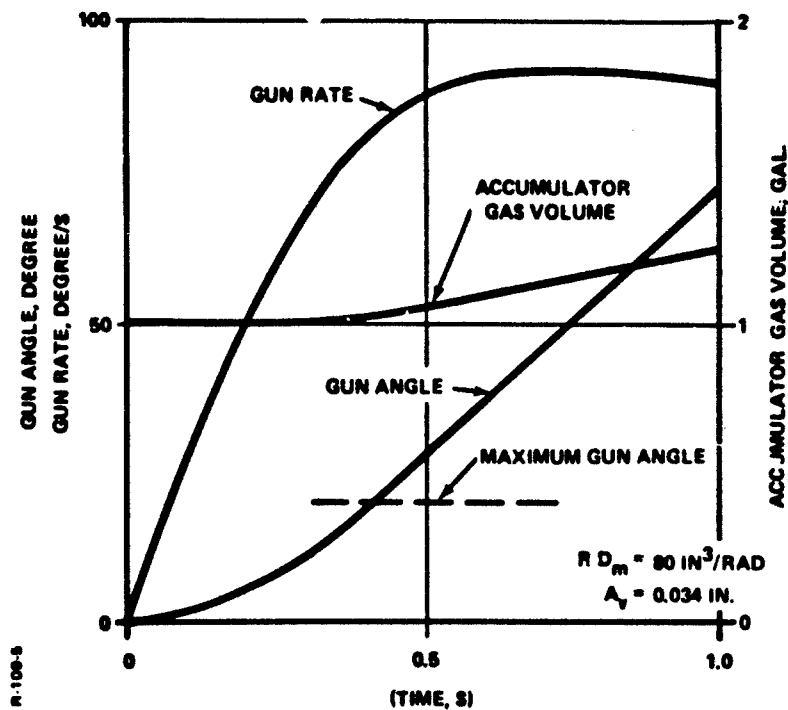


Figure 4-5. Time History Plot for Elevation Angle, Rate, and Gas Volume

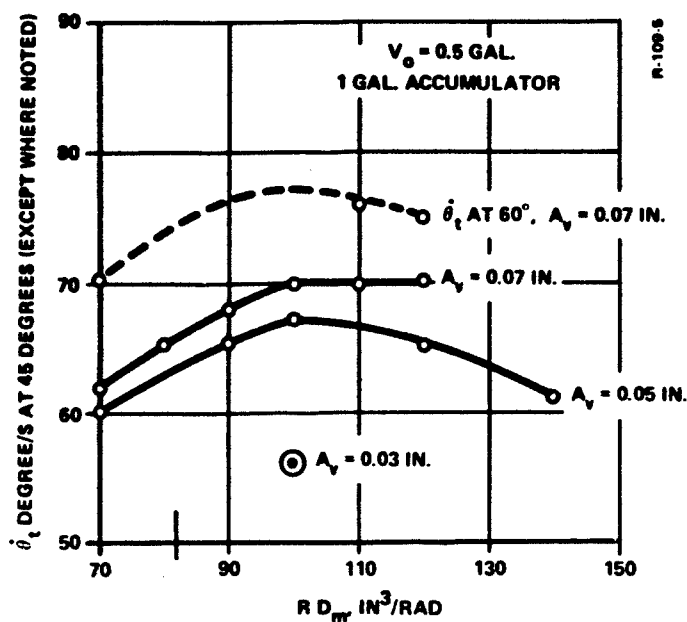


Figure 4-6. Hardware Selection Results for Azimuth Axis

#### 4.5 AZIMUTH AXIS HARDWARE SELECTION

Results of the hardware selection for the azimuth axis are summarized in Figure 4-6. Turret azimuth rate achieved at an azimuth displacement angle of 40 deg is plotted for three different servovalve areas and numerous values of the product  $RD_m$ . It is evident from Figure 4-6 that a rate of 90 deg/s in 45 deg displacement is not possible. It was therefore decided to choose the valve and motor combination which maximizes rate at 45 deg displacement.

In accordance with Figure 4-6, a turret speed of 70 deg/s is obtained with  $A_v$  equal to  $0.07 \text{ in}^2$  and  $RD_m$  values of 100 to  $120 \text{ in}^3/\text{rad}$  using a 1 gallon accumulator. Therefore, these are the recommended values.

Figure 4-7 demonstrates the effect of reducing or increasing the accumulator volume from 1 gallon. Increasing the accumulator volume above 1 gallon has no effect, while reducing the volume has only a minimal effect.

Figure 4-8 shows the turret rotation required to accelerate to a speed of 90 deg/s. If this criterion had been used, an  $RD_m$  of 90 and a 2 gal accumulator would have been selected. However, an  $RD_m$  of 100 appears to be the best overall choice when allowance is made for torque loss due to friction.

As indicated by Figure 4-4, the selected  $0.07 \text{ in}^2$  servovalve flow area is equivalent to 40 gpm flow at 1000 psi pressure drop. This is within the range of the MOOG 72 and 73 flow control servovalves.

A time history plot for the turret azimuth angle, the turret azimuth rate, and the accumulator gas volume is contained in Figure 4-9 for the selected values of  $RD_m$  and  $A_v$ .

A curve for each of the azimuth and elevation axes, which shows the range of values available for  $R$  and  $D_m$ , is contained in Figure 4-10. The product  $RD_m$  was determined to be 100 and 80 for the azimuth and elevation axes respectively.

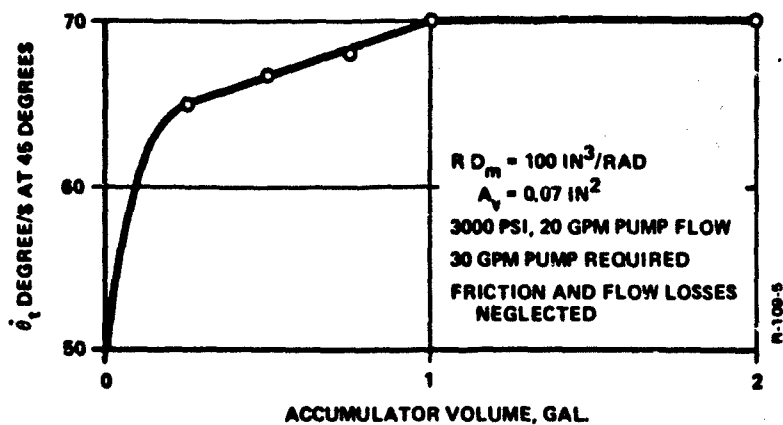


Figure 4-7. Accumulator Sizing Effects Diagram

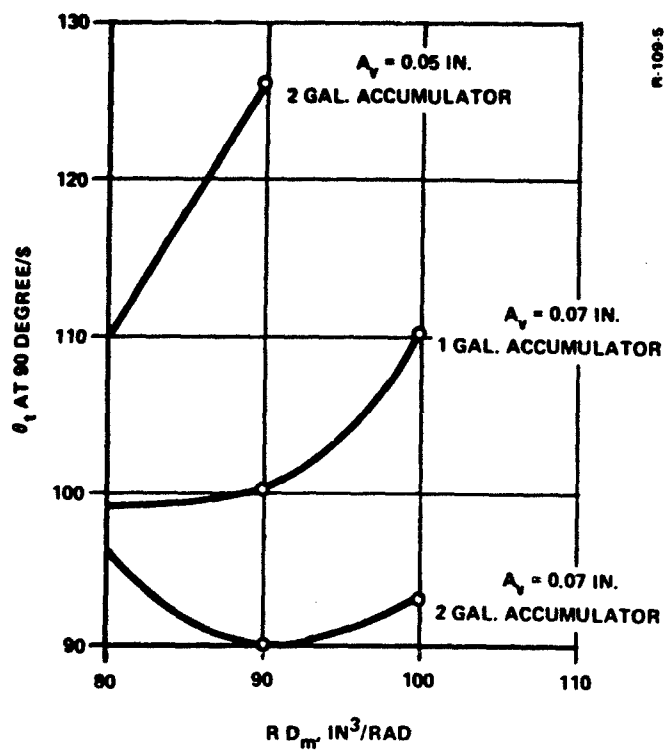


Figure 4-8. Azimuth Rotation Requirements Plot

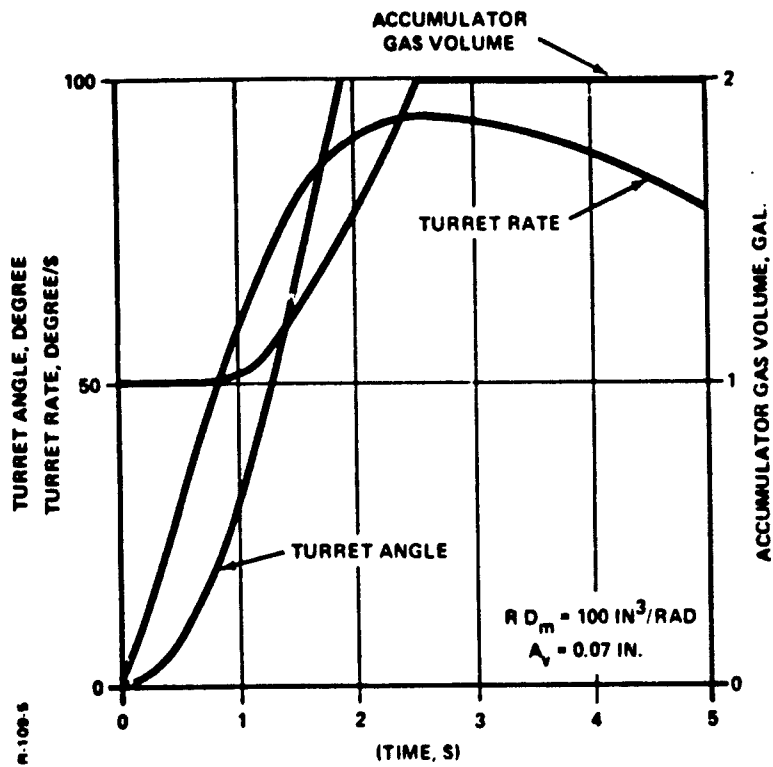


Figure 4-9. Time History Plot for Azimuth Angle, Rate, and Gas Volume

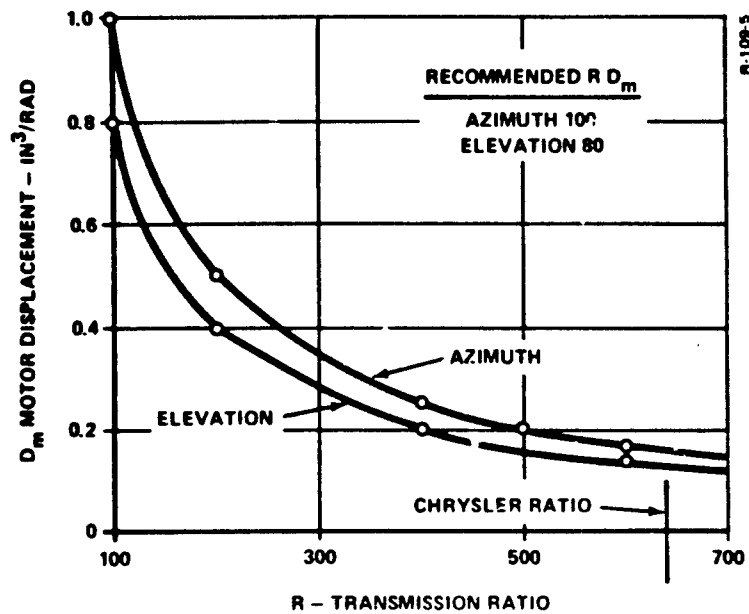


Figure 4-10. Curves of Constant Products of Transmission Ratio and Motor Displacement for Azimuth and Elevation Axes

## SECTION 5

### ANALYTICAL SYSTEM ANALYSIS

#### 5.1 GENERAL

This section contains an extensive analytical analysis of the rate and position control systems which was performed prior to the simulation analysis. By means of this study, it was possible to gain valuable insight into the behavior of the system and to derive starting values for the rate, proportional, and integral gains of the control law. In addition, two methods were derived to reduce system tracking errors. An analytical sensor error study and an outline of a coupling model are also presented.

For the purpose of this analytical study, the control system was linearized and simplified, where possible, while retaining basic system functions. This simplification of the system is described in Sections 5.2.1 and 5.2.2 for the rate and the position control concepts, respectively. Transfer functions for the system described here showed how the systems should be designed to achieve the desired performance.

Initial gain values for the nonlinear computer models were also determined from the linear study. These gain values were varied, and compensation was used to achieve the desired performance with the nonlinear flow equations, sensor dynamics, and other nonlinear effects and lags.

#### 5.2 LINEAR SYSTEM DERIVATION

##### 5.2.1 Rate Control System

This section contains the derivation of the linearized rate control system shown in Figure 5-1. The following nonlinearities were omitted in order to simplify the transfer function:

- (1) Friction
- (2) Sensor dynamics
- (3) Compliance

Since pressure feedback dynamics have negligible phase shift at the natural frequency, this effect can be represented by a constant as shown in Figure 5-1.

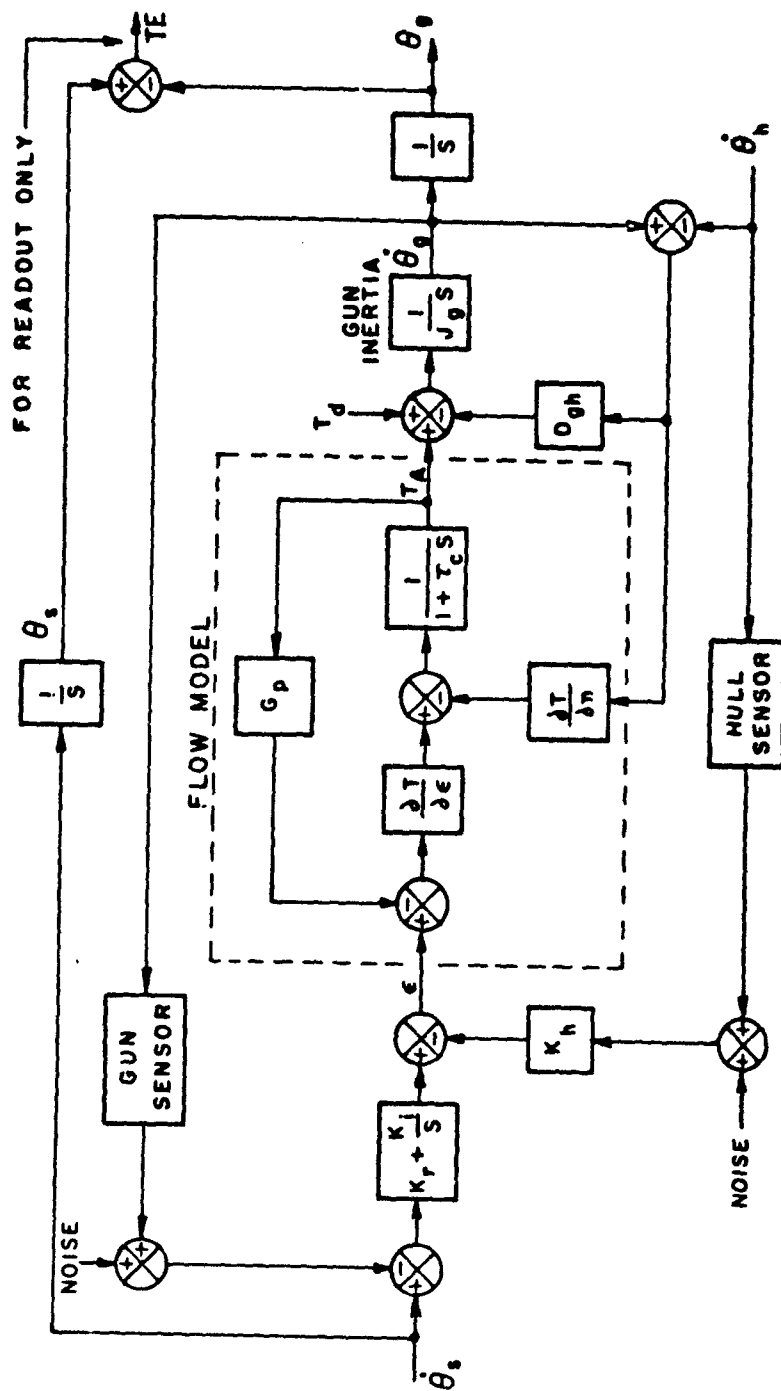


Figure 5-1. Block Diagram of Linearized Rate Control System



The nonlinear flow and torque equations from Figure 2-6 are given by:

$$CA_v \sqrt{\frac{P_s - P_m}{2}} = RD_m (\dot{\theta}_g - \dot{\theta}_h) + LP_m + \frac{V}{2B} sP_m \quad (5-1)$$

$$T_A = RD_m P_m \quad (5-2)$$

Linearizing by partial differentiation yields:

$$C \sqrt{\frac{P_s - P_o}{2}} \Delta A_v - \frac{CA_{vo}}{2 \sqrt{2(P_s - P_o)}} \Delta P_m = RD_m s (\Delta \theta_g - \Delta \theta_h) + L \Delta P_m + \frac{V}{2B} s \Delta P_m \quad (5-3)$$

Substituting for  $P_m$  from equation (5-2) and rearranging terms gives:

$$C \sqrt{\frac{P_s - P_o}{2}} \Delta A_v = \frac{Q_o}{2RD_m (P_s - P_o)} \Delta T_A + \frac{V}{2BRD_m} s \Delta T_A + RD_m s (\Delta \theta_g - \Delta \theta_h) \quad (5-4)$$

where

$$Q_o = CA_{vo} \sqrt{\frac{P_s - P_o}{2}} + 2L (P_s - P_o)$$

Rearranging terms, and defining new parameters gives:

$$T_A = \frac{1}{1 + \tau_c s} \left[ \frac{\partial T}{\partial \epsilon} \epsilon - \frac{\partial T}{\partial n} (\dot{\theta}_g - \dot{\theta}_h) \right] \quad (5-5)$$

where

$$\begin{aligned} \epsilon &= A_v \\ \tau_c &= \frac{v}{Q_o} \frac{(P_s - P_o)}{B} \\ \frac{\partial T}{\partial n} &= 2 (RD_m)^2 \frac{(P_s - P_o)}{Q_o} \\ \frac{\partial T}{\partial \epsilon} &= \frac{2CRD_m (P_s - P_o)}{CA_{vo} + 2L \sqrt{2 (P_s - P_o)}} \end{aligned}$$

Assuming  $T_d = T_{fg} = 0$ , gun acceleration is given by:

$$\ddot{\theta}_g = \frac{T_A}{J_g} \quad (5-6)$$

The control law developed in Section 3.2 is:

$$\epsilon = \left( K_r + \frac{K_1}{s} \right) (\dot{\theta}_s - \dot{\theta}_g) - K_h \dot{\theta}_h \quad (5-7)$$

Upon solving equations (5-4) through (5-7), the system transfer function is:

$$\frac{\dot{\theta}_g}{\dot{\theta}_s} = \frac{1 + \frac{K_r}{K_1} s}{1 + \frac{s}{\beta \omega_{NS}} + \frac{\alpha s^2}{\beta \omega_{NS}^2} + \frac{s^3}{\beta \omega_{NS}^3}} \quad (5-8)$$

where

$$\omega_{NS}^2 = \frac{1}{J_g \tau_c} \left( \frac{\partial T}{\partial n} + K_r \frac{\partial T}{\partial \epsilon} \right) \quad (5-9)$$

$$\beta\omega_{NS} = \frac{K_1 \frac{\partial T}{\partial \epsilon}}{\frac{\partial T}{\partial n} + K_r \frac{\partial T}{\partial \epsilon}} \quad (5-10)$$

$$\alpha\omega_{NS} = \frac{1 + K_p \frac{\partial T}{\partial \epsilon}}{\tau_c} \quad (5-11)$$

The steady state  $T_d/\dot{\theta}_g$  gain is:

$$\frac{T_d}{\dot{\theta}_g} = \frac{K_1 \frac{\partial T}{\partial \epsilon}}{1 + K_p \frac{\partial T}{\partial \epsilon}} \quad (5-12)$$

The response to terrain input is:

$$\frac{\dot{\theta}_g}{\dot{\theta}_h} = \frac{\frac{\partial T}{\partial n} - K_h \frac{\partial T}{\partial \epsilon}}{K_1 \frac{\partial T}{\partial \epsilon}} \cdot \frac{1}{G_c} \quad (5-13)$$

where

$G_c$  = cubic denominator of the transfer function

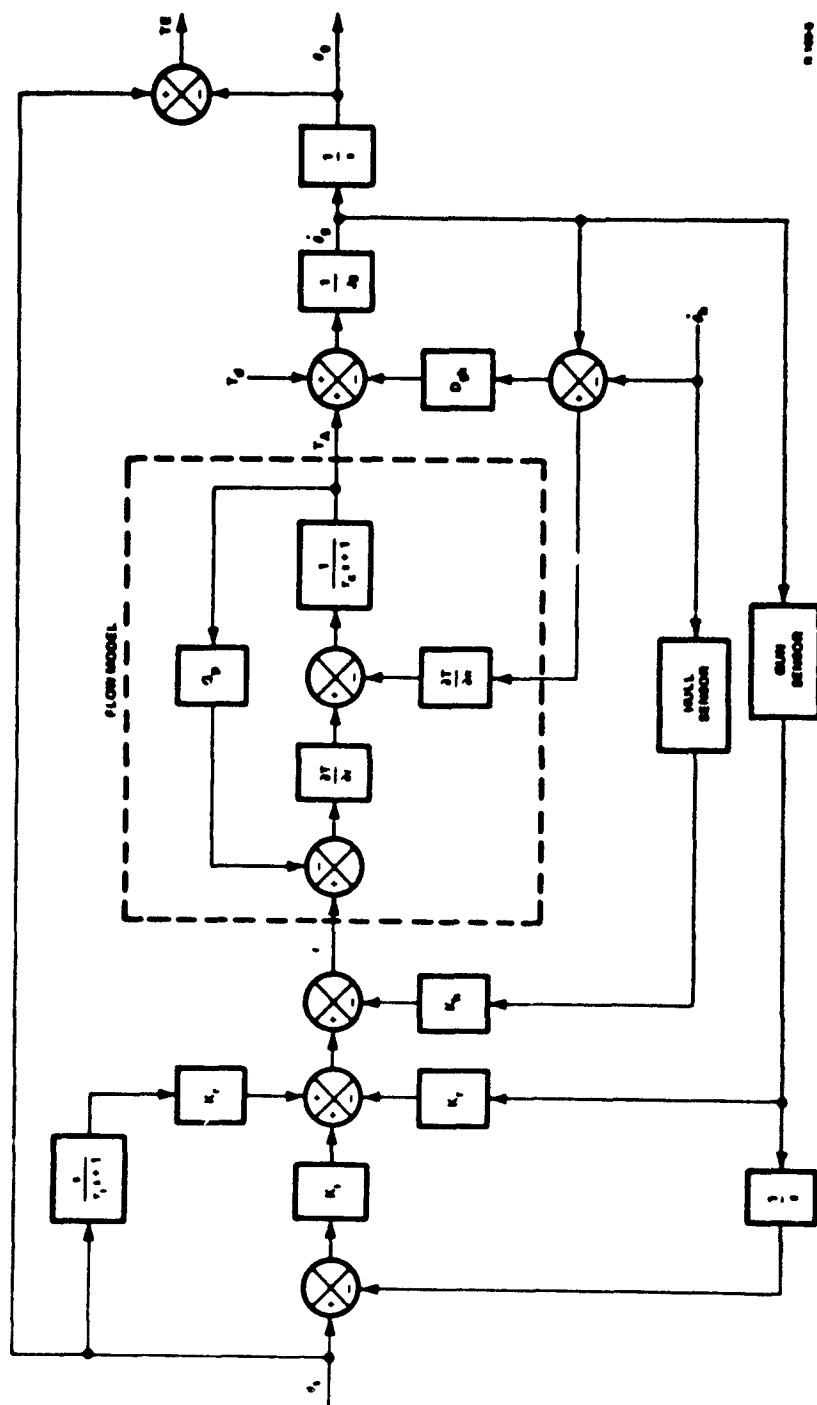
It follows that the gun motion, or tracking error, for a terrain input is zero if

$$K_h = \frac{\partial T/\partial n}{\partial T/\partial \epsilon}$$

However, since the torque speed slope is nonlinear, perfect cancellation over the complete range of operation may not be possible.

### 5.2.2 Position Control System

This section contains a derivation of the linearized elevation position control system shown in Figure 5-2. As indicated in the figure, the inertial gun position feedback signal is obtained by integrating



**Figure 5-2. Block Diagram of Linearized Position Control System**

the gun rate sensor signal. A potentiometer was not used for the position feedback because it would have been necessary to add the hull position to the potentiometer signal to obtain the inertial position signal. This would have made the system very sensitive to hull sensor gain variation.

The control law for this system is given by:

$$c = K_i (\theta_s - \theta_g) + K_r (\dot{\theta}_s - \dot{\theta}_g) - K_h \dot{\theta}_h$$

Since a position command input will generally be used for the position command system, the command rate signal  $\dot{\theta}_s$  will not be readily available. This signal can be omitted, or it can be obtained by pseudo differentiation of the input command signal as shown below.

$$\dot{\theta}_s = \frac{s}{1 + \tau_i s} \theta_s$$

Including the  $\dot{\theta}_s$  term results in faster response to command inputs.

Other equations for the position control concept take the same form as those for the rate control concept.

It follows therefore that the position control transfer function is given by:

$$\frac{\theta_g}{\theta_s} = \frac{1 + \frac{K_r}{K_i} \cdot \frac{1}{1 + \tau_i s} s}{1 + \frac{s}{\beta \omega_{NS}} + \frac{\alpha s^2}{\beta \omega_{NS}^2} + \frac{s^3}{\beta \omega_{NS}^3}}$$

The steady state response to a hull rate step input is:

$$\dot{\theta}_g = \frac{\frac{\partial T}{\partial n} - K_h \frac{\partial T}{\partial \epsilon}}{K_i \frac{\partial T}{\partial \epsilon}} \dot{\theta}_h$$

The transfer functions for hull rate and disturbance torque inputs are the same as for the rate control system. Therefore, the response to these inputs is identical for the two systems.

The response to identical rate command inputs will be the same for both systems if the pseudo differentiator lag  $\tau_1$  equals zero. This condition was used in the simulation for showing the equivalence of the two systems, as discussed in Section 6.3. Since position command inputs will generally be used with the position control system, the response time to rotate the gun to a commanded angle will be shorter whether or not the input rate signal  $\theta_g$  is used.

### 5.2.3 Linear Model Parameters

For model development and preliminary studies, the actuator displacement and volume under compression were obtained from the bibliography. The parameters developed here corresponded to the elevation axis:

$$RD_m = LA_p = 38.4 \text{ in.} \times 4.72 \text{ in}^2 \times 16.37 \frac{\text{cm}^3}{\text{in}^3}$$

where

$$RD_m = 2970 \text{ cm}^3/\text{rad}$$

$$V = 895 \text{ cm}^3$$

The supply pressure and maximum servovalve area were taken to be

$$P_s = 210 \text{ kg/cm}^2$$

$$A_{v \text{ max}} = 0.2 \text{ cm}^2$$

The linearized actuator parameters are:

$$\frac{\partial T}{\partial n} = \frac{2}{100} (RD_m)^2 \frac{P_s - P_o}{Q_o}$$

$$\tau_c = \frac{V}{B} \frac{P_s - P_o}{Q_o}$$

$$\frac{\partial T}{\partial \epsilon} = \frac{2}{100} RD_m \frac{P_s - P_o}{A_{vo}}$$

The effective servovalve area  $A_{vo}$  and pressure  $P_o$  are typically taken to be equal to one-half the maximum servovalve area and one-half supply pressure for linear analysis. Thus, the linearized servovalve parameters are:

$$Q_o = CA_{vo} \sqrt{\frac{P_s - P_o}{2}} = 955 \times 0.1 \sqrt{\frac{210 - 105}{2}}$$

giving

$$Q_o = 695 \text{ cm}^3/\text{s}$$

Also

$$\frac{\partial T}{\partial n} = \frac{2}{100} (2970)^2 \frac{105}{695} = 2.66 \times 10^4 \text{ kg-m-s}$$

$$\tau_c = \frac{895}{10,550} \times \frac{105}{695} = 0.013 \text{ s}$$

$$\frac{\partial T}{\partial \epsilon} = 0.02 \times 2970 \times \frac{105}{0.1} = 6.25 \times 10^4 \text{ kg-m/rad}$$

Generally, the linearized parameters must be varied somewhat to obtain the best correlation with the nonlinear system. To achieve this correlation, the following values were used:

$$\frac{\partial T}{\partial n} = 2.08 \times 10^4 \text{ kg-m/s}$$

$$\tau_c = 0.016 \text{ s}$$

$$\frac{\partial T}{\partial \epsilon} = 6.25 \times 10^4 \text{ kg-m/rad}$$

The initial gain values for  $K_1$ ,  $K_r$  and  $K_p$  were determined as follows. The bandpass goal was 15 Hz. Thus, a bandpass of 16 Hz was selected for the linear system, to allow some performance margin. From equation (5-9)

$$K_r = \frac{1}{\partial T / \partial \epsilon} \left( \tau_c J \omega_{NS}^2 - \frac{\partial T}{\partial n} \right)$$

and for

$$\omega_{NS} = 2\pi \times 16 = 100 \text{ rad/s}$$

it follows that:

$$K_r = \frac{1}{6.25 \times 10^4} (0.016 \times 527 \times 100^2 - 2.08 \times 10^4) = 1.0$$

From equation (5-10)

$$K_1 = \beta \omega_{NS} \left( K_r + \frac{\partial T / \partial n}{\partial T / \partial \epsilon} \right)$$

In this expression,  $\beta$  should equal 0.35 for the best linear system response, and  $\alpha$  should equal 0.7. Thus:

$$K_1 = 0.35 \times 100 \left( 1.0 + \frac{2.08}{6.25} \right) = 47$$

Equation (5-11) specifies that:

$$K_p = \frac{1}{\partial T / \partial \epsilon} (\alpha \tau_c \omega_{NS} - 1)$$

Upon substituting numerical values, it follows that:

$$K_p = \frac{1}{6.25 \times 10^4} (0.016 \times 0.7 \times 100 - 1) = 2 \times 10^{-6}$$



#### 5.2.4 Linear System Response

Based on the transfer functions presented in Sections 5.2.1 and 5.2.2, a digital computer solution for time and frequency response was obtained. Using ideal sensors and the parameters for the elevation axis, the resulting time response is illustrated in Figure 5-3. Note that the time response exhibits a 7 percent overshoot. A solution for system frequency response is illustrated in Figure 5-4. This figure shows that a bandpass of 18 Hz was achieved.

#### 5.3 TRACKING ERROR REDUCTION

This section deals with the method of reducing the gun tracking error due to motions of the hull. Referring to Figure 5-5, which is a simplified block diagram of the rate command stabilization system with proportional and integral control, assume a constant hull rate input  $\dot{\theta}_h$  and no command input ( $\dot{\theta}_g = 0$ ). In steady state,  $\dot{\theta}_g = 0$  and  $Q_v = Q_m$ . Then:

$$\epsilon = - \frac{\partial T / \partial n}{\partial T / \partial \epsilon} \dot{\theta}_h$$

Since the tracking error TE is given by:

$$TE = \theta_s - \theta_g$$

$$\epsilon = K_r (\dot{\theta}_s - \dot{\theta}_g) + K_i (\theta_s - \theta_g)$$

then the steady state tracking error is:

$$TE = -\theta_g = \frac{\epsilon}{K_i} = - \frac{\partial T / \partial n}{K_i \partial T / \partial \epsilon} \dot{\theta}_h$$

It is desired to reduce this tracking error so that the gun is stabilized with respect to hull motions.

Two methods can be used to reduce the tracking error: The first method is a cancellation technique. Equation (5-13) shows that hull motion will cause no gun motion, and thus will result in no tracking error, if

$$K_h = \frac{\partial T / \partial n}{\partial T / \partial \epsilon}$$

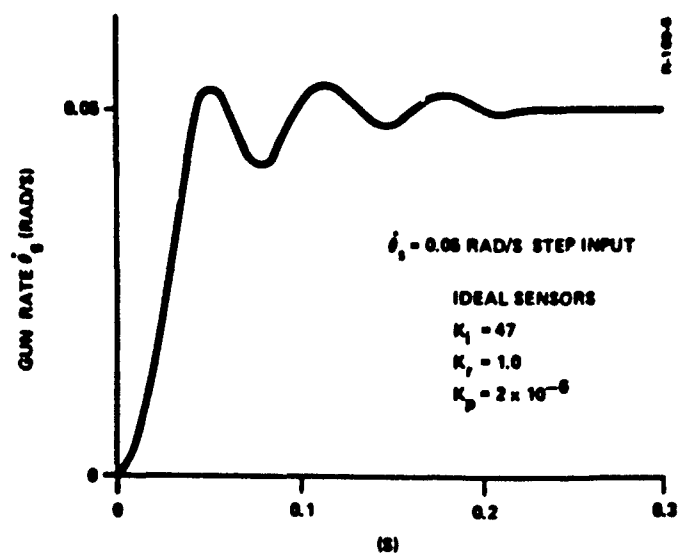


Figure 5-3. Linear System Step Response

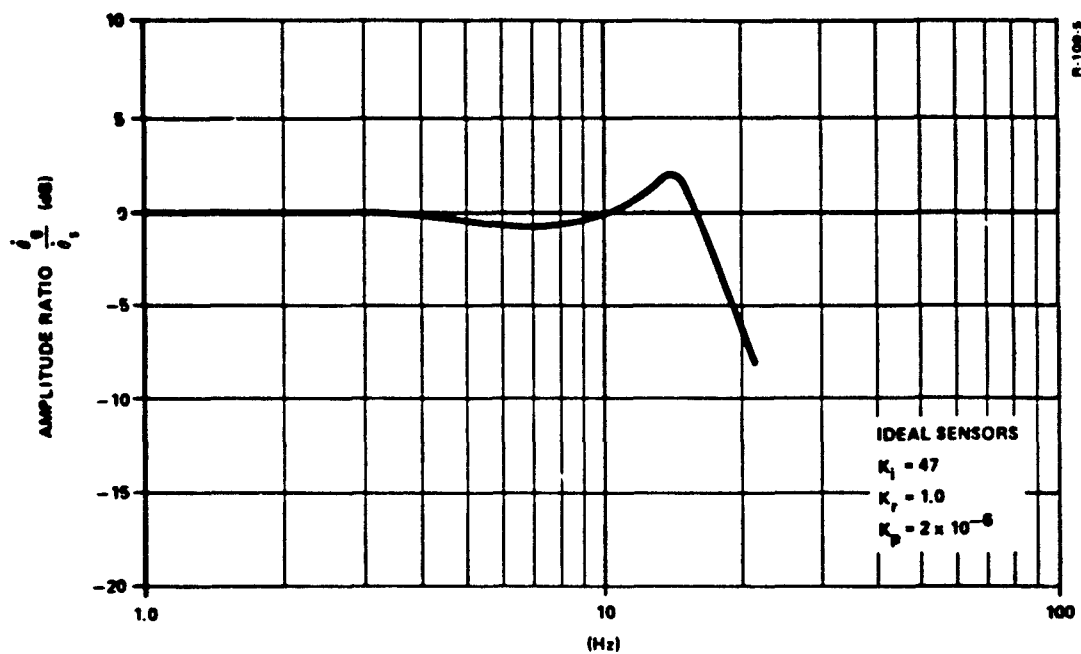


Figure 5-4. Linear System Frequency Response

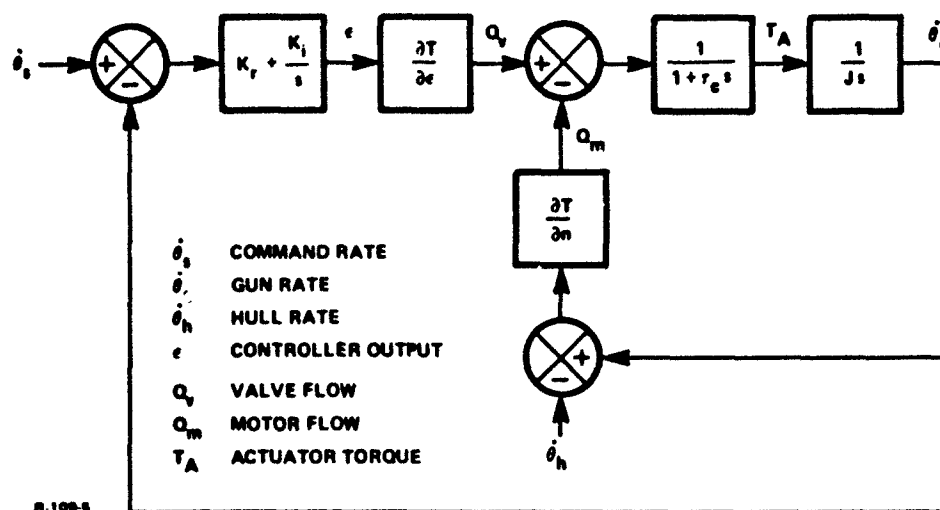


Figure 5-5. Simplified Block Diagram of the Rate Command System

Therefore, hull motion can be effectively cancelled by using the output signal of an inertial hull sensor with gain  $K_h$  in the control law. Complete cancellation of hull motion coupling for the actual system, however, cannot be obtained due to nonlinearities and compressibility. In addition, gain variations due to such factors as aging and temperature will cause some tracking error. The cancellation technique, however, is very effective in reducing gun motion.

The second method uses only a single gun sensor. No vehicle motion signal input is provided for this system. The tracking error for the system will be inversely proportional to the error signal amplification and, therefore, a very high gain must be used. A lag-lead network is required to lower the high frequency gain for stability. The design technique is to first adjust the gains for the stable operation with the desired response. The gain required to lower the tracking error to an acceptable level is then determined. The lag-lead network is designed to maintain the high frequency gains at the original values for stable operation, while a higher gain is obtained at low frequency for low tracking error.

The higher gain and lag-lead network can also be used with the cancellation technique to reduce the sensitivity to variation of some of the gains.

## 5.4 SENSOR ERROR ANALYSIS

### 5.4.1 General

The effect of sensor errors on the control and stabilization system is the subject of this section. Sensor errors considered here are gain errors and offset errors. Only the rate command system is analyzed in this section. It is assumed that in both the rate and position command systems, the gunner will zero the tracking error with or without sensor errors. On that basis, sensor errors have an effect only on stabilizing the gun in the presence of hull motions. As described in Section 5.2.2, the response to hull motions is identical for both the rate and the position command systems.

The analysis presented here will consider a system utilizing:

- (1) Two rate sensors (Section 5.4.2)
- (2) A single acceleration sensor in the gun axis (Section 5.4.3)

### 5.4.2 Two Rate Sensors

Figure 5-6 is a simplified block diagram of the rate control system with two rate sensors. The transfer function of the sensors is assumed to be  $Y(s)$  with unity gain. For zero sensor gain error,  $K_{bh} = K_{bg} = 1$ . The sensor offset errors are  $\epsilon_h$  and  $\epsilon_g$ .

A hull sensor gain error has the same effect as a change in  $K_h$ . The effects of changes in  $K_h$  were investigated using the simulation and are described in Section 6.

A hull sensor offset error will give a constant steady state tracking error. This is determined as follows: Assuming that  $\theta_s = \theta_h = 0$  and  $\theta_g = 0$  (in steady state), then  $\epsilon = 0$  and  $-K_1 \theta_g = K_h \epsilon_h$ . Since the tracking error  $TE = -\theta_g$ ,

$$TE = \frac{K_h}{K_1} \epsilon_h$$

Typical gain values are  $K_h = 0.15$  and  $K_1 = 50$ . Therefore, the tracking error in radians for an offset error in rad/s is  $TE = 0.003 \epsilon_h$ .

A gun sensor gain error will have little effect on the stabilization system since it only changes the loop gain and is like changing  $K_r$  and  $K_1$ . The effect of a gun sensor offset error  $\epsilon_g$  is analyzed as follows: Assuming  $\theta_s = 0$ , and  $\theta_h = \text{constant}$ , then in steady state,  $\theta_{gs} = 0$ . Therefore,  $\theta_g = -\epsilon_g$ . Since the tracking error is given by  $TE = \theta_s - \theta_g$ , it follows that  $TE = \int \epsilon_g dt$ . The tracking error for a gun

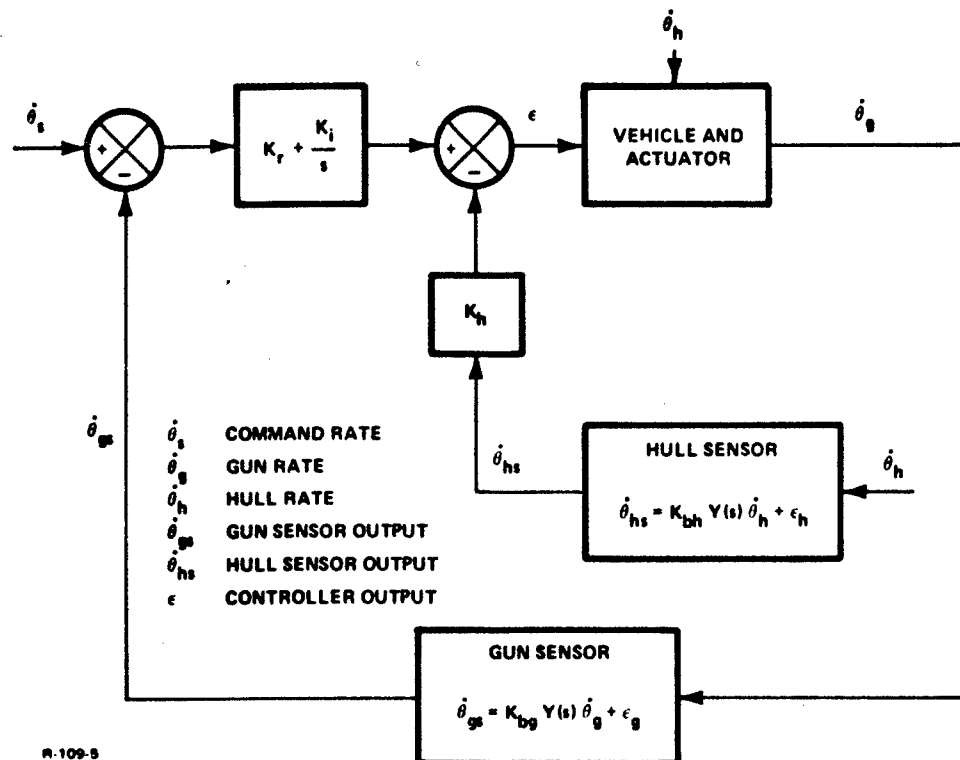


Figure 5-6. Simplified Block Diagram of the Rate Control System with Two Rate Sensors

sensor offset will therefore increase with time. As a result, it is mandatory to minimize the gun sensor offset error or devise a method of eliminating its effect. Section 5.5 discusses a method of eliminating sensor offset errors.

#### 5.4.3 Single Acceleration Sensor

A simplified block diagram of a rate control system with a single acceleration sensor in the gun axis is presented in Figure 5-7. Both an integrating and a pneumatic accelerometer are shown, along with the corresponding rate computation networks.

For both of these acceleration sensors, a gain error (i.e.,  $K_{bg} \neq 1$ ) has the same effect as the rate sensors. The loop gain changes with the gain error, but the influence on the tracking error is small.

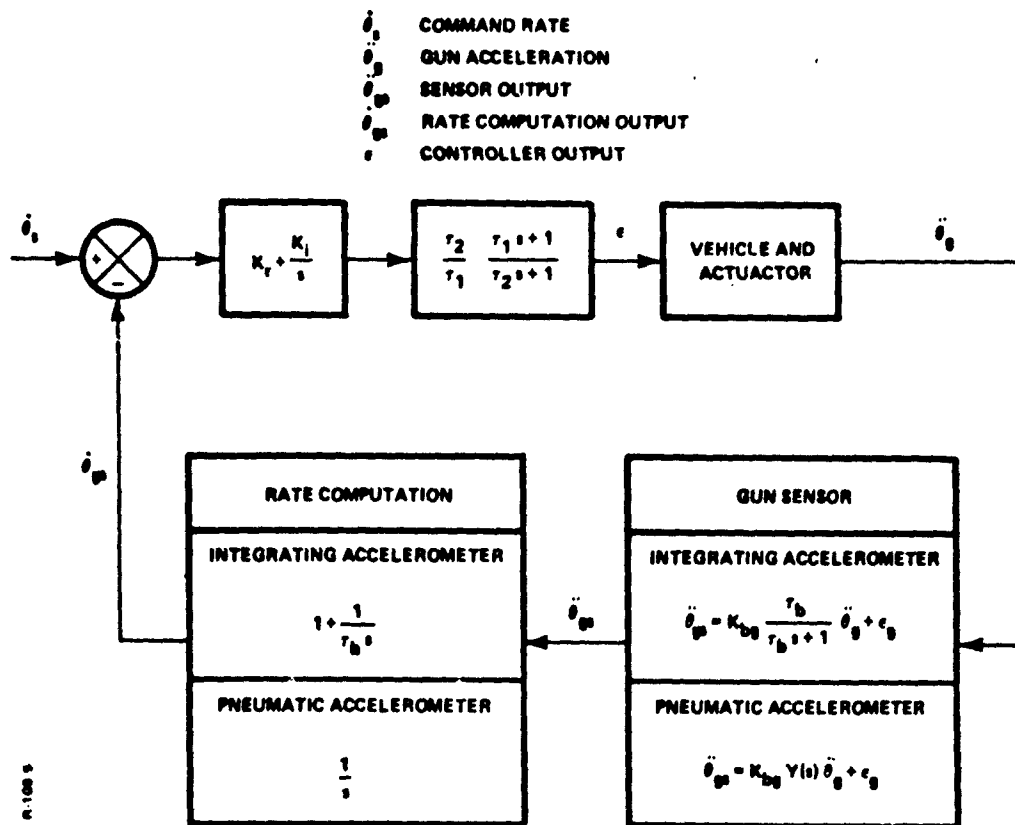


Figure 5-7. Simplified Block Diagram of the Rate Control System with a Single Acceleration Sensor

For the integrating accelerometer, the effect of an offset error is synthesized as follows: In steady state,  $\dot{\theta}_{gs} = 0$ . Therefore,

$$\dot{\theta}_{gs} = 0 = \left( 1 + \frac{1}{\tau_b s} \right) \left( \frac{\tau_b}{\tau_b s + 1} \ddot{\theta}_g + \epsilon_g \right)$$

It follows that

$$\ddot{\theta}_g = -s \epsilon_g - \frac{1}{\tau_b s} \epsilon_g$$

Since the tracking error is  $TE = -\theta_g$ ,

$$TE = \int \epsilon_g dt + \iint \frac{\epsilon_g}{\tau_b} dt dt$$

The tracking error will therefore increase with time for a constant offset error. Its effect, however, can be minimized by increasing the time constant  $\tau_b$ . For the pneumatic accelerometer, the effect of an offset error is determined in a similar manner. In steady state,  $\dot{\theta}_{gs} = 0$  so that

$$\dot{\theta}_{gs} = 0 = \frac{1}{s} (\ddot{\theta}_{gs} + \epsilon_g)$$

The tracking error is therefore

$$TE = \theta_g = \iint \epsilon_g dt dt$$

Again the tracking error will increase with time for a constant sensor offset error.

#### 5.5 METHOD OF ELIMINATING THE EFFECTS OF INTEGRATOR DRIFT AND SENSOR OFFSETS

The sensor error analysis in Section 5.4 shows that gun sensor offsets cause the gun to drift, and hence the tracking error to increase with time. A method has been developed which automatically nulls out the effects of integrator drift and sensor offsets. This method is illustrated in Figure 5-8. The sensor and integrator shown are parts of the original system. The track/store amplifier operates such that the output tracks (equals) the input when in the track mode, and stores (holds) the output when in the store mode.

The amplifier is ordinarily in the store mode so that the output of the sensor  $\dot{\theta}_{gs}$  (gun rate in this case) is integrated to give the gun angle. At the time it is desired to null the sensor offset  $\epsilon_g$  and the integrator drift  $\epsilon_d$ , the track/store amplifier is put into the track mode. This must be accomplished during a period when the actual gun rate  $\dot{\theta}_g$  is zero. Under these conditions, when the network reaches steady state, its output is zero, effectively nulling the offset and drift. Switching the amplifier to the store mode returns the system to normal operation except that now the drift and offset are nulled out.

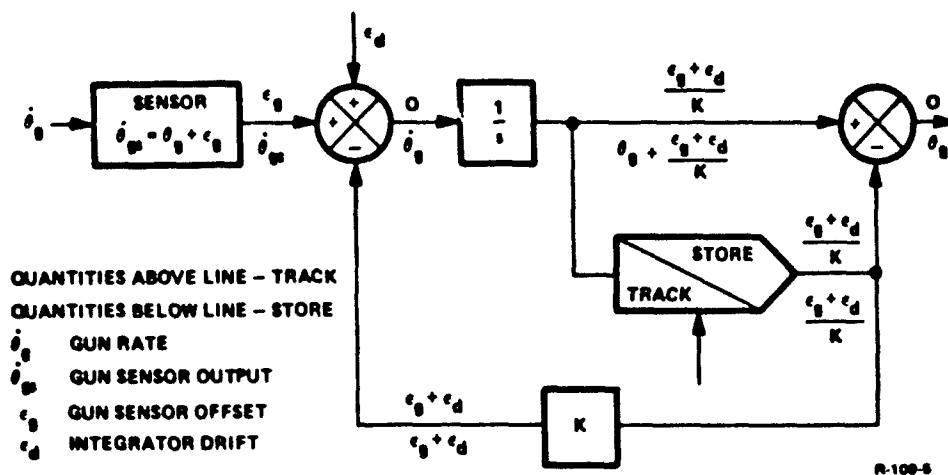


Figure 5-8. Method of Eliminating the Effects of Integrator Drift and Sensor Offsets

Two of these networks are required for an acceleration sensor system while only one is required for a rate sensor system. In addition, the network is required only for the gun sensor since an offset error in the hull sensor results in a small tracking error which does not increase with time.

In order to implement this type of drift and offset elimination, the gunner can be provided with a pushbutton switch which activates all the nulling networks. A timer could also be utilized to automatically return the system to normal operation after the nulling process is completed.

#### 5.6 EFFECT OF GYROS ON RESPONSE

A block diagram for analysis of the effect of rate sensor dynamics on response is presented in Figure 5-9. The transfer function for a command input is

$$\frac{\dot{\theta}_g}{\dot{\theta}_d} = \frac{\left(1 + \frac{K_r}{K_1} s\right) D_g}{1 + \left(\frac{K_r}{K_1} + \frac{D_a D_g}{K_a K_1}\right) s}$$



where

$D_g$  = denominator of gyro transfer function

$K_a$  = gain of actuator and gun transfer function

$D_a$  = denominator of the actuator and gun transfer function

The gyro dynamics, represented by  $D_g$ , add a lead term to the transfer function, which will increase the overshoot to a step input command. The gyro dynamics also add phase lag to the open loop system, and thus will reduce the damping. Therefore, compensation must be used to reduce the overshoot and phase lag resulting from the sensor dynamics.

### 5.7 FORWARD VERSUS FEEDBACK COMPENSATION

Lead-lag compensation must be used to reduce the overshoot and phase lag resulting from the sensor dynamics. In this section, a comparison is made between using this type of compensation in the forward or in the feedback path.

The block diagram in Figure 5-10 shows a system open loop transfer function with a gain term  $K_g$  in the numerator, dynamic terms  $D_g$  in the denominator, and with feedback compensation  $N_c/D_c$ .

The closed loop transfer function is

$$\frac{\dot{\theta}_g}{\dot{\theta}_s} = \frac{K_g D_c}{K_g N_c + D_c D_g}$$

The block diagram in Figure 5-11 shows the same system with forward compensation. The closed loop response is

$$\frac{\dot{\theta}_g}{\dot{\theta}_s} = \frac{K_g N_c}{K_g N_c + D_c D_g}$$

The response with feedback compensation has a lead term equal to the compensation denominator  $D_c$ . The response with forward compensation has a lead term equal to the compensation numerator  $N_c$ . Generally, the smaller the lead time constant, the less the overshoot will be. Therefore, when lead-lag compensation is used, it should be in the feedback path. When lag-lead compensation is used, it should be in the forward path.

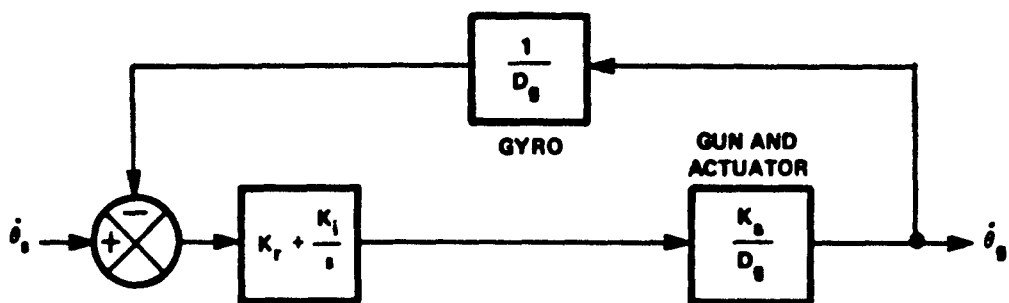


Figure 5-9. Block Diagram for Effect of Rate Sensor Dynamics on Response R-109-8

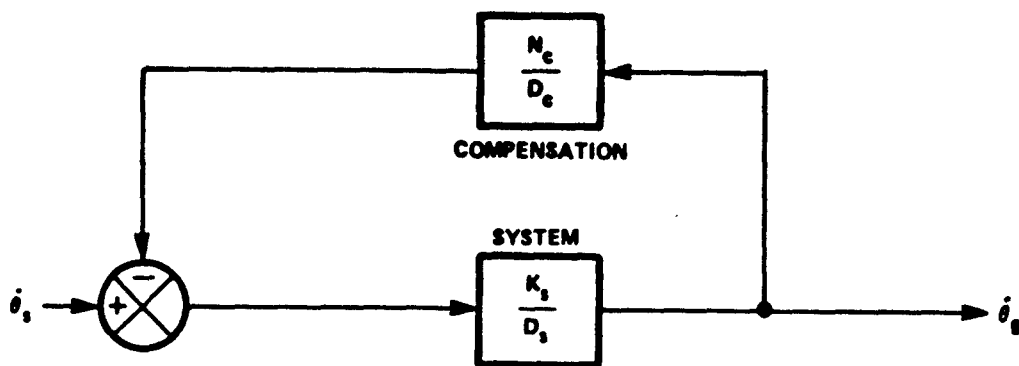


Figure 5-10. Simplified Block Diagram of System with Feedback Compensation

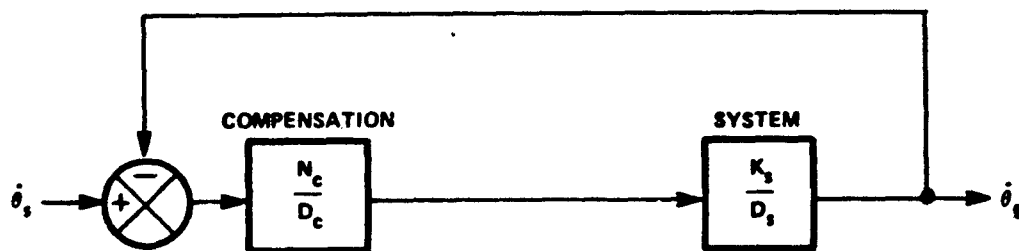


Figure 5-11. Simplified Block Diagram of System with Compensation in Forward Path

## 5.8 SYSTEM STIFFNESS

Stiffness is the reciprocal of the response to a disturbance torque. The static stiffness can be considered as the equivalent spring rate of the system. The tracking error resulting from a constant disturbance torque equals the disturbance torque divided by the static stiffness. The stiffness transfer function is

$$\frac{T_d}{\theta_g} = K_o \frac{1 + a_1 s + a_2 s^2 + a_3 s^3 + a_4 s^4}{(1 + \tau_p s) (1 + \tau_c s) + K_p \tau_p \frac{\partial T}{\partial \epsilon} s^2}$$

where

$$a_1 = \frac{1}{K_i \frac{\partial T}{\partial \epsilon}} \left[ \frac{\partial T}{\partial n} + (K_r + K_i \tau_p) \frac{\partial T}{\partial \epsilon} \right]$$

$$a_2 = \frac{1}{K_i \frac{\partial T}{\partial \epsilon}} \left[ J_g + \tau_p \frac{\partial T}{\partial n} + K_r \frac{\partial T}{\partial \epsilon} \right]$$

$$a_3 = \frac{1}{K_i \frac{\partial T}{\partial \epsilon}} \left[ J_g (\tau_p + \tau_c) + K_p \tau_p \frac{\partial T}{\partial \epsilon} \right]$$

$$a_4 = \frac{1}{K_i \frac{\partial T}{\partial \epsilon}} J_g \tau_c \tau_p$$

The static stiffness is given by

$$K_o = K_i \frac{\partial T}{\partial \epsilon}$$

The stiffness should be high to minimize the effects of disturbance torques. Disturbance torques can result from gun bending mode vibrations, coupling, and bumps in the road. The static stiffness can be increased by increasing the system integral gain  $K_i$ . When  $K_i$  is increased, a low frequency dipole must be used in the forward path to lower the gain at high frequency as required for stability.

The analysis in Section 5.9 shows the effect of stiffness on tracking errors resulting from coupling.

### 5.9 COUPLING

The azimuth and elevation stabilization systems are coupled by both gyroscopic and hull roll rate effects. This section discusses this coupling and methods of minimizing their effect.

One cause of coupling between the azimuth and elevation axes is gyroscopic moments. According to the law for a gyroscope, if a mass is rotating about the x axis with angular momentum  $H_x$ , simultaneous rotation about the y axis will result in a torque about the z axis. Thus,

$$T_z = \omega_y \cdot H_x \quad (5-14)$$

where

$$H_x = J_x \omega_x \quad (5-15)$$

Applying this law to the elevation axis, with the gun pointing straight ahead, gives:

$$T_{gc} = \dot{\phi}_h \cdot J_{gy} \dot{\psi}_t + \dot{\psi}_t \cdot J_{gx} \dot{\phi}_h \quad (5-16)$$

where

$J_{gx}$  = inertia of the gun about the roll axis

$J_{gy}$  = inertia of the gun about the yaw axis

Thus, when turret rotation and roll occur simultaneously, the resulting torque  $T_{gc}$  is applied to the gun.

The gun moments of inertia about the yaw and roll axes are not known. A rough estimate of the inertia about the yaw axis was calculated by multiplying the gun mass by the square of the estimated distance from the gun's center of gravity to the center of turret rotation.

$$J_{gy} = 1.5^2 \times 100 = 225 \text{ kg-m-s}^2$$

The gun moment of inertia about the roll axis is assumed to be negligible.

The peak roll rate, from the HITPRO program, is 0.24 rad/s. The peak turret rate is assumed to be 1.571 rad/s.

The torque applied to the gun by the gyroscopic effect is thus

$$T_{gc} = \dot{\phi}_h \cdot J_{gy} \cdot \dot{\psi}_t = 0.24 \cdot 225 \times \frac{\pi}{2} = 85 \text{ kg-m}$$

The tracking error resulting from this torque is  $T_{gc}$  divided by the stiffness. The stiffness curve determined from the nonlinear computer simulation of the rate control system with electric gyro is shown in Figure 5-12. The stiffness at 0.6 Hz, the fundamental frequency of the bump course, is  $5.6 \times 10^5 \text{ kg-m/rad}$ . Thus

$$TE = \frac{85}{5.6 \times 10^5} = 0.00015 \text{ rad (0.15 mil)}$$

When the gun is not pointing straight ahead, the coupling equation becomes

$$T_{gc} = (J_{gx} \cdot J_{gy}) (\dot{\phi}_h \cos \psi_t + \dot{\theta}_h \sin \psi_t) \dot{\psi}_t \quad (5-17)$$

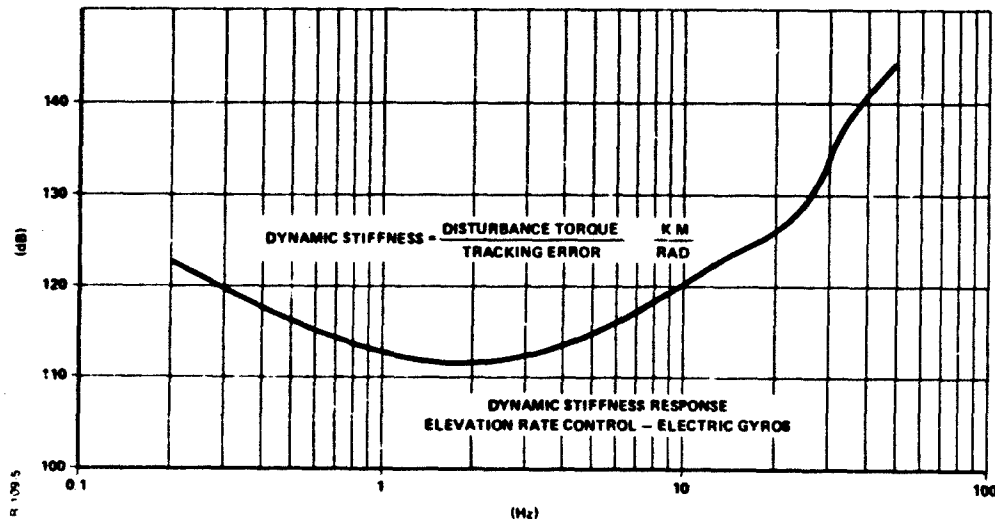


Figure 5-12. Dynamic Stiffness Curve

Similarly, for the azimuth axis, a torque  $T_{tc}$  is applied to the turret when the vehicle rolls while the gun is being rotated about the pitch axis, and when vehicle pitch and roll occur simultaneously.

$$T_{tc} = \dot{\phi}_h \cdot J_{tz} \dot{\theta}_h + \dot{\theta}_h \cdot J_{tx} \dot{\phi}_h + \dot{\phi}_h \cdot J_g \dot{\theta}_g$$

where

$J_{tx}$  = inertia of turret about the roll axis

$J_{tz}$  = inertia of turret about the pitch axis

The turret inertias are assumed to be equal about all three axes. The torque applied to the turret, using hull rates from HITPRO, is

$$\begin{aligned} T_{tc} &= 2 J_t \dot{\phi}_h \dot{\theta}_h + \dot{\phi}_h J_g \dot{\theta}_g = 2 \times 3140 \times 0.24 \times 0.18 + 0.24 \times 527 \times 1.05 \\ &= 408 \text{ kg-m} \end{aligned}$$

The dynamic stiffness curve was not run for azimuth. However, azimuth stiffness should be about six times higher than for elevation, due to the higher inertia of the turret. The higher stiffness in azimuth was indicated by the smaller tracking errors on the bump course with this axis. The estimated tracking error resulting from gyro effect is thus

$$TE = \frac{408}{6 \times 5.6 \times 10^5} = 0.00012 \text{ rad (0.12 mil)}$$

The estimated tracking errors are within the goal of 0.25 mil, but are large enough to be significant. The calculated values could be considerably in error if the estimated inertias are significantly in error.

Two methods can be used to reduce the tracking errors resulting from coupling. The first method is by increasing the stiffness. Stiffness is proportional to the integral gain  $K_I$ , as shown in Section 5.8. Therefore, increasing  $K_I$  will reduce the coupling error. It is also important to tune the system parameters and compensation so that the stiffness remains high over the frequency band of coupling.

The second method uses a hull roll rate sensor to cancel the coupling effect. The coupling error can probably be reduced to an acceptable level by increasing the stiffness. Therefore, the second method was not studied in detail.

A second coupling effect, for elevation, is the result of hull pitch angle, as seen by the gun, varying with turret angle. When the turret angle is zero, the torque acting on the gun is given by

$$T_{gl} = T_A - \left( D_{gh} + \frac{\partial T}{\partial n} \right) (\dot{\theta}_g - \dot{\theta}_h)$$

where  $D_{gh}$  is the trunnion friction and  $\partial T / \partial n$  is the slope of the actuator torque-speed curve. The significance of the  $\partial T / \partial n$  term will now be discussed. Figure 5-13 shows the gun with a piston actuator, and with the servovalve represented by orifice area  $A_v$ . If there were no oil in the cylinder, and no trunnion friction, hull pitch motion would not apply any torque to the gun. The gun inertia would then maintain the gun orientation fixed in space. However, when there is oil in the cylinder, any motion of the hull relative to the gun results in oil being pumped through orifice  $A_v$ . This flow causes a differential pressure to act on the piston, and thus applies a torque to the gun. The ratio of the torque to the relative velocity is the term  $\partial T / \partial n$ .

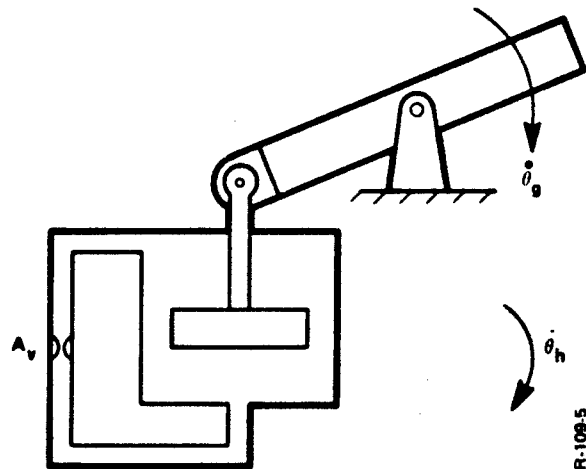


Figure 5-13. Schematic for Coupling of Hull Pitch Rate to Gun

The equation for friction coupling becomes considerably more complex when vehicle roll is considered.

A torque also results from hull acceleration. This torque is proportional to the gun mass unbalance  $L_g M_g$  and the distance between the hull centers of rotation and the trunnion.

$$T_{g2} = L_g M_g (R_p \ddot{\theta}_h \cos \psi_t + R_h \ddot{\phi}_h \sin \psi_t)$$

The mass unbalance of the gun is small. Therefore, this term will probably be negligible.

A second coupling for azimuth results from the turret mass unbalance with respect to the turret center of rotation  $L_t M_t$ . When the turret angle is zero, hull roll motion applies a torque to the turret, but hull pitch has no effect. When the turret angle is 90 deg, hull pitch motion applies a torque to the turret, and hull roll has no effect. The equation for the torque applied to the turret by these effects is

$$T_t = L_t M_t (\sin \theta_h \sin \psi_t + \sin \phi_h \cos \psi_t)$$

#### 5.10 SUMMARY OF ANALYTICAL SYSTEMS STUDY

Following is a summary of the results obtained and methods of synthesis accomplished by means of the analytical systems study. Some of these items proved to be invaluable in the subsequent computer analysis of the system.

- (1) A linear model of the stabilization system was derived which aided in the design of the control system and allowed the establishment of an initial set of control gains.
- (2) For terrain inputs, the responses of the rate and position command systems are identical.
- (3) Two methods can be used to reduce the tracking error resulting from hull motion. These are:
  - (a) Cancellation of hull rate using hull rate sensor
  - (b) Lag-lead compensation with high loop gain
- (4) A sensor error analysis restricted to gain and offset errors showed that only gun sensor offset errors are significant and cause the tracking error to increase with time.
- (5) A method was developed for eliminating the effects of sensor offset errors and integrator drift.



- (6) Sensor dynamics increase overshoot and reduce stability.
- (7) Lead-lag compensation will be required if the gun rate sensor phase lag is significant.
- (8) When a lead-lag configuration is used, it should be in the feedback path.
- (9) The integral control gain  $K_I$  should be high to maximize system stiffness and minimize the effects of disturbance torques.
- (10) When vehicle roll rates occur, significant coupling between the azimuth and elevation axes takes place in the form of gyroscopic moments and other effects.

## SECTION 6

### SIMULATION ANALYSIS RESULTS

#### 6.1 GENERAL

This section contains a detailed description of the analog and digital computer results obtained in this study for the gun stabilization system. The objectives of the computer study are outlined here. The organization of the relevant computer program as well as a summary of the types of inputs and outputs are discussed in Section 6.2. Section 6.3 compares the rate and position command systems. The results of a preliminary computer analysis evaluating the effects of nonlinear valve flow, vehicle hull dynamics, and hydraulic fluid compressibility are described in Section 6.4. The effects of these nonlinearities needed to be evaluated in order to judge whether or not they would be included in the detailed simulation models. A detailed evaluation of the complete stabilization system and the five candidate sensors are contained in Sections 6.5 and 6.6.

The following is a summary of the objectives considered in this systems study.

- (1) Evaluate the effects of the nonlinear valve flow equations versus linearized flow, and the effect of hull dynamics.
- (2) Derive stabilization system configurations for each of the sensors, and define gains and compensation networks required to meet the performance goals with each system.
- (3) Investigate the effects of sensor gain variation, sensor dead-band, and actuator friction on tracking error.
- (4) Perform detailed evaluation of the five prospective sensors for this application.
- (5) Determine the system time response for step and sinusoidal inputs.
- (6) Determine the tracking error for sinusoidal hull rates and for bump course using the HITPRO digital computer program.
- (7) Determine the gain and phase margins for each configuration.

The following sensors were analyzed in detail.

##### (1) Elevation Axis

Two electric gyros.

Two hydraulic rate sensors (GE).

Two laminar vortex rate sensors (Honeywell).

One integrating accelerometer (Bendix).

(2) Azimuth Axis

Two electric gyros

One pneumatic accelerometer (AiResearch)

Rate sensors can be used in either one (gun) or two (gun and hull) sensor configurations. Only the rate sensor configuration utilizing two sensors was analyzed in this study.

Accelerometers can be used in only the one sensor configuration. Since the accelerometer signal is integrated to obtain the rate signal, an accelerometer for sensing hull rate would have an open loop integrator which would result in excessive drift rates. On the other hand, the gun sensor integrator is within a closed control loop. Under those conditions, integrator drift will have less effect.

6.2 SIMULATION APPROACH

The organization of the computer programming approach utilized for the analog, hybrid, and digital computer programs is described in this section. The types of inputs, organization of the analog computer boards, analog computer switch logic utilized, and the hybrid computer inputs are also described. Detailed listings of all digital computer programs are contained in Appendix A. Analog and hybrid program wiring diagrams are presented in Appendix B.

Digital computer solutions were obtained for each system configuration as checks on the analog computer simulations. The following inputs were generally used for the digital computer simulations.

- (1) Command rate step inputs
- (2) Hull rate step inputs
- (3) Hull rate sinusoidal inputs

The analog program of the control system was initiated by means of a simplified linear model. The nonlinear flow equations, sensor dynamics, nonlinear friction, and deadband were added one at a time, so that their effects could be assessed individually.

In order to facilitate the selection of the various components for a particular computer simulation run, automated computer switches were used for the following purposes.

- (1) Selecting one of the five sensors.
- (2) Selecting either a two sensor configuration using a hull sensor and a gun sensor or a one sensor configuration using only a gun sensor.
- (3) Selecting either forward or feedback path compensation.

The following inputs were used for the hybrid computer study for determining time and frequency responses.

- (1) Step command rate input.
- (2) Sinusoidal rate command input.
- (3) Step hull rate input.
- (4) Sinusoidal hull rate input.
- (5) Simulated bump course from program HITPRO.

The step and sinusoidal command inputs used were sufficiently small to avoid saturation of any component of the simulation. A hull rate input in the azimuth axis as used here is representative of a pivot steer maneuver. The digital program HITPRO was used to provide realistic hull motion for the simulated bump course. To implement these rates in the simulation, the hull pitch motion was stored in the digital part of the hybrid computer from a magnetic tape and was then used for providing the elevation hull rate inputs. The same rate input was used for the azimuth axis because no comparable azimuth rates were available from the HITPRO program. This is a valid procedure since the power spectral densities of the hull pitch and yaw motions are approximately the same.

The sinusoidal hull rate input used was of approximately the same amplitude as the maximum hull rate observed for the bump course, and of approximately the same frequency as the bump course fundamental frequency.

### 6.3 COMPARISON STUDY OF RATE AND POSITION CONTROL FOR STABILIZATION

As discussed in Section 5.2.2, the transfer functions for hull rate inputs are the same for both the rate and position command systems. Also, the transfer functions for command inputs are the same if, for the position control system, the input compensation term  $s/(\tau_1 s + 1)$  is a perfect derivative.

In order to verify the above results, a computer study was made to compare responses using the rate and position command simulation models. Each of the rate and position command systems was modelled with equal gains and compensation networks and with two electric gyro sensors. A perfect derivative of the position command  $\theta_s$  was used in the position system. The frequency response curves of the two systems are shown in Figures 6-1 and 6-2. Figure 6-1 verifies that the two systems are, in fact, equivalent.

In an actual system, a perfect derivative cannot be generated; instead, an approximation to it must be generated by the input compensation network of the type  $s/(\tau_1 s + 1)$ . Figure 6-2 shows the system frequency response using a perfect derivative and with  $\tau_1 = 0.01$  s for the input compensation network.

As a result of this study, the performance evaluations of each of the five sensor configurations were conducted using the rate command system only.

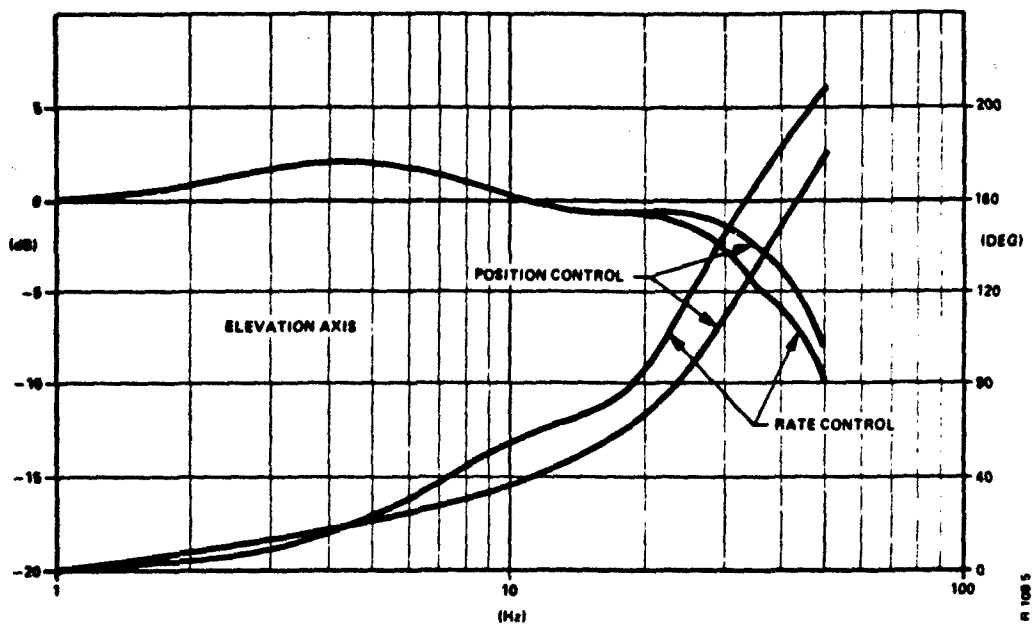


Figure 6-1. Frequency Response - Position Control

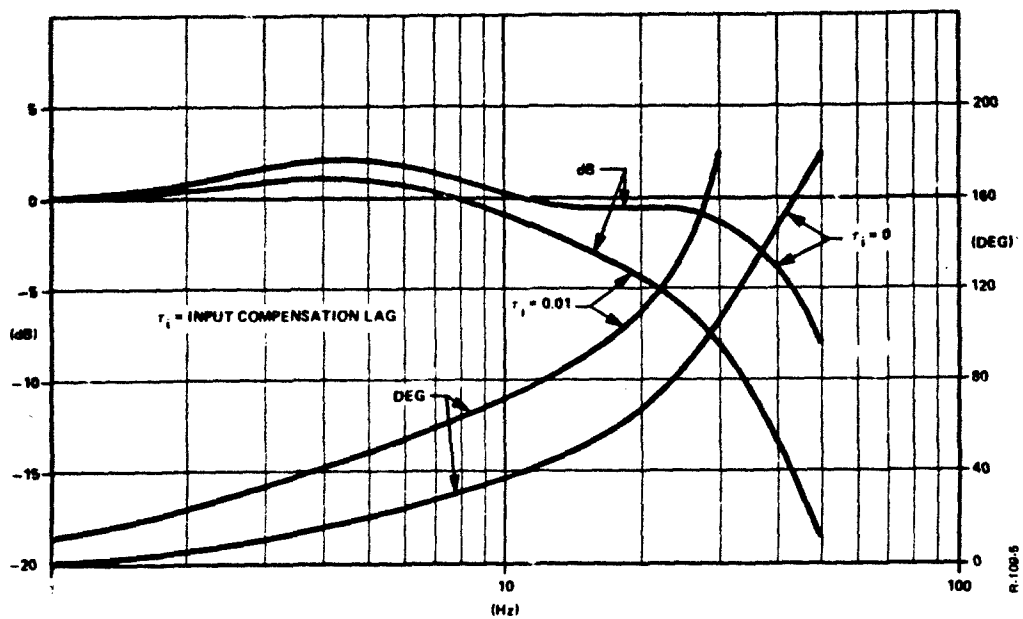


Figure 6-2. Frequency Response - Position Control and Rate Control

## 6.4 PRELIMINARY SIMULATION STUDIES

### 6.4.1 Effect of Nonlinear Valve Flow

It was found by means of the computer simulation that the rate command system (with ideal sensors) exhibited basically the same step response with the linear flow models as with the nonlinear. When the electric gyro dynamics were added, the nonlinear model became unstable, while the linear model remained stable. This indicates that the nonlinear flow model has less phase margin than the linear model. The nonlinear flow model was therefore used for all subsequent system studies.

### 6.4.2 Effect of Hull Dynamics

Hull dynamics modelled in this simulation represent the effect observed when the actuator applies torque to the gun. At that point an equal and opposite torque is applied to the hull and the resulting hull motion affects the gun. The simulation was used to determine system response to command and hull rate inputs, with and without hull dynamics. It was found that hull dynamics had a negligible effect on system response and on the tracking error. As a result, hull dynamics were not included in the model for subsequent studies.

### 6.4.3 Effects of Changes in Fluid Compressibility and Bulk Modulus

The volume under compression  $V$  modelled in this simulation consists of the high pressure fluid in the servovalve, actuator, and connecting lines. The numerical value for  $V$  used here was obtained from the bibliography. This value seems large and indicates that in the past, long connecting lines were used. To determine whether this significantly affected system performance, a step response simulation was performed with a 10 percent reduction in the value of  $V$ . A negligible change in the step response of the system was observed.

The effect of a drastic change in the bulk modulus was also determined. This effect represents the process of air entering the hydraulic fluid. It was found that reducing the fluid bulk modulus  $B$  by 50 percent, has a negligible effect on the system step response and stability. These results indicate that the amount of pressure feedback used here was great enough to compensate for the large volume under compression.

Compressibility may have different effects when compliance and gun bending modes are added to the model. These effects should therefore be evaluated again in future studies when bending modes and compliance have been modelled.

## 6.5 EVALUATION OF SENSORS IN ELEVATION AXIS

Sections 6.5 and 6.6 contain the evaluation of each of the five sensors studied. Time and frequency response plots are presented to

illustrate the results. The results for the sensors in Section 6.5.1 through 6.5.4 are based on elevation (gun) axis parameters while the results for the pneumatic accelerometer in Section 6.6 are based on azimuth axis parameters. All simulation results presented here are based on the rate command control concept as pointed out in Section 6.3.

#### 6.5.1 Evaluation of Electric Rate Gyros

A block diagram of the rate command system with two electric rate gyros is presented in Figure 5-6. A hull rate sensor is used for cancellation of hull rate coupling, and a gun rate sensor is used for closing the control loop. Lead-lag compensation is used to compensate for sensor phase lag, and thus to improve stability. Compensation is in the feedback path. This is the best location for the compensation in order to reduce the overshoot to command step inputs. (See Section 5.7.)

Using the hybrid computer simulation, the gains were adjusted to obtain the desired frequency response, stability, and tracking error. The resulting gain values and compensation are:

$$K_i = 50 \text{ 1/s}$$

$$K_r = 7.5$$

$$K_p = 2.5 \times 10^{-4} \text{ rad/(s-kg-cm}^2\text{)}$$

$$K_h = 0.15$$

$$\text{Feedback compensation } \frac{1 + 0.008 \text{ s}}{1 + 0.0016 \text{ s}}$$

The response to a command step input  $\dot{\theta}_g = \text{rad/s}$  was obtained, and is shown in Figure 6-3. The gun rate reaches the commanded speed, 0.05 rad/s, in about 0.04 s, overshoots about 15 percent, then settles to the commanded speed with no oscillations. The absence of oscillations indicates good damping. In the absence of a specific performance requirement for step response, the speed of response as well as damping appear satisfactory.

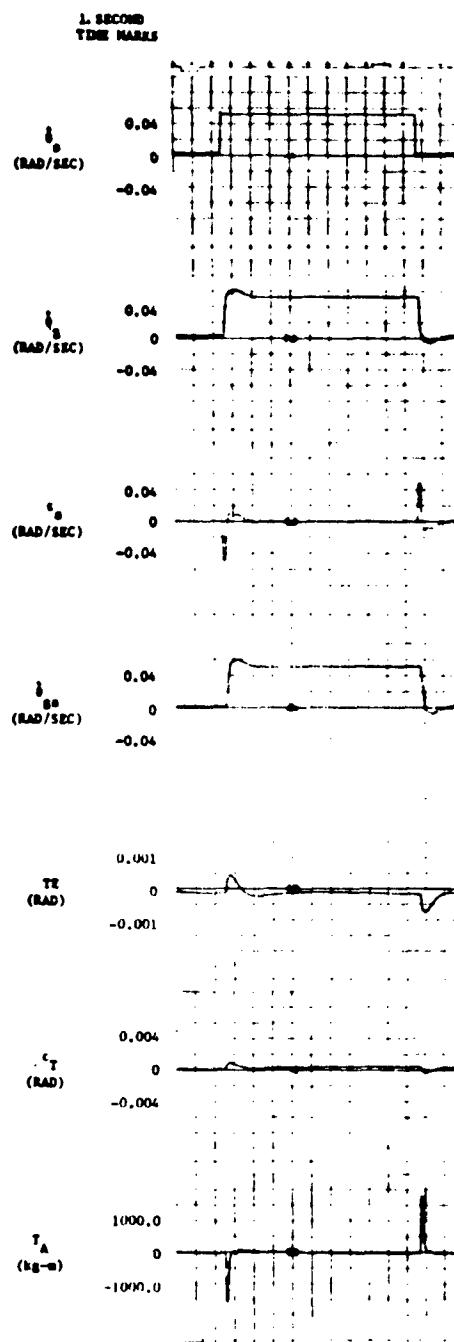


Figure 6-3. Response to a Command Step Input - Elevation Rate Control with Electric Gyros



The error rate signal  $\epsilon_s$  given by  $\epsilon_s = \dot{\theta}_s - \dot{\theta}_{gs}$  is the input for the open loop response. The open loop response is given by:

$$\frac{\dot{\theta}_{gs}}{\dot{\theta}_s - \dot{\theta}_{gs}}$$

The variable  $\dot{\theta}_{gs}$  (sensed gun rate) is the output signal for the open loop response. The variables  $\dot{\theta}_s$  and  $\dot{\theta}_{gs}$  are used mainly for determining frequency response, gain margin, and phase margin. These variables are also used as an aid in tuning the system.

The tracking error TE and the sensed tracking error  $\epsilon_T$  are defined as:

$$TE = \theta_s - \theta_g$$

$$\epsilon_T = \int \epsilon_s dt = \theta_s - \theta_{gs}$$

The variable TE originates at an open loop integration and therefore tends to drift. The variable  $\epsilon_T$  is obtained from within the closed loop and generally does not drift. These two signals are nearly identical except when sensor deadband or a high frequency input are used. The actuator torque  $T_A$  is also shown.

The response to a sinusoidal hull pitch rate is shown in Figure 6-4. The pitch rate input is:

$$\dot{\theta}_h = 0.18 \sin (2\pi \times 0.6 t) \text{ rad/s}$$

This input is approximately equal to the most severe pitch rate experienced on the simulated bump course. As shown in Figure 6-4, the resulting tracking error has a peak-to-peak amplitude of 0.18 mil and thus meets the pointing performance specification of the system.

The closed loop frequency response curves  $\dot{\theta}_g/\dot{\theta}_s$  are shown in Figure 6-5. The open loop frequency response curves were used for determining the gain and phase margins. An input signal amplitude of

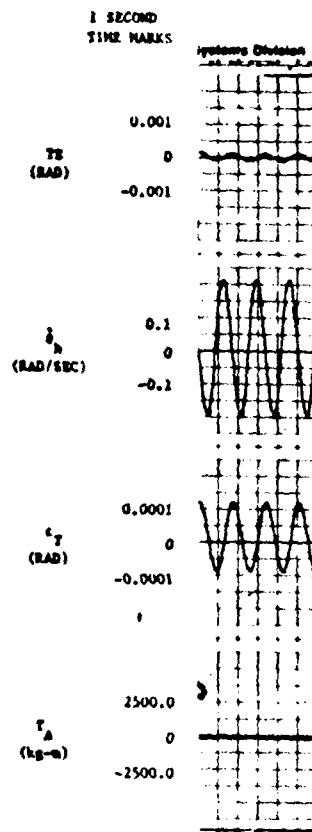


Figure 6-4. Response to Sinusoidal Pitch Rate - Elevation Rate Control with Electric Gyros

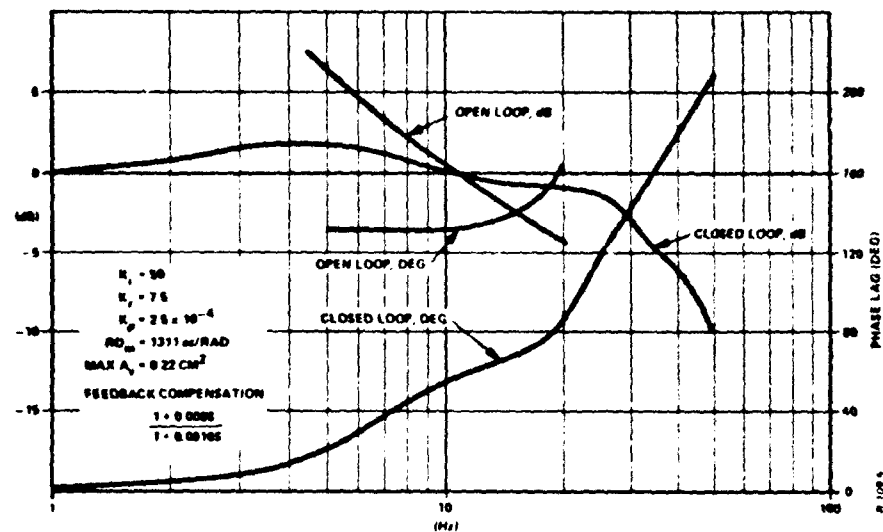


Figure 6-5. Closed Loop Frequency Response Curves - Elevation Rate Control with Electric Gyros

0.03 rad/s peak to peak was used for all frequency response tests. The frequency response results are as follows.

System bandpass: 28 Hz  
Open loop phase lag  
at gain crossover: 108 deg  
Phase margin: 180 deg - 108 deg = 72 deg  
Gain margin: 0.5

The system gain margin was determined by increasing the gain until instability resulted. It was found that the system was stable with the gain doubled, and unstable with higher gain. This is equivalent to a gain margin of 0.5.

The frequency response results meet or exceed the performance criteria specified in Section 3.3.

Time response plots using inputs consisting of vehicle rates from the simulated bump course of the HITPRO program are given in Figures 6-6 through 6-8. Graphs summarizing these results are presented in Figures 6-9 and 6-10. The maximum tracking error for nominal conditions ( $K_h = 0.15$ ) was 0.18 mil peak to peak.

Figure 6-9 shows that a hull sensor gain error of 50 percent increased the tracking error to 0.4 mil.

Figure 6-10 shows that combined coulomb and stiction friction of 50 kg-m each, increased the tracking error to 0.4 mil.

The 0.18 mil tracking error observed with nominal conditions is well within the 0.5 mil criteria listed in Section 3.2. Note however that the combined effects of sensor gain error, sensor deadband, friction, noise, and cross-coupling of axes will increase the tracking error. The tracking error can then be reduced by increasing the system low frequency gain. This can be accomplished by adding lag-lead compensation in the forward path, of the form:

$$\frac{\tau_2}{\tau_1} \frac{1 + \tau_1 s}{1 + \tau_2 s}$$

A graph demonstrating dynamic stiffness is presented in Figure 6-11. These results were obtained by applying a sinusoidal disturbance torque, and recording tracking error as the output signal. Stiffness is the ratio of the torque amplitude to the tracking error. The minimum stiffness was found to be 111 dB. The dynamic stiffness should be high to minimize the effects of coupling and bending mode vibration.

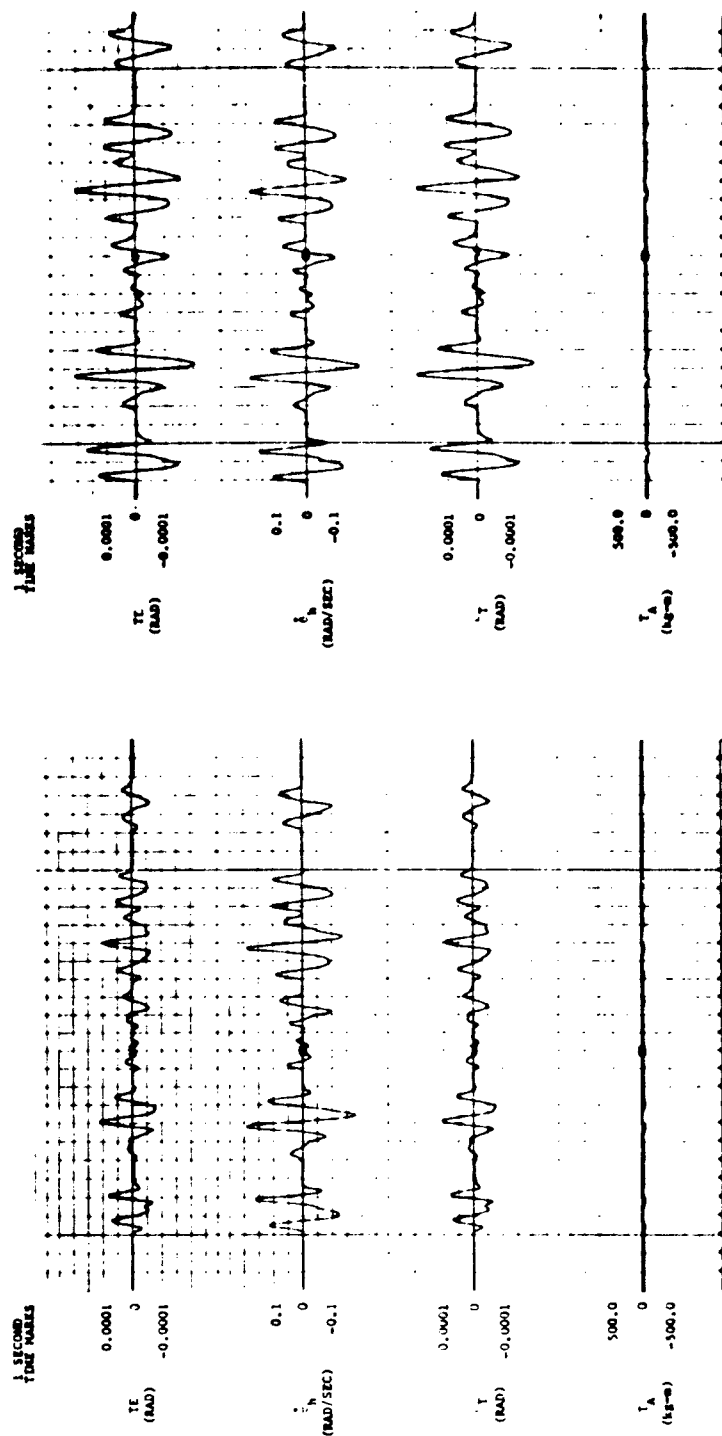


Figure 6-6. Response to HITPRO Simulated Bump Course - Elevation Rate Control with Electric Gyros

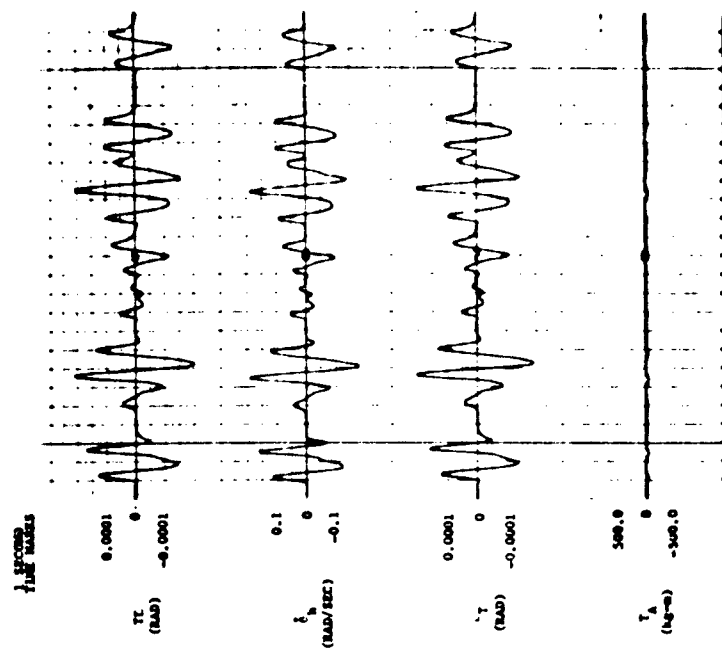


Figure 6-7. Response to HITPRO Bump Course with Hull Sensor Gain Error - Elevation Rate Control with Electric Gyros

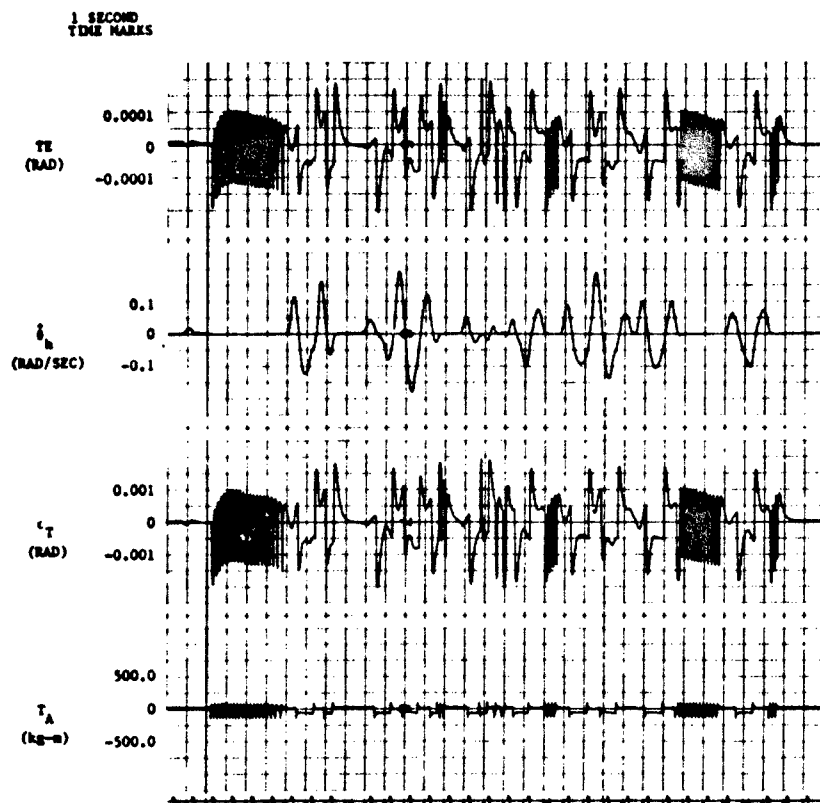


Figure 6-8. Response to HITPRO Bump Course with Coulomb and Stiction Friction - Elevation Rate Control with Electric Gyros

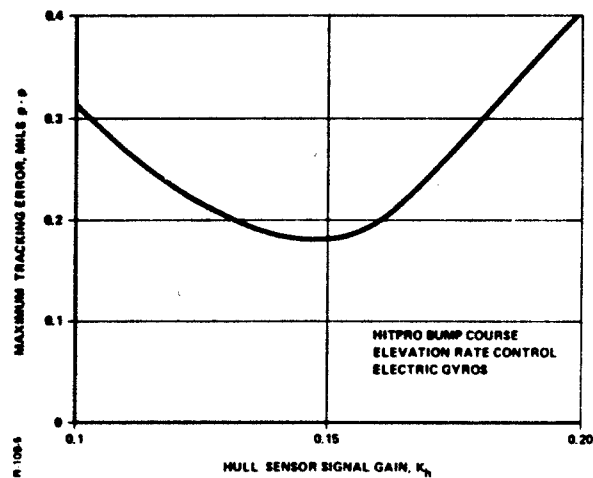


Figure 6-9. Graph of Tracking Error Versus Hull Sensor Gain Error - Elevation Rate Control with Electric Gyros

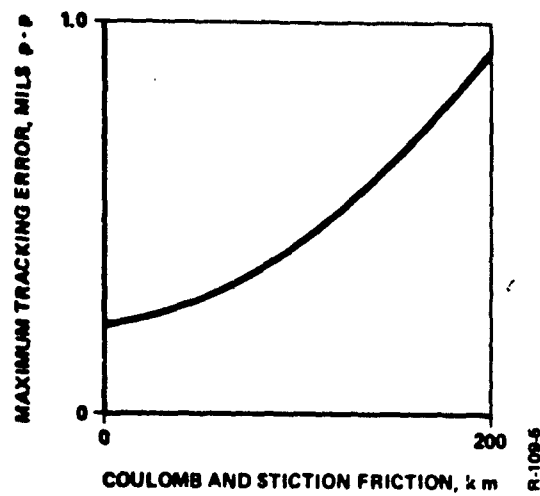


Figure 6-10. Graph of Tracking Error Versus Friction - Elevation Rate Control with Electric Gyros

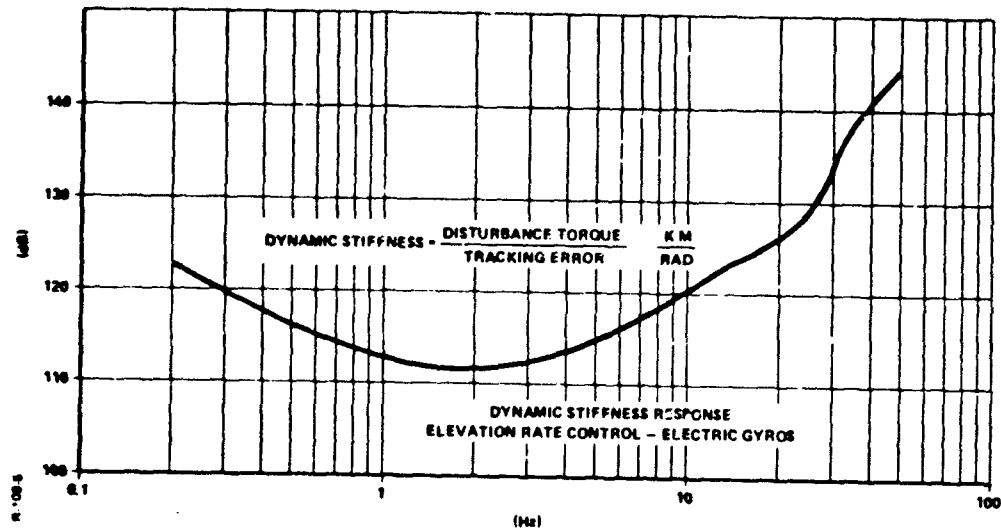


Figure 6-11. Dynamic Stiffness - Elevation Rate Control with Electric Gyros

### 6.5.2 Evaluation of Hydraulic Rate Sensor

Figure 5-6 is a block diagram of the stabilization system with the hydraulic rate sensor. The basic configuration is the same as with two electric gyros. Control gains and feedback compensation were identical to those used with the electric gyros.

Using the hybrid computer, the response of the system to a step rate command  $\theta_g$  was obtained. As shown in Figure 6-12, this step response exhibited a 20 percent overshoot. The response time is about the same as for the system with two rate gyros.

The response to sinusoidal hull motion of an amplitude equivalent to the bump course is shown in Figure 6-13. The tracking error observed was 0.12 mil peak to peak.

The frequency response results shown in Figure 6-14 can be summarized as follows.

Bandpass: 13 to 37 Hz

Gain margin: 0.53

Phase margin: 50 deg

Note that the amplitude ratio is down 3 dB at 13 Hz and drops to -5 dB at 21 Hz. Modification of the feedback compensation network will allow improvement of the system bandpass.

The response to the HITPRO simulated bump course is shown in Figures 6-15 through 6-17. Figure 6-18 shows the maximum observed tracking error versus hull sensor gain error. A gain error of 33 percent increased the maximum tracking error to 0.38 mil. The maximum tracking error versus sensor threshold or deadband is shown in Figure 6-19. A deadband of 6 mils increases the tracking error to 2 mils. The effects of combined deadband and friction effects are shown in Figure 6-20. A small amount of friction has the effect of reducing the tracking error. Larger amounts of friction increase the tracking error. The combination of deadband and friction increases the tendency for limit cycling.

### 6.5.3 Evaluation of Integrating Accelerometer

The block diagram of the stabilization system with an integrating accelerometer is presented in Figure 5-7. Only a single sensor, mounted on the gun, is used. The processed sensor signal is proportional to acceleration at frequencies below 0.1 Hz and to rate above 0.1 Hz. In the simulation model, the compensation network was designed to integrate the sensor signal below 0.1 Hz and thus to provide a rate signal at all frequencies.

Since there is no hull sensor, the system gain must be increased to reduce the tracking error. To achieve this, compensation was used in the forward path. The type of compensation increases the

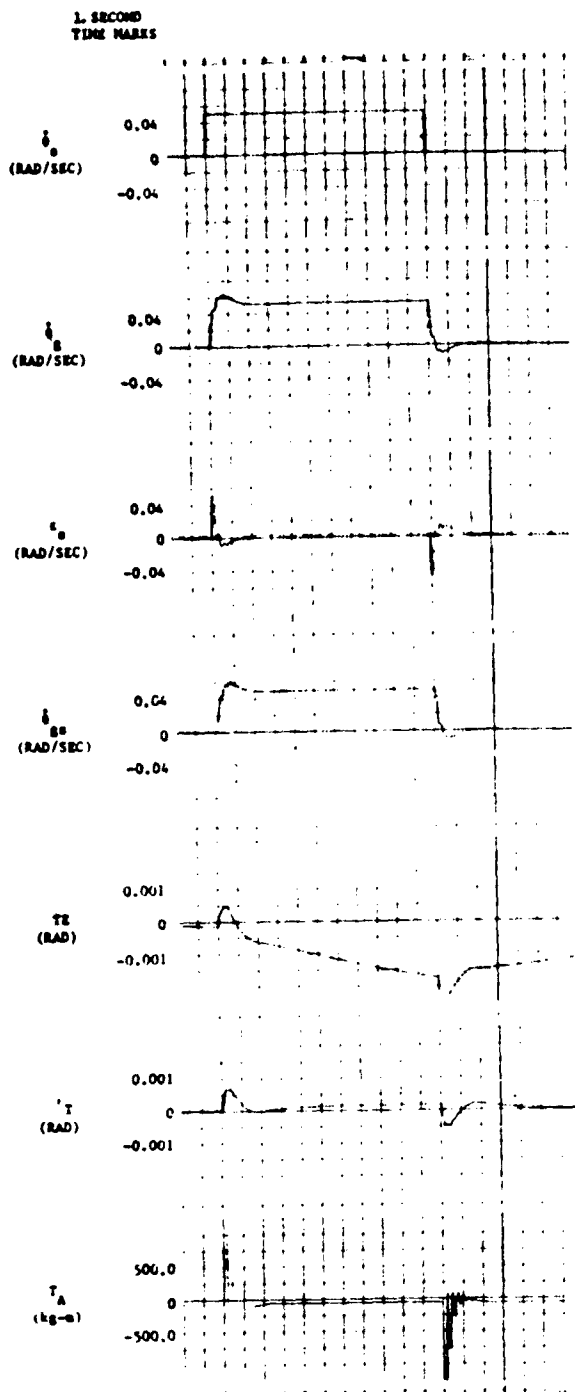


Figure 6-12. Response to a Step Rate Command - Hydraulic Rate Sensor.

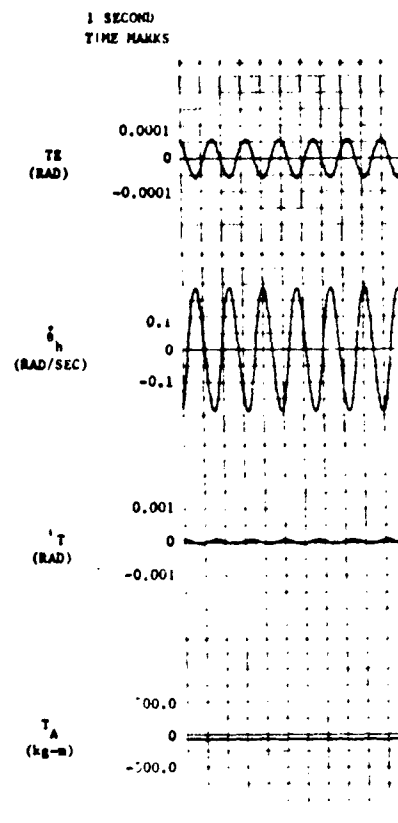


Figure 6-13. Response to Sinusoidal Hull Motion - Hydraulic Rate Sensor



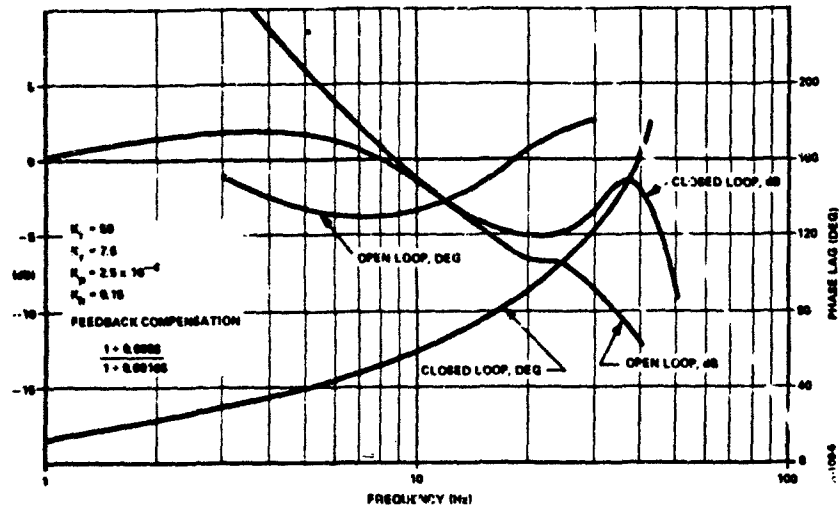


Figure 6-14. Frequency Response - Hydraulic Rate Sensor

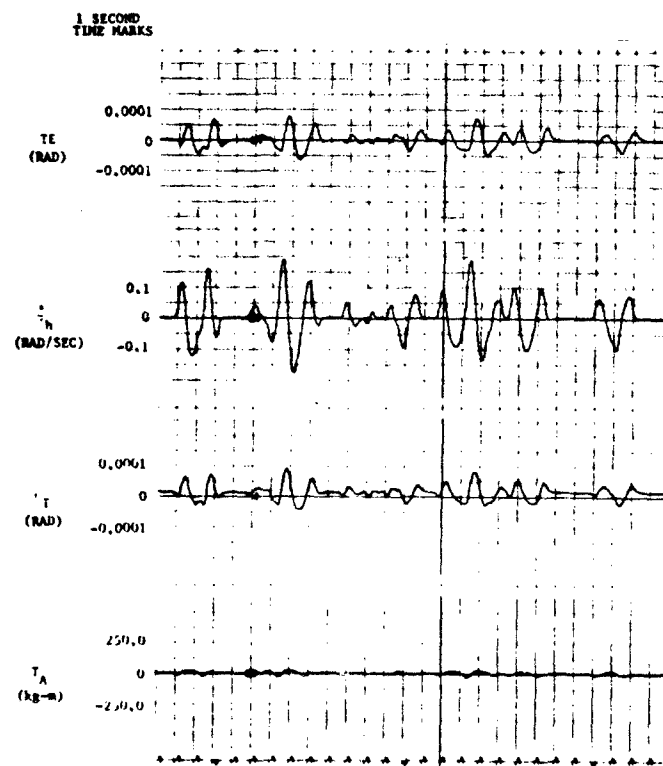


Figure 6-15. Response to HITPRO Simulated Bump Course - Hydraulic Rate Sensor

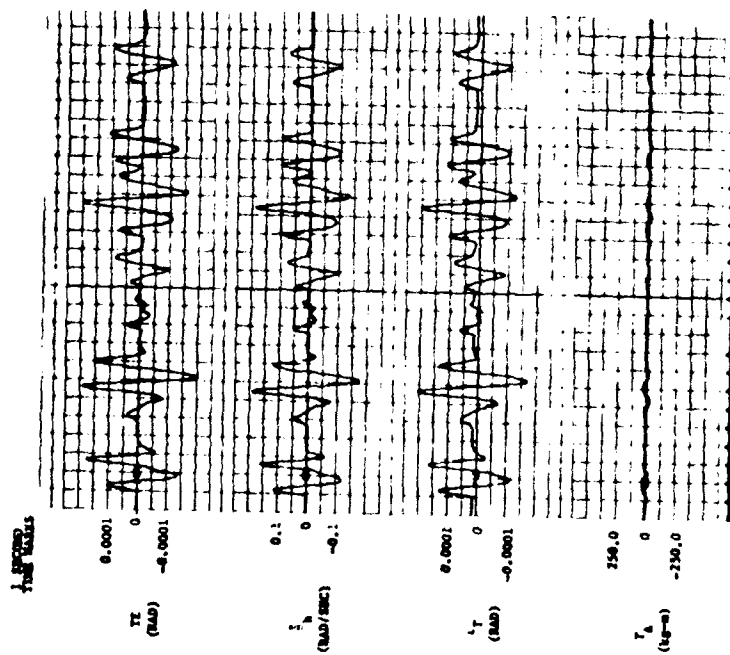


Figure 6-16. Response to HITPRO Bump Course with Hull Sensor Gain Error - Hydraulic Rate Sensor

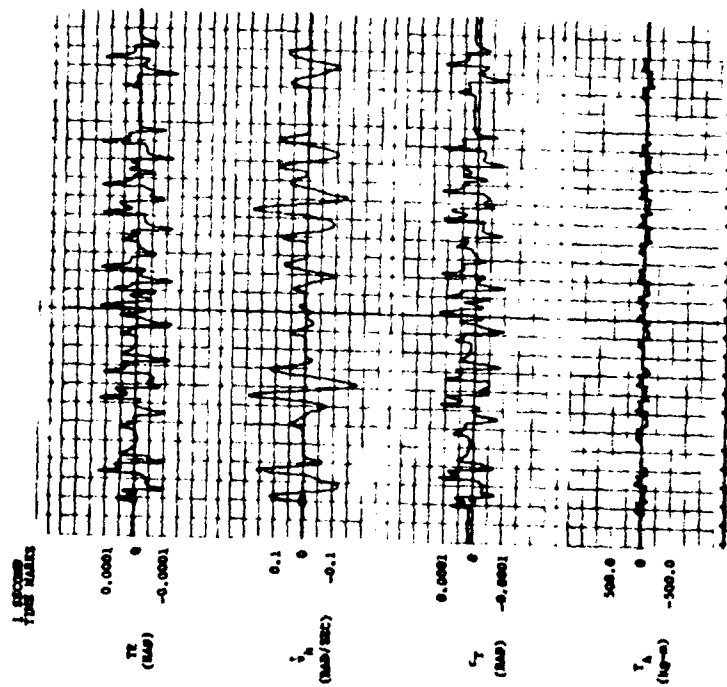


Figure 6-17. Response to HITPRO Bump Course with Coulomb and Stiction Friction - Hydraulic Rate Sensor

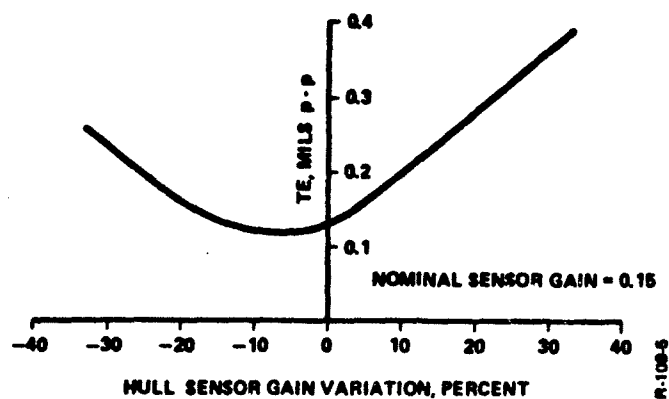


Figure 6-18. Graph of Tracking Error Versus Hull Sensor Gain Error - Hydraulic Rate Sensor

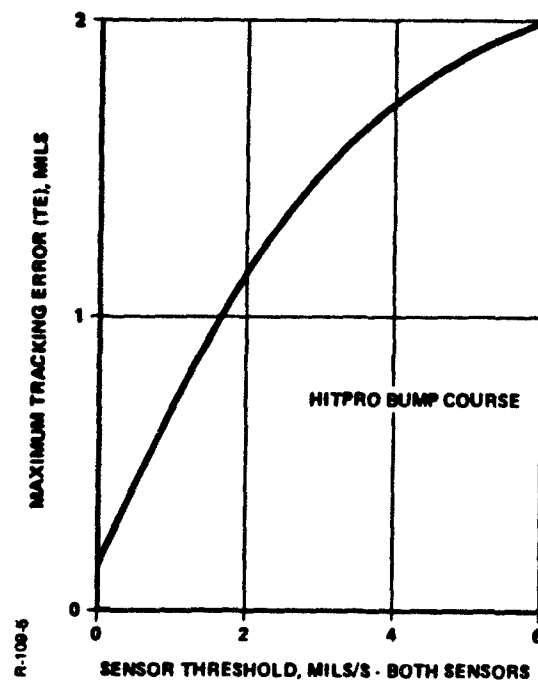


Figure 6-19. Tracking Error Versus Sensor Deadband - Hydraulic Rate Sensor

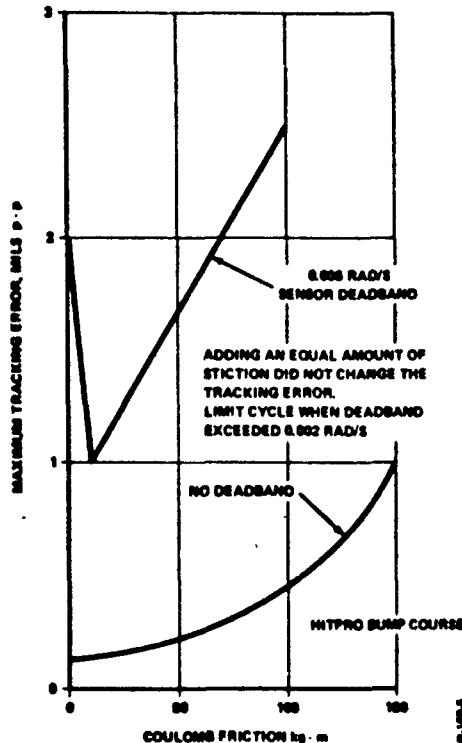


Figure 6-20. Effects of Combined Hull Sensor Deadband and Friction - Hydraulic Rate Sensor

gain at low frequencies and decreases the gain at high frequencies as is required for stability. The forward compensation and system gains established using the hybrid computer simulation are as follows:

Forward compensation:  $10 \frac{1 + 0.055 s}{1 + 0.55 s}$

Control gains:  $K_I = 250$   
 $K_T = 7.5$   
 $K_P = 2.5 \times 10^{-4}$

The step response of this sensor configuration to a command rate input is shown in Figure 6-21. The response has a 60 percent overshoot. This overshoot can be reduced significantly using a lead-lag feedback compensation.

The response to a sinusoidal hull input, equivalent in amplitude to the most severe motion for the bump course, is shown in Figure 6-22. The tracking error observed was 0.11 mil peak to peak. The response to the actual HITPRO bump course is shown in Figure 6-23.

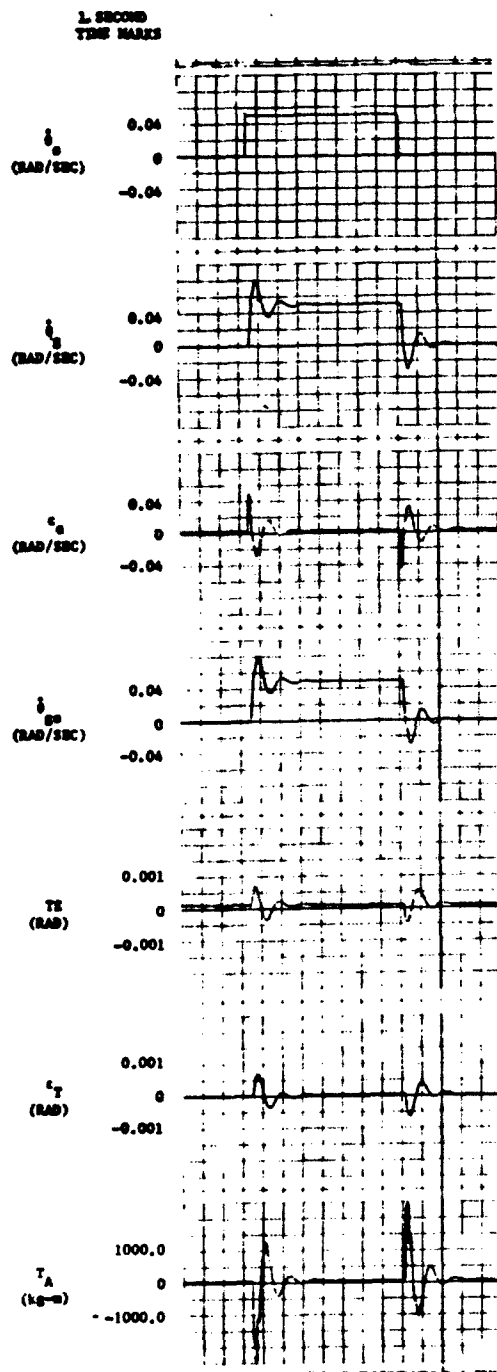


Figure 6-21. Response to a Step Rate Command - Integrating Accelerometer

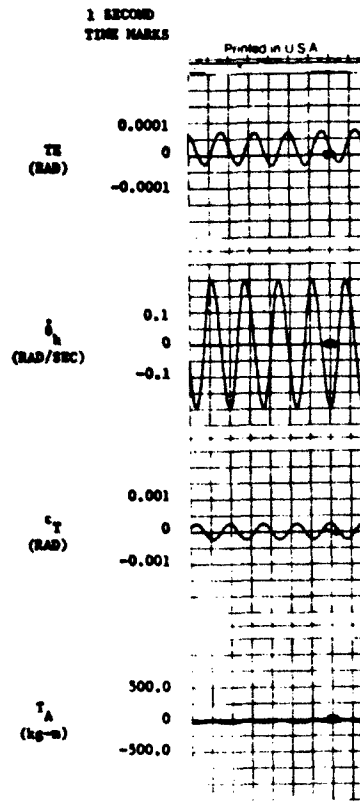


Figure 6-22. Response to Sinusoidal Hull Motion - Integrating Accelerometer

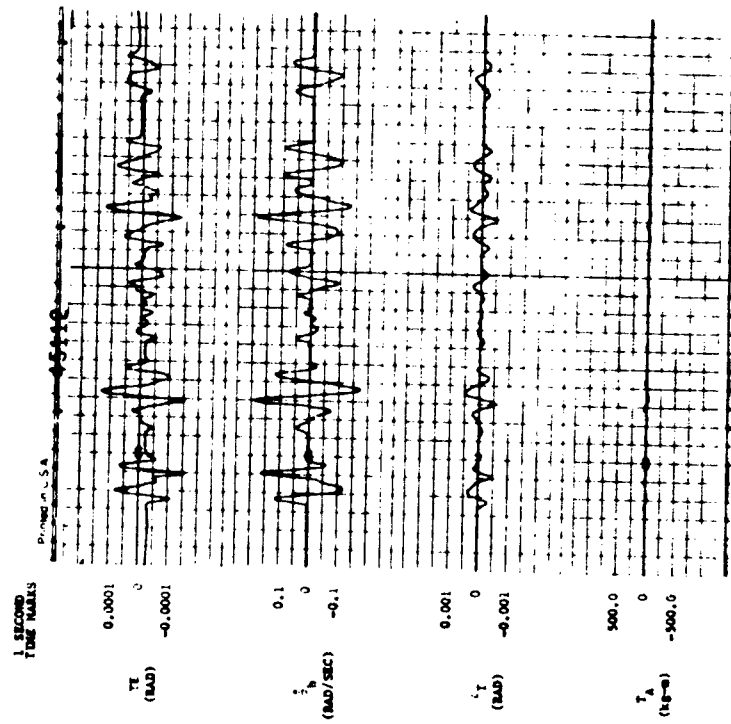


Figure 6-23. Response to HITPRO Bump Course - Integrating Accelerometer

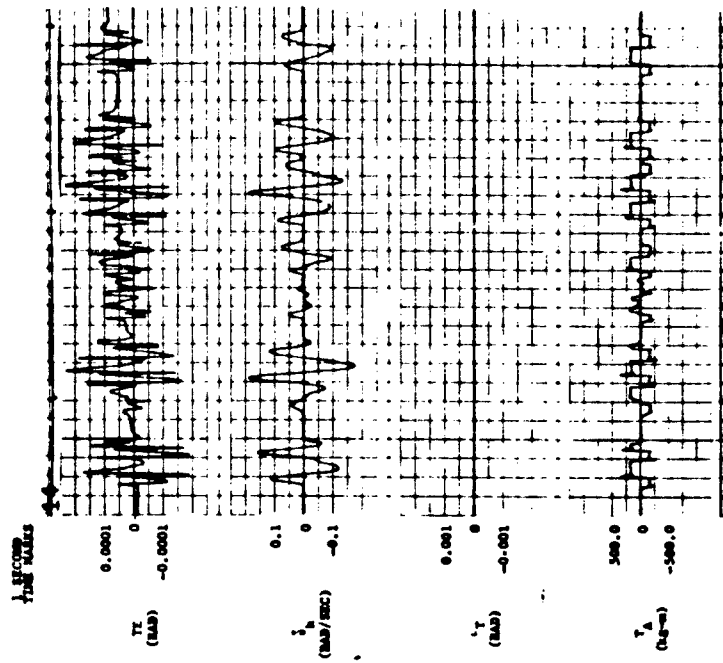


Figure 6-24. Response to HITPRO Bump Course with Coulomb and Stiction Friction - Elevation Rate Control with Integrating Accelerometer

The maximum tracking error observed is 0.28 mil peak to peak. The tracking error for the HITPRO input could be reduced by modifying the compensation.

Coulomb and stiction friction values of 100 kg-m each increased the tracking error to 0.38 mil, as shown in Figure 6-24. Figure 6-25 shows the maximum tracking error for various values of friction.

The frequency response curves are shown in Figure 6-26. The frequency response results can be summarized as follows.

Bandpass: 16 Hz  
Gain margin: 0.53  
Phase margin: 45 deg

#### 6.5.4 Evaluation of Laminar Vortex Sensor

A block diagram for the rate command system with two laminar vortex sensors is shown in Figure 5-6. The sensor models consist of a single lag and a time delay. The time delay was simulated in the digital part of the hybrid computer. To achieve this for the gun rate sensor, the gun rate signal from the analog computer was fed into the digital computer, delayed in time, and then fed back to the analog computer. For the hull sensor, the hull rate from the HITPRO program was delayed in time in the sensor path, but used without time delay in the remaining portion of the simulation.

The step response to a rate command input is shown in Figure 6-27. A 30 percent overshoot is exhibited.

The response to the simulated bump course using the HITPRO program is shown in Figures 6-28 and 6-29 for hull sensor gains of 0.15 and 0.20, respectively. A plot of tracking error versus hull sensor gain  $K_h$  is shown in Figure 6-30. From this graph, a gain of 0.15 was selected as the nominal value with the minimum tracking error. Increasing  $K_h$  to 0.20 increased the tracking error to 0.6 mil. Decreasing  $K_h$  to 0.10 increased the tracking error to 0.28 mil.

The sinusoidal hull input response was not obtained for this sensor.

The frequency response curves are shown in Figure 6-31. The frequency response results are summarized as follows.

Bandpass: 25 Hz  
Gain margin: 0.50  
Phase margin: 50 deg

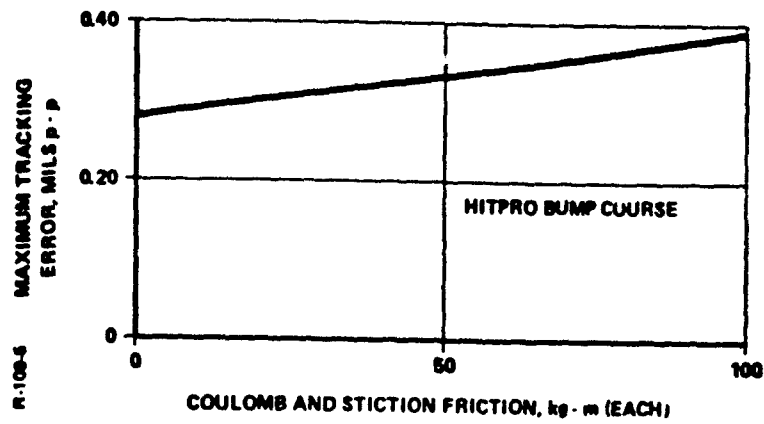


Figure 6-25. Maximum Tracking Error Versus Friction - Integrating Accelerometer

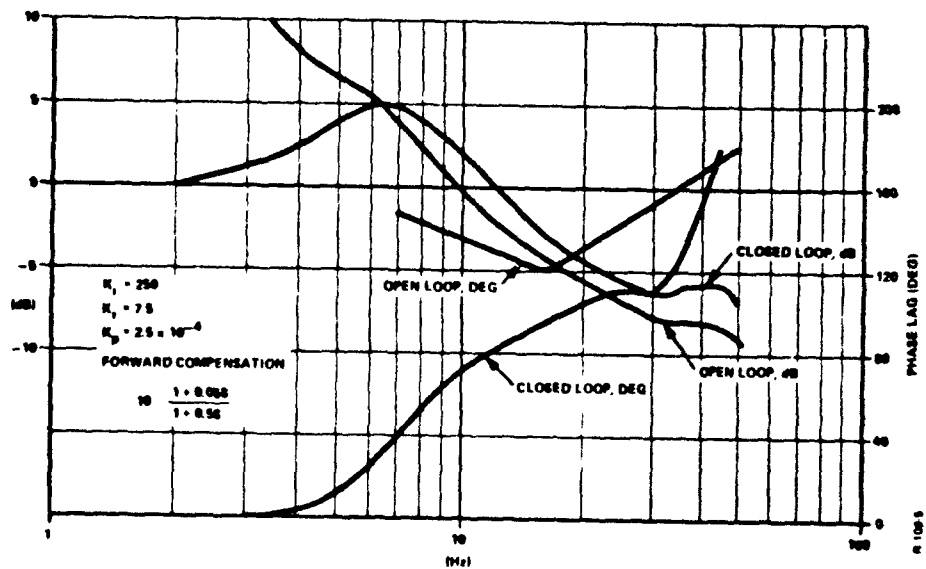


Figure 6-26. Frequency Response - Integrating Accelerometer



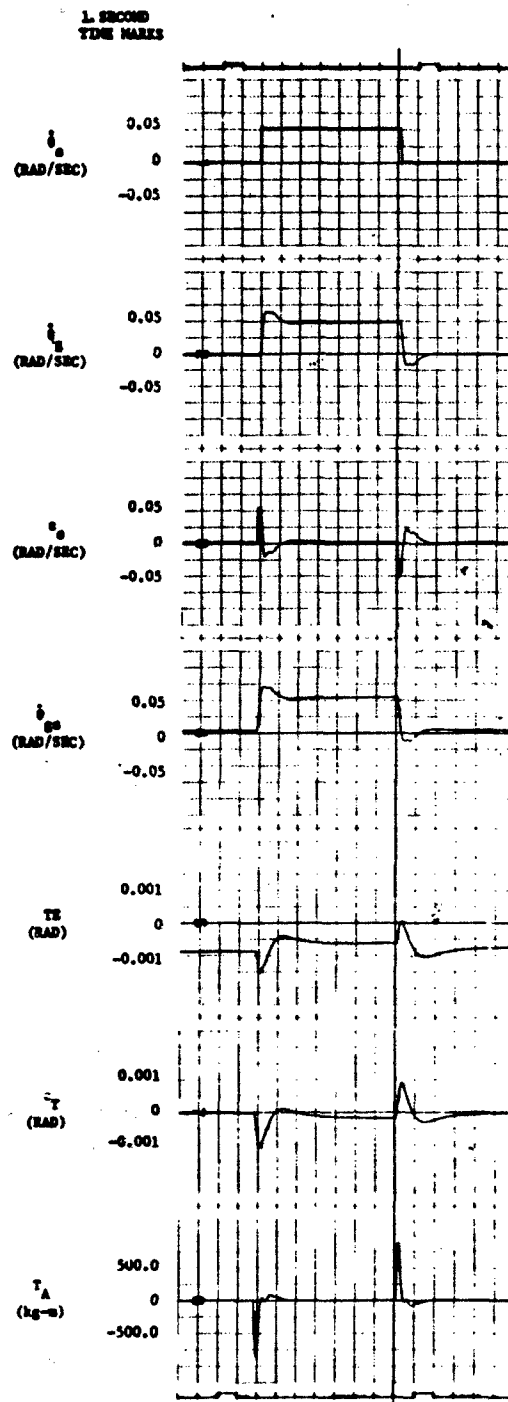


Figure 6-27. Response to a Step Rate Command - Laminar Vortex Sensor

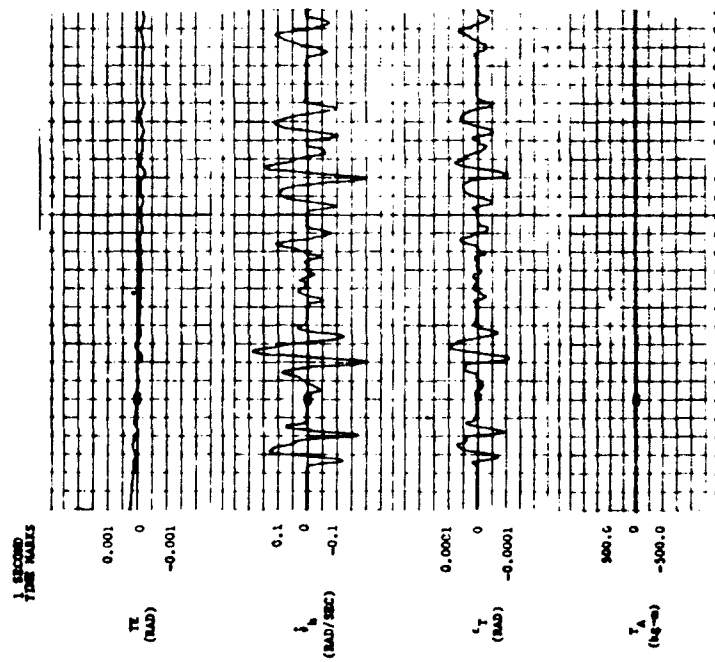


Figure 6-28. Response to HITPRO Bump Course - Laminar Vortex Sensor

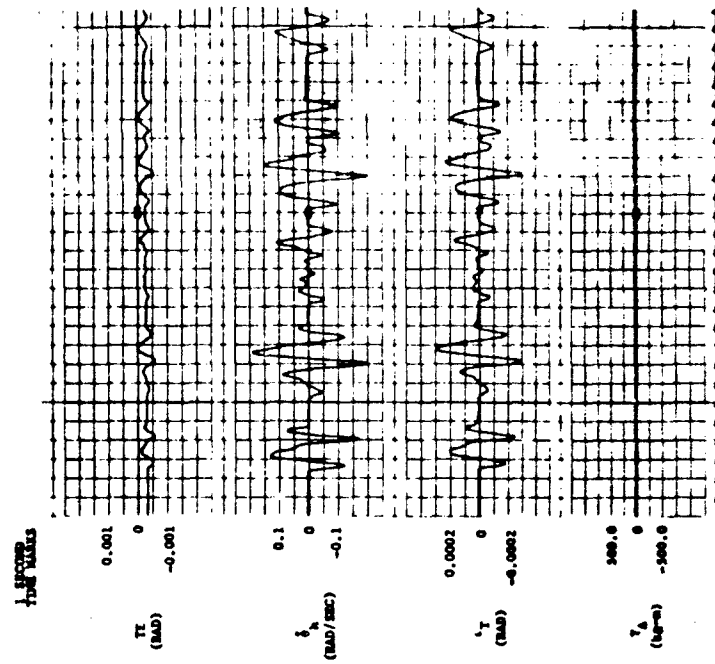


Figure 6-29. Response to HITPRO Bump Course with Hull Sensor Gain Error - Laminar Vortex Sensor

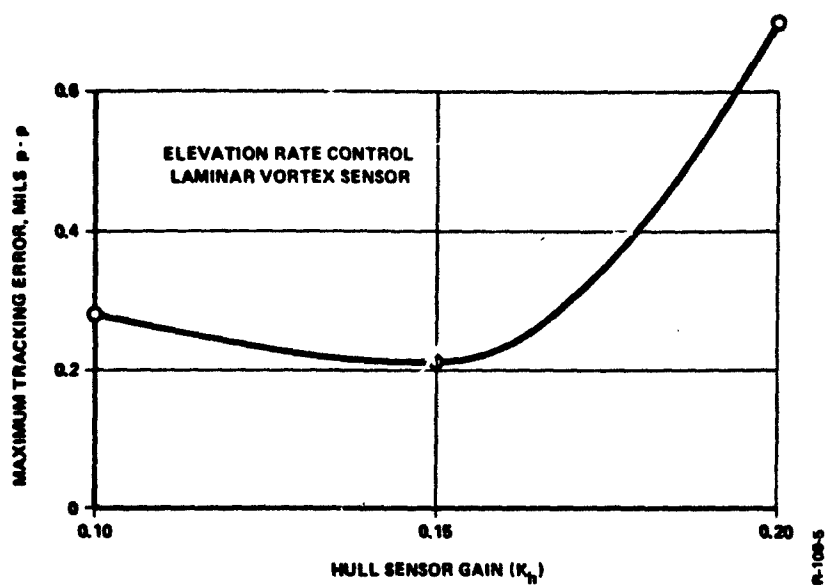


Figure 6-30. Tracking Error Versus Hull Sensor Gain - Laminar Vortex Sensor

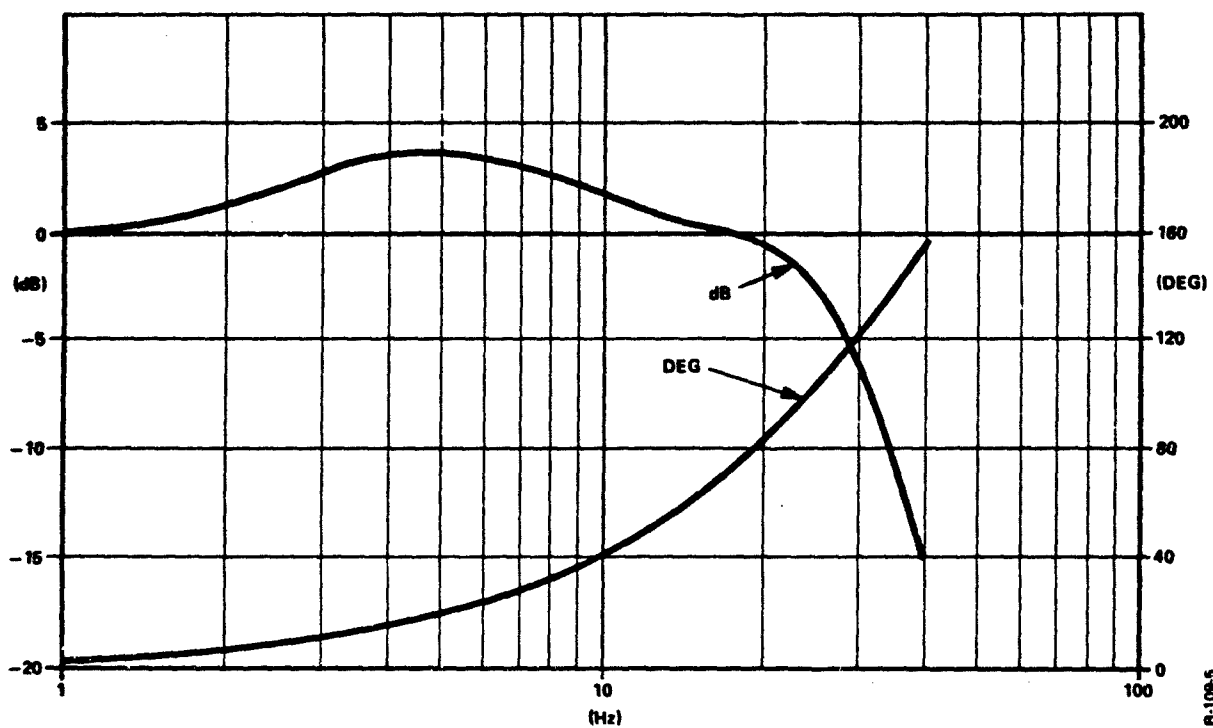


Figure 6-31. Frequency Response - Laminar Vortex Sensor

## 6.6 EVALUATION OF SENSORS IN AZIMUTH AXIS

### 6.6.1 Evaluation of Electric Rate Gyros

The block diagram of the azimuth stabilization system with two electric rate gyros, shown in Figure 5-6, is identical to the block diagram for the elevation axis. Only some of the parameters were changed to allow for hardware differences between axes. (See Tables 2-1 and 2-2.)

Using the hybrid computer simulation, the step response of the system to a rate command input was obtained and is shown in Figure 6-32. The observed overshoot is 36 percent.

The response to sinusoidal hull motion equivalent in amplitude to the bump course is shown in Figure 6-33. The observed tracking error was 0.03 mil peak-to-peak.

The response of the system using the simulated HITPRO bump course is shown in Figure 6-34. The maximum tracking error is also 0.03 mil peak to peak. The response of the system to the bump course with a 20 percent hull sensor gain error was also investigated and the results are as shown in Figure 6-35. This amount of gain error increased the tracking error to about 0.07 mil.

The effects of hull sensor gain variations for both the bump course and sinusoidal hull motions are summarized in Figure 6-36. It is evident that the observed tracking error is approximately the same for either input.

Additional simulation studies were conducted to determine the response to the bump course with 2 and 6 mil/s sensor deadband. These results are shown in Figures 6-37 and 6-38, respectively.

The effects of deadband and friction on the tracking error are summarized in Figure 6-39. Coulomb friction of 100 kg-m increased the tracking error to 0.5 mil. The addition of an equal amount of stiction friction did not change the tracking error. Sensor deadband of 2 mils/s increased the tracking error to 0.07 mil. For larger deadband, a limit cycle occurred, and the tracking error was greatly increased.

The frequency response curves are shown in Figure 6-40. The frequency response results are numerated as follows.

Bandpass: 21 Hz  
Gain margin: 0.50  
Phase margin: 55 deg

The response of the system to a pivot steer maneuver is shown in Figure 6-41. In a pivot steer maneuver, the tank is turning at its maximum rate, i.e., 180 deg in 8 s. In practice, it is desired that the gun remain pointing at the target during this maneuver. As shown in Figure 6-41, there is a momentary tracking error of 0.09 mil. The tracking error then decays to zero after about 0.5 s.

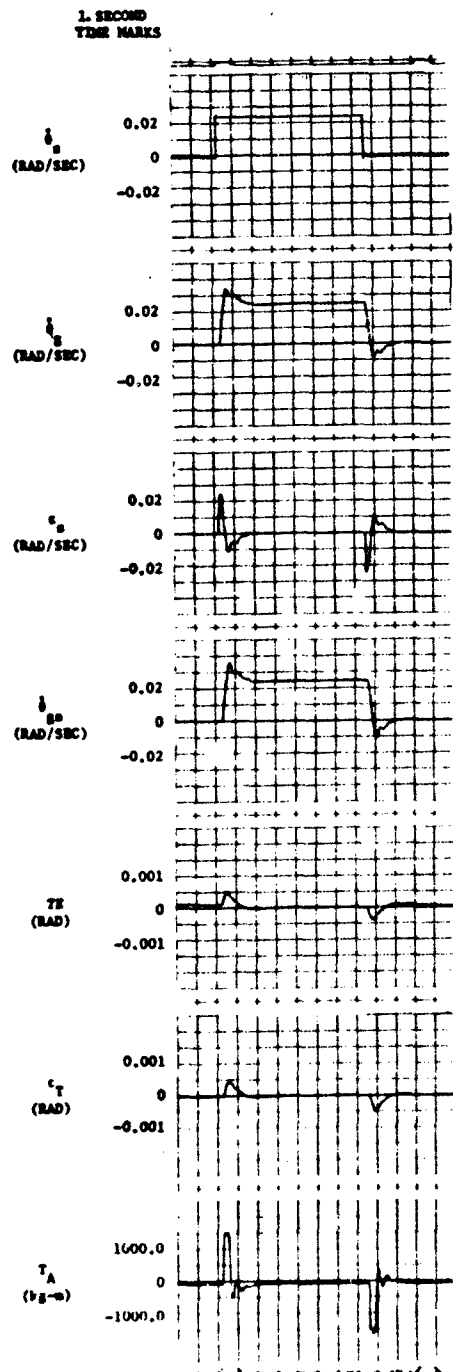


Figure 6-32. Step Response - Azimuth Rate Control with Electric Gyros

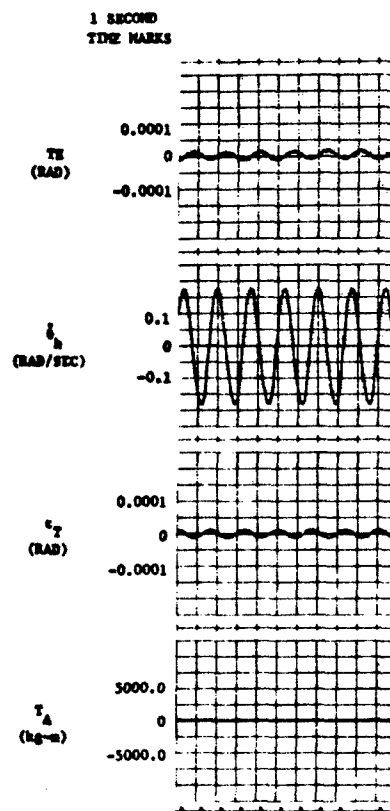


Figure 6-33. Response to Sinusoidal Hull Motion - Azimuth Rate Control with Electric Gyros

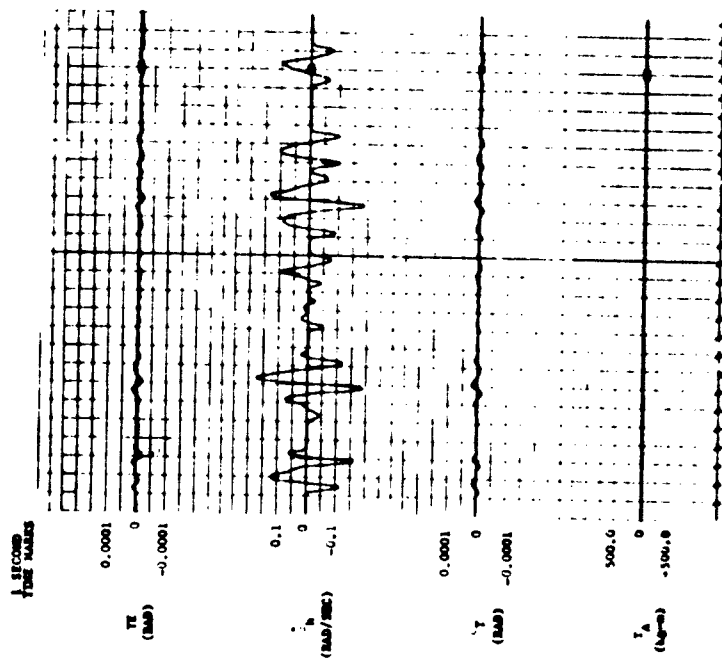


Figure 6-34. Response to HITPRO Bump Course - Azimuth Rate Control with Electric Gyros

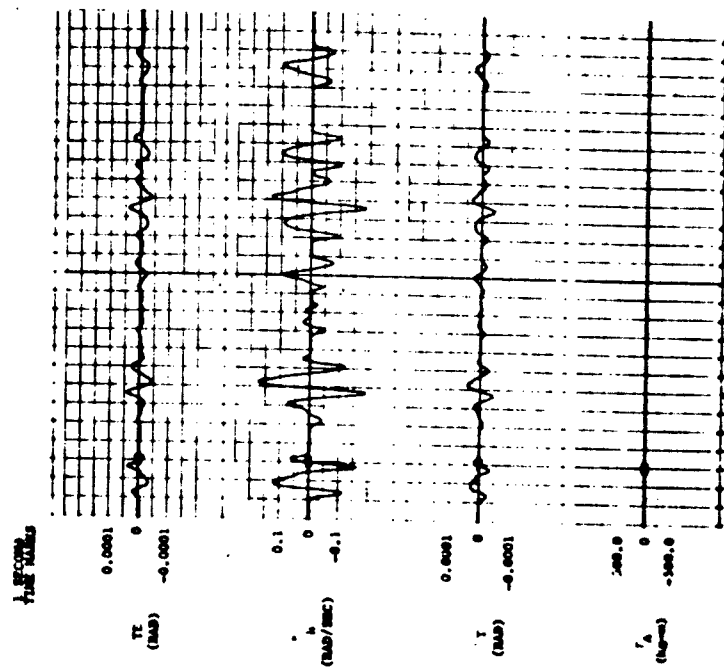


Figure 6-35. Response to HITPRO Bump Course with Hull Sensor Gain Error - Azimuth Rate Control with Electric Gyros

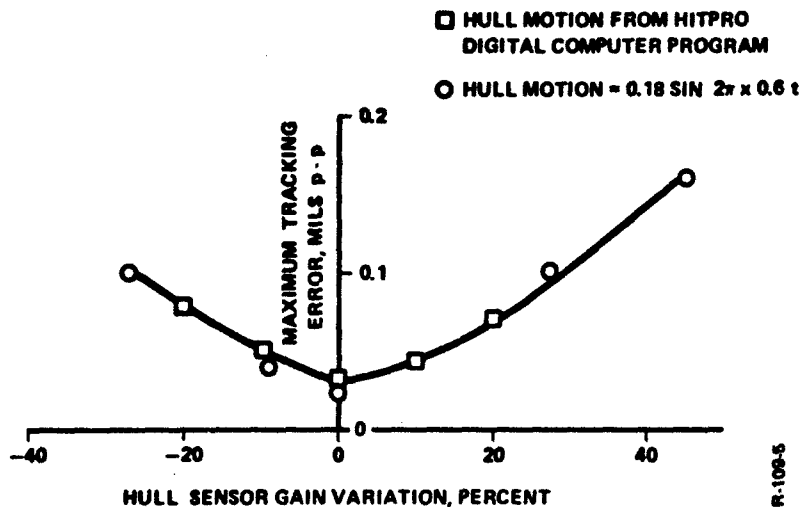


Figure 6-36. Tracking Error Versus Hull Sensor Gain Error - Azimuth Rate Control with Electric Gyros

#### 6.6.2 Evaluation of Pneumatic Accelerometer

A block diagram of the stabilization system for the azimuth axis using a pneumatic accelerometer is presented in Figure 5-7. The accelerometer is used for sensing turret acceleration, and the sensor output signal is integrated to obtain the turret rate signal.

A high gain loop was used to reduce the tracking error. Compensation was used in the forward path for reducing the gain at high frequency to achieve stability. In addition, feedback compensation was used to reduce the effects of sensor phase lag.

The gains and compensation networks used in the simulation were as follows.

$$K_i = 35$$

$$K_r = 10$$

$$K_p = 2 \times 10^{-4}$$

$$\text{Forward compensation: } 10 \frac{1 + 0.05 s}{1 + 0.5 s}$$

$$\text{Feedback compensation: } \frac{1 + 0.010 s}{1 + 0.002 s}$$

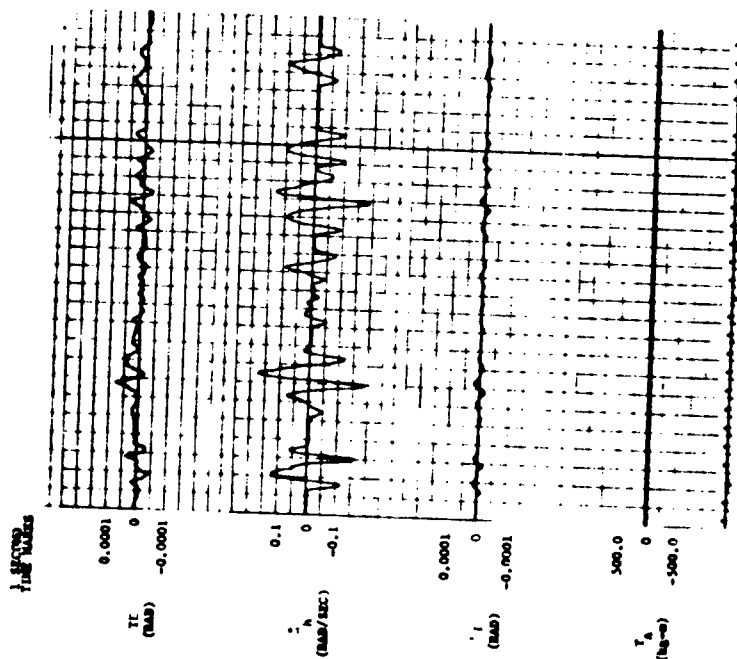


Figure 6-37. Response to HITPRO Bump Course with 2 ms Sensor Deadband - Azimuth Rate Control with Electric Gyros

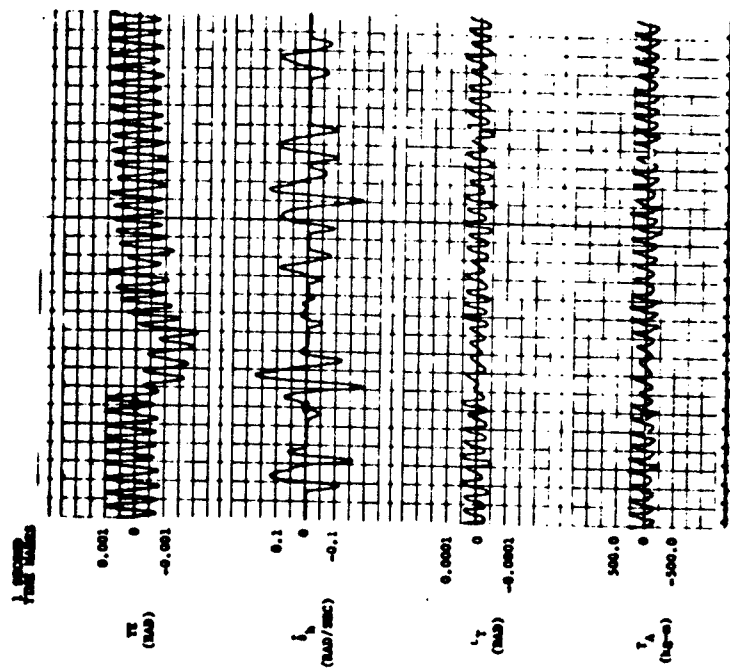


Figure 6-38. Response to HITPRO Bump Course with 6 ms Sensor Deadband



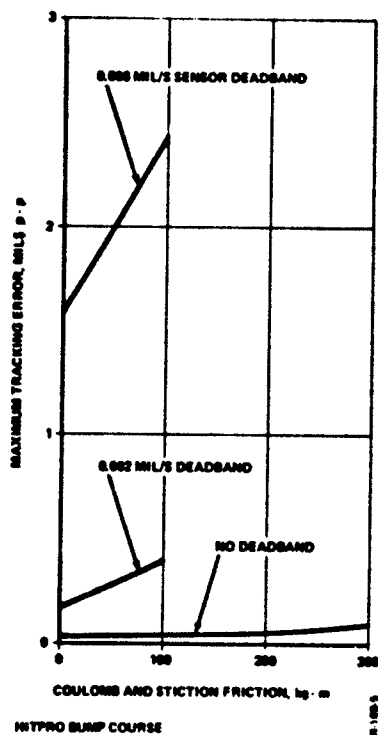


Figure 6-39. Effects of Deadband and Friction on Tracking Error - Azimuth Rate Control with Electric Gyros

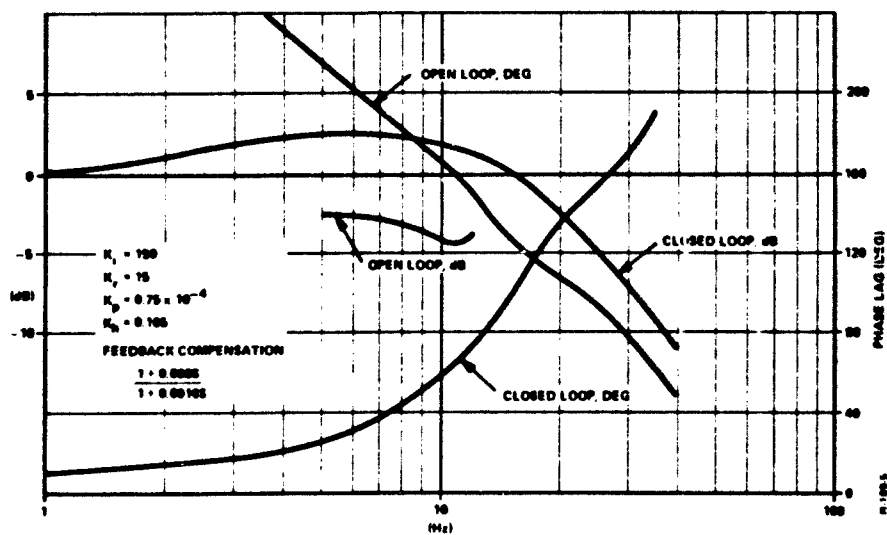


Figure 6-40. Frequency Response - Azimuth Rate Control with Electric Gyros

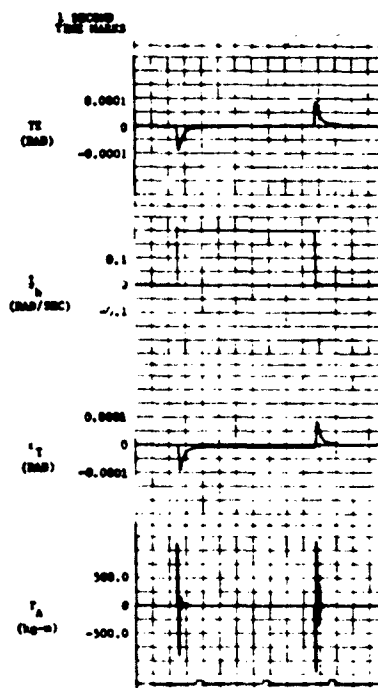


Figure 6-41. Response to Pivot Steer Maneuver - Azimuth Rate Control with Electric Gyros

The step response obtained with the hybrid computer simulation is shown in Figure 6-42. An overshoot of 44 percent was observed.

The response to sinusoidal hull motion equivalent in amplitude to the maximum hull motion for the bump course is shown in Figure 6-43. The observed tracking error was 0.2 mil peak to peak.

The response to the HITPRO bump course is shown in Figure 6-44. The effect of adding 100 kg-m coulomb friction and 100 kg-m stiction friction is shown in Figure 6-45. The resulting maximum tracking error was again 0.25 mil. The friction caused an increase in the maximum tracking error for larger values of friction, as shown in Figure 6-46.

The frequency response curves are shown in Figure 6-47. The frequency response results can be summarized as follows.

Bandpass:	15 Hz
Gain margin:	0.55
Phase margin:	36 deg

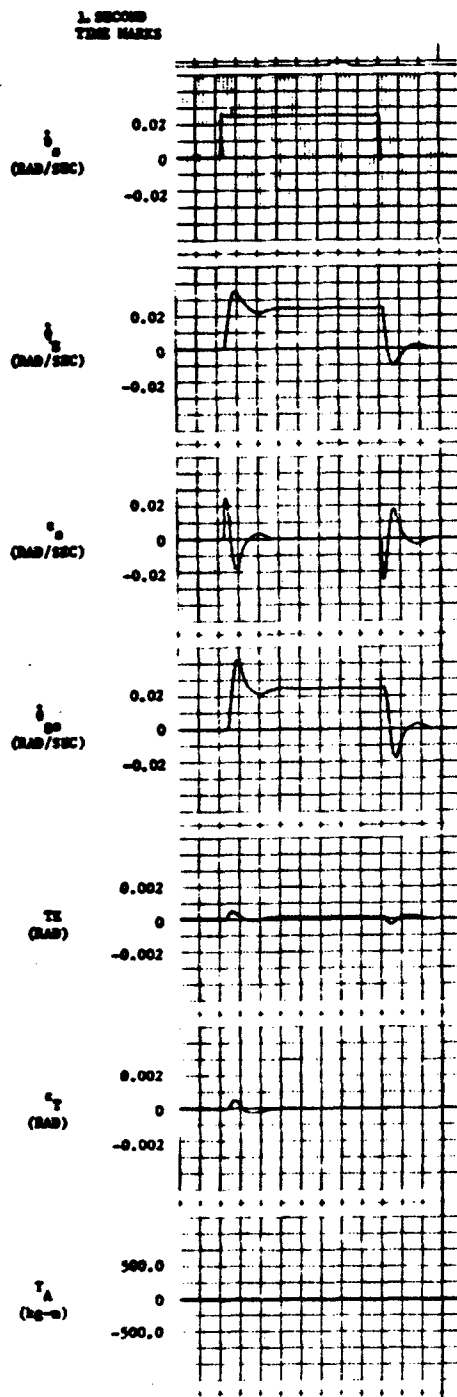


Figure 6-42. Step Response - Pneumatic Accelerometer

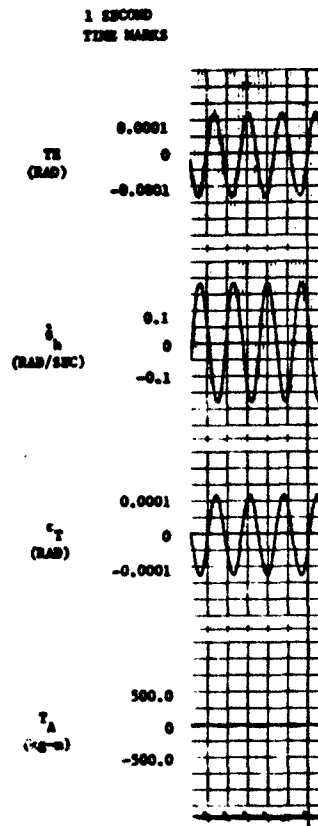


Figure 6-43. Response to Sinusoidal Hull Motion - Pneumatic Accelerometer

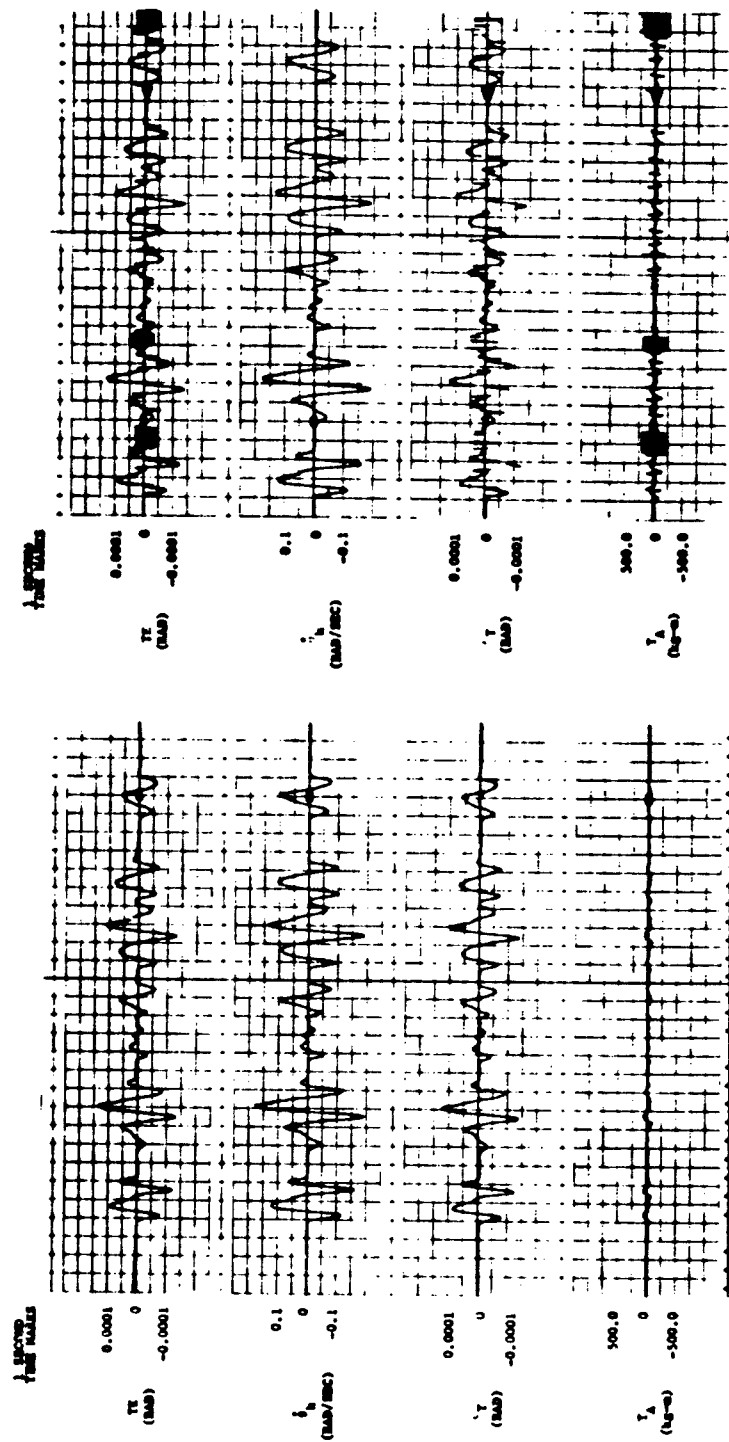


Figure 6-44. Response to HITPRO Bump Course - Pneumatic Accelerometer

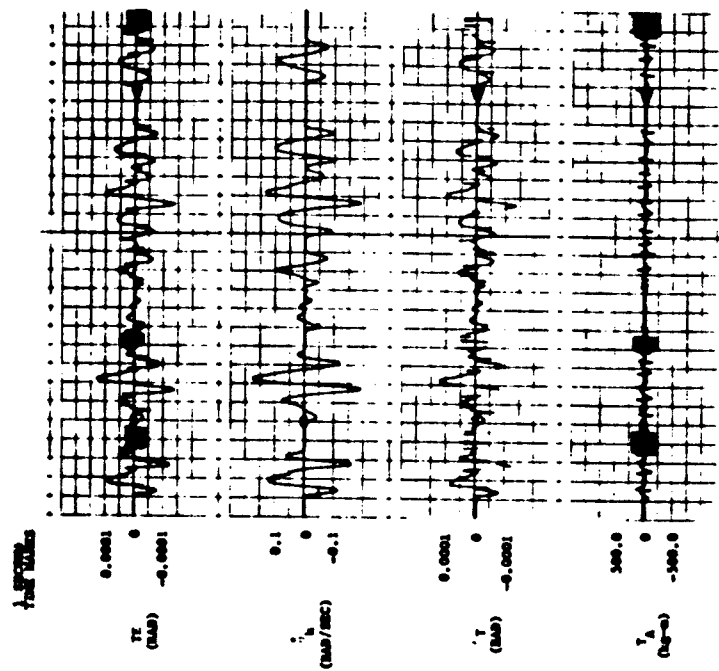


Figure 6-45. Response to HITPRO Bump Course with Coulomb and Stiction Friction - Pneumatic Accelerometer

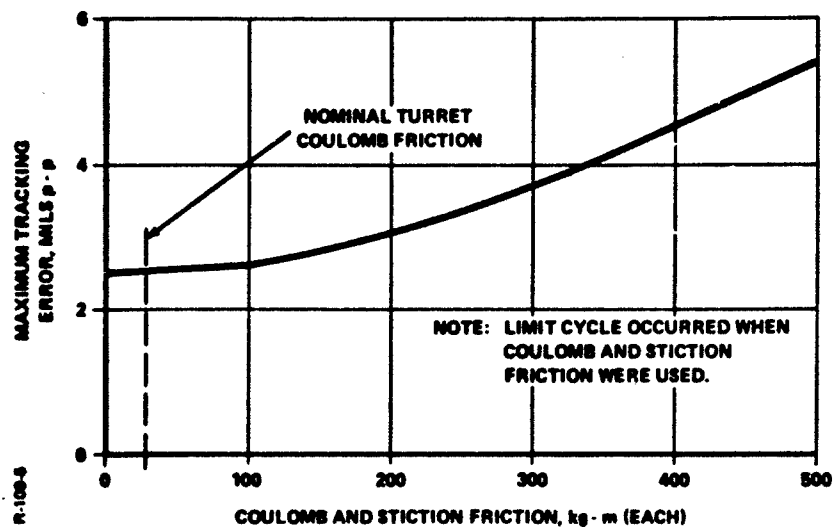


Figure 6-46. Maximum Tracking Error Versus Friction - Pneumatic Accelerometer

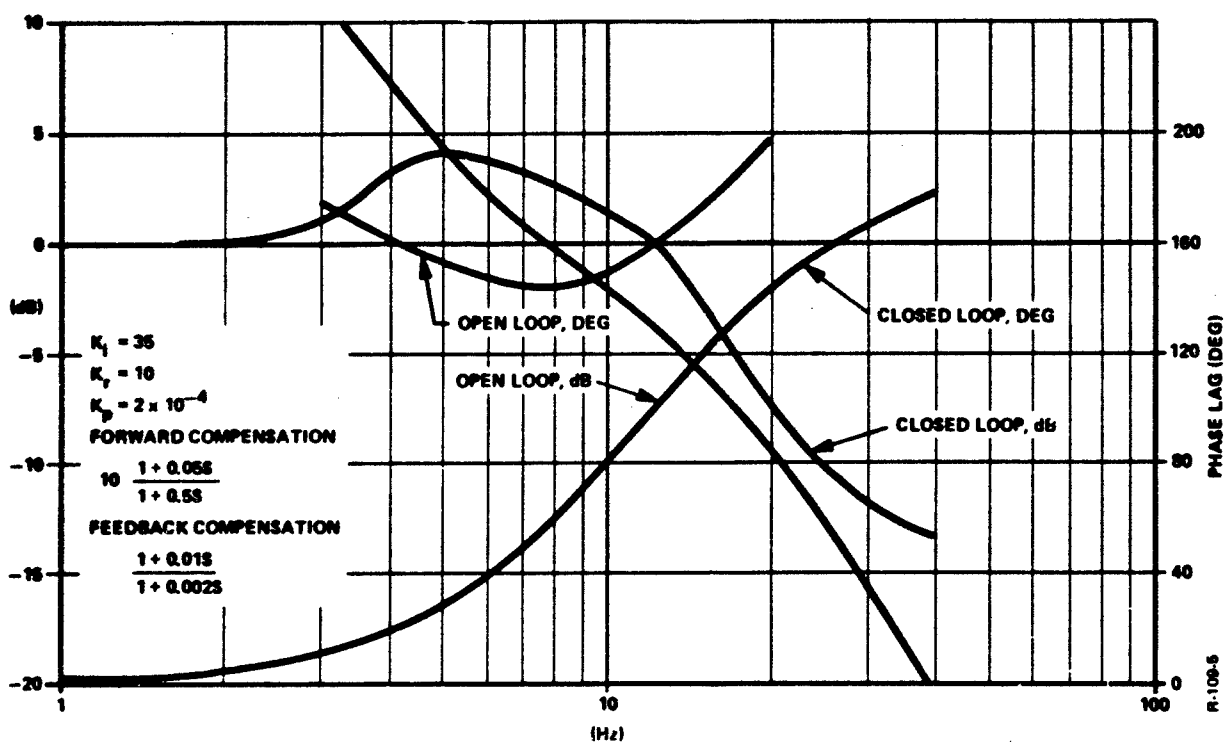


Figure 6-47. Frequency Response - Pneumatic Accelerometer

## 6.7 SUMMARY OF SIMULATION RESULTS

Simulation of the stabilization system has produced the following results.

- (1) The linear servovalve-actuator model provided a means of estimating required system gains but was not sufficiently accurate to be used in the sensor evaluation study. The nonlinear model was therefore used for sensor evaluations.
- (2) Hull dynamics had a negligible effect on system response and tracking error.
- (3) A summary of the results of the sensor evaluation study is presented in Table 6-1.

Table 6-1(A) summarizes the results for the rate sensors. For each of these systems, a rate sensor was mounted in the gun and the hull axis.

Table 6-1(B) summarizes the results for the accelerometers studied. For both of these systems, an accelerometer was mounted in the gun axis only.

It is to be noted that the systems represented in the table were not fully optimized in this study. An absolute comparison of the sensors is therefore not entirely justified.

As indicated by Table 6-1, the performance criteria were met for all of the sensor configurations.

- (4) A hull sensor gain error of 25 percent can be tolerated without exceeding the tracking error requirement.
- (5) Coulomb friction of 100 kg-m increases the tracking error by approximately 0.5 mil peak to peak.

Table 6-1 - Results of the Sensor Evaluation Study

A. Rate Sensor Results - Gun and Hull Sensor

Axis	System Command	Sensors	Gains				Bandpass (Hz)	Phase Margin (deg)	Gain Margin	Max. Tracking Error (Mils P-P)	Step Response Overshoot (percent)
			$K_i$	$K_r$	$K_p$	$K_h$					
Elevation	Rate	2 Elec. Gyros	50	7.5	$2.5 \times 10^{-4}$	0.15	28	72	0.50	0.18	15
Elevation	Rate	2 G.E. Fluidic	50	7.5	$2.5 \times 10^{-4}$	0.15	13	50	0.53	0.11	20
Elevation	Rate	2 Honeywell Vortex	40	4.8	$2.0 \times 10^{-4}$	0.15	25	50	0.50	0.21	30
Azimuth	Rate	2 Elec. Gyros	50	4.0	$0.2 \times 10^{-4}$	0.165	14	45	0.50	0.03	36
Elevation	Position	2 Elec. Gyros	50	7.5	$2.5 \times 10^{-4}$	0.135	28	72	0.50	0.13	15

B. Acceleration Sensor Results - Gun Sensor Only

Axis	System Command	Sensors	Gains				Bandpass (Hz)	Phase Margin (deg)	Gain Margin	Max. Tracking Error (Mils P-P)	Step Response Overshoot (percent)
			$K_i$	$K_r$	$K_p$	$K_h$					
Elevation	Rate	1 Bendix Int. Accel.	250	7.5	$2.5 \times 10^{-4}$	-	16	45	0.53	0.28	60
Azimuth	Rate	1 Aitesearch Accel.	35	10.0	$2.0 \times 10^{-4}$	-	15	36	0.55	0.25	44

NOTE: Tracking error values are for HITPRO bump course with nominal gains, no coulomb and stiction friction, and no deadband.

R-100-5

## SECTION 7

### CONCLUSIONS

A mathematical model of a suitable stabilization system for the M60A1 tank main gun was formulated and programmed. The model was defined so as to include most of the significant nonlinearities such as nonlinear valve flow and hull dynamics due to gun motion. A hybrid computer analysis was performed to determine the operating characteristics of the stabilization system, to evaluate prospective sensors for sensing gun and/or hull rate, and to determine whether a rate or position command control concept is preferable with respect to specified performance criteria.

An analytical study revealed that both the rate and position control concepts required a proportional plus integral control law in order to minimize the gun tracking error. It was also shown that the rate and position concepts are equivalent in terms of nulling out the effects of hull motions and thus of stabilizing the gun after the target is in the sight. A computer analysis which followed verified this equivalence. In addition, it was possible to show that the effect of hull motions on the system can be minimized by either a high control loop gain along with a lead-lag compensation network or by using a hull sensor signal in the control law.

The extensive computer simulation analysis revealed several conclusions in the areas of stabilization control philosophy and sensor applicability. In the process of arriving at a full computer model of the system for sensor evaluation, it was found that the effect of hull dynamics on the gun was negligible. Nonlinear valve flow, however, was found to have a significant influence on system performance. A linearized flow model was not sufficiently accurate for use in this study.

It was found that all five of the sensors studied in this program meet the performance criteria set forth by the Contracting Officer's Representative. In addition, this study indicates that these criteria can be met by using only a gun sensor. If verified by further studies, the need for a corresponding hull sensor may be eliminated.

The detailed sensor study immediately revealed that automatic offset and integrator drift nulling circuits are required when using an acceleration sensor. A method which can be used for this purpose is described in Section 5.5.

In addition, increasing the gain of the acceleration sensor will decrease the sensor offset effects, and hence the drift rate. More generally, it was found in the sensor study that sensor gain errors have a small effect on the tracking error. Also, a combination of sensor deadband and gun or turret friction will cause the system to limit cycle.

In order to compensate for sensor phase lag, feedback compensation is required. Forward path compensation is desirable for obtaining stability with higher loop gains for this system.



## SECTION 8

### RECOMMENDATIONS

It is recommended that in future system studies and sensor evaluations, additional system characteristics including realistic bending modes and hardware compliance model be added. The effects of providing an extensive model for friction which separates the effects of stiction coulomb, and running friction through switching logic should also be investigated. Sensor models should be utilized allowing for sensor errors due to noise, deadband, gain variation, and offset. It is also of importance on a complete system model to include integrator drift and an automatic nulling circuit. When these effects have been included in the model, a statistical analysis of the pointing error output data will be required in order to afford more detailed and objective performance comparisons. Inclusion of the additional nonlinearities mentioned will also make it possible to better optimize the system gains and compensation networks.

It is also recommended that a gunner model be included in the tracking loop, in order to arrive at additional data for comparing the rate and position command control concept. A display of tracking error can be developed using an oscilloscope for which a human operator can issue realistic commands to the system to simulate target tracking. Inclusion of an operator model in the simulation will allow a determination of the tracking and stabilization capabilities of the system.

In addition, it is recommended that steps be taken to verify the results of this simulation study by means of field test data with an M60A1 tank. Additional insight into the system operating characteristics and the effects of significant nonlinearities could be gained. A verification study would consist of obtaining recordings of measurements of hull and gun rates and achieved pointing accuracies on various types of terrain, and comparing these with the results of this study.

An optimal control theory approach to the stabilization of the M60A1 tank gun may also be desirable in the future if additional performance specifications are defined for the gun or if conventional techniques fail to achieve the goals when additional nonlinearities are included.

SECTION 9  
BIBLIOGRAPHY

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7. Final Report - Product Improvement Test of Gun Mount, 105-MM, M140, for M601A Tank, Report No. APG-MT-4064, Aberdeen Proving Ground, May 1972.

**APPENDIX A**  
**ACTUATOR SIZING STUDY AND**  
**DIGITAL CHECK SOLUTIONS**

**PART I**  
**ACTUATOR SIZING**  
**STUDY**

[illegible]

PAGE 1

SYSTEMS REAL-TIME MONITOR-4.0

CO. 03.40 ACTUATOR

8308 ACTUATOR-03.03  
80PT/1/5.3  
NEWLOGIC FORTRAN

[illegible]





SYSTEMS 9506 FORTRAN IV COMPILER (REV D)

PAGE 4

```

1  C
2  SUBROUTINE DIFFC(V,VP,T,N)
3  DIMENSION V(4),DV(4),VH(4),VP(4),RI(4,4)
4  REAL T
5  COMPLEX TARESON,C2,J,OP,OT,VL,TH
6  TH=(ON + V(4) - C2*V(1)*V(2))/12
7  VP(1)=TA/J
8  VP(2)=V(1)
9  OP=ON*V(1)
10 VP(3)=0
11 IF(OT GT OP) VP(3)=ON-OP
12 IF(V(3) GT VH)V(3)=VH
13 VP(4)=-1 + V(4) + VP(3)/V(3)
14 RETURN
15 END

```

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01130  
01140  
01150  
01160  
01170  
01180  
01190  
01200  
01210  
01220  
01230

```

1  C *** SUBROUTINE RUC(J,V,VO,DV,XOS,DX,VL,VP,NL,AL,DIFEC)
2  C *** N SPECIFIES NUMBER OF EDIS K=1 PROVIDES VP'S AT END POINTS
3  C *** DIMENSION YOUN, DY(N), VN(N), VP(N), AL(N,4)
4  C *** DOUBLE PRECISION XO
5  C *** XO = XOS
6  C *** I = 1
7  C *** GO TO (1,2,3,4,5),I
8  C *** 1 XA = XO
9  C *** GO 6 J = 1,N
10 C *** VN(J) = VO(J)
11 C *** CALL DIFEC(VL,VP,NL,N)
12 C *** DO 7 J = 1,N
13 C *** 7 AL(J,1) = DV*VP(J)
14 C *** I = I + 1
15 C *** GO TO 6
16 C *** 2 XA = XO + 0.5*DX
17 C *** DO 9 J = 1,N
18 C *** 9 VN(J) = VO(J) + 0.5*AL(J,1)
19 C *** GO TO 10
20 C *** 3 GO 11 J = 1,N
21 C *** 11 VN(J) = VO(J) + 0.5*AL(J,2)
22 C *** GO TO 10
23 C *** 4 XA = XO + DX
24 C *** DO 12 J = 1,N
25 C *** 12 VN(J) = VO(J) + AL(J,3)
26 C *** GO TO 10
27 C *** 5 DO 13 J = 1,N
28 C *** DV(J) = (0.5*AL(J,1)+AL(J,4))/AL(J,3)+AL(J,3))/2.0
29 C *** 13 VO(J) = VO(J) + DV(J)
30 C *** IF (K.EQ.1) RETURN
31 C *** CALL DIFEC(VL,VP,NL,N)
32 C *** RETURN
33 C *** END
34

```

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01390  
01400  
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01470  
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01490  
01500  
01510  
01520  
01530  
01540  
01550  
01560  
01570

```

1 SUBROUTINE SLOT(X,V,IP,XMIN,XMAX,VMIN,VMAX,NT,NC)
2 DIMENSION IPT(76),SCLX(6),IDR(9),X(INC,IP),V(INC,IP)
3 DATA IDR/1,2,3,4,5,6,7,8,9/
4 IF(IP.LT.2)RETURN
5 X=X/255.
6 V=V/255.
7 XMIN=XMIN
8 VMIN=VMIN
9 IF(NT.NE.0)GO TO 50
10 VMAX=VMAX
11 XMAX=X(1.1)
12 VMAX=V(1.1)
13 DO 28 J=1,IP
14 DO 28 I=1,NC
15 IF(X(I,J).GT.XMAX)VMAX=X(I,J)
16 IF(V(I,J).GT.VMAX)VMAX=V(I,J)
17 IF(X(I,J).LT.XMIN)VMIN=X(I,J)
18 IF(V(I,J).LT.VMIN)VMIN=V(I,J)
19 CONTINUE
20 CONTINUE
21 XFACT=75./XMAX-VMIN
22 VFACT=45./VMAX-VMIN
23 BITX=(XMAX-XMIN)/3.0
24 BITV=(VMAX-VMIN)/3.0
25 SCLX(1)=XMIN
26 SCLV(1)=VMIN
27 DO 30 K=2,5
28 SCLX(K)=SCLX(K-1)+BITX
29 SCLV(K)=SCLV(K-1)+BITV
30 CONTINUE
31 SCLX(6)=XMAX
32 SCLV(6)=VMAX
33 PRINT 999,(SCLX(N),N=1,6)
34 DO 100 I=1,46
35 DO 110 J=1,76
36 IPT(I,J)=IDBK
37 CONTINUE
38 DO 111 J=1,76,15
39 IPT(J)=ILNE
40 CONTINUE
41 DO 120 K=1,NC
42 DO 120 J=1,NC
43 INT=(VMX-V(J,K))*VFACT+1.5
44 IF(INT.NE.1)GO TO 120
45 IX=X(I,J,K)-XMIN)*VFACT+1.5
46 IF(IX.GT.0.AND.IX.LE.76)IPT(IX)=IDR(J)
47 CONTINUE
48 DO 130 K=1,6
49

```

00.0.51 RETURN SYSTEMS REAL-TIME MONITOR-4.0

PAGE 7

SYSTEMS 05/86 FORTRAN IV COMPILER (REV. 6)

SLOT

```

54      INT=(VWZ-SOLV(K))*PCT+1.5
55      IF (INT EQ 1) GO TO 140
56      130 CONTINUE
57      PRINT 998, (IPT(J), J=L, 76)
58      GO TO 100
59      DO 150 J=L, 76
60      IF (IPT(J) EQ 1000) IPT(J)=1050
61      150 CONTINUE
62      PRINT 997, SOLV(K), (IPT(J), J=L, 76)
63      100 CONTINUE
64      PRINT 996, (SOLX(N), N=L, 6)
65      RETURN
66      997 FORMAT(' ', 3X, D14.6, 2X, 76H)
67      998 FORMAT(' ', 21X, 76H)
68      999 FORMAT('1', 9X, 6F13.2)
69      996 FORMAT(' ', 9X, 6F13.2)
70      END

```

```

00110
00111
00112
00113
00114
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00116
00117
00118
00119
00120
00121
00122
00123
00124
00125
00126
00127

```

```

EXTERNAL DIFEC
DIMENSION IOB(20)
DIMENSION AVV(20), VCC(20)
DIMENSION OTH(1000), OTHD(1000), DEGV(1000), TT(1000)
DIMENSION DUMMY(3)
DIMENSION Y(4), DY(4), YN(4), YP(4), AI(4,4)

```

REAL J

COMMON/NAMES/DH, C2, J, GP, DT, VM, TA

C	Y(1)=THD	YP(1)=THD0
C	Y(2)=TH	YP(2)=TH0
C	Y(3)=V	YP(3)=V0
C	Y(4)=PS	YP(4)=PS0

IN=0

A1 2

3 IN=IN+1

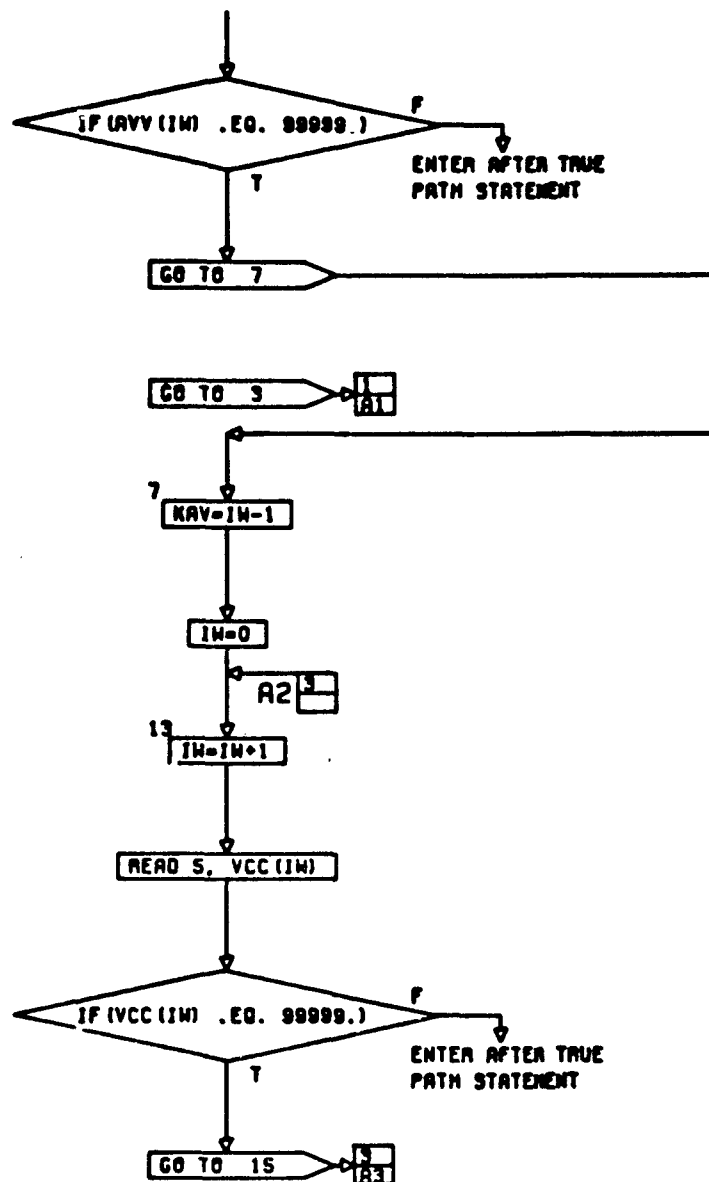
READ 5, AVV(IN)

5 FORMAT(F10.3)

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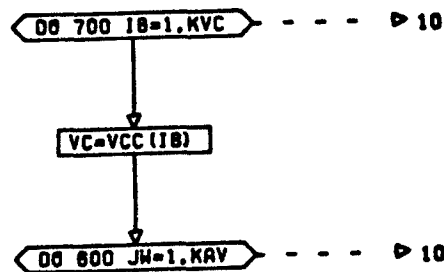
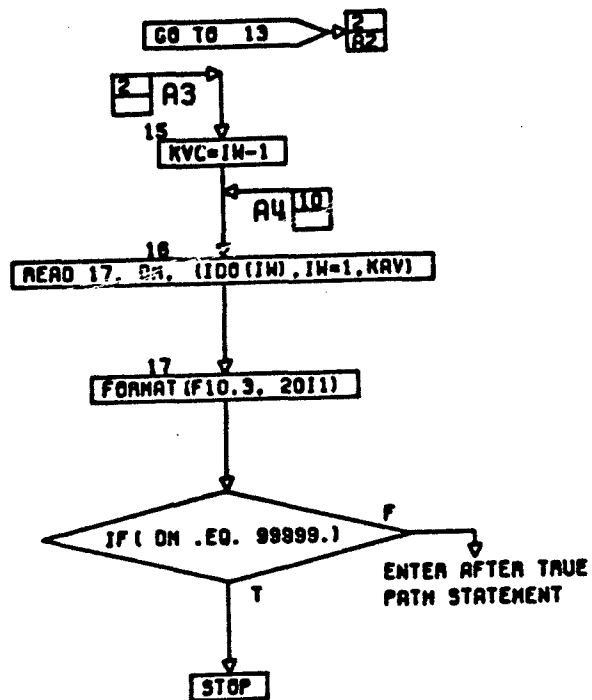
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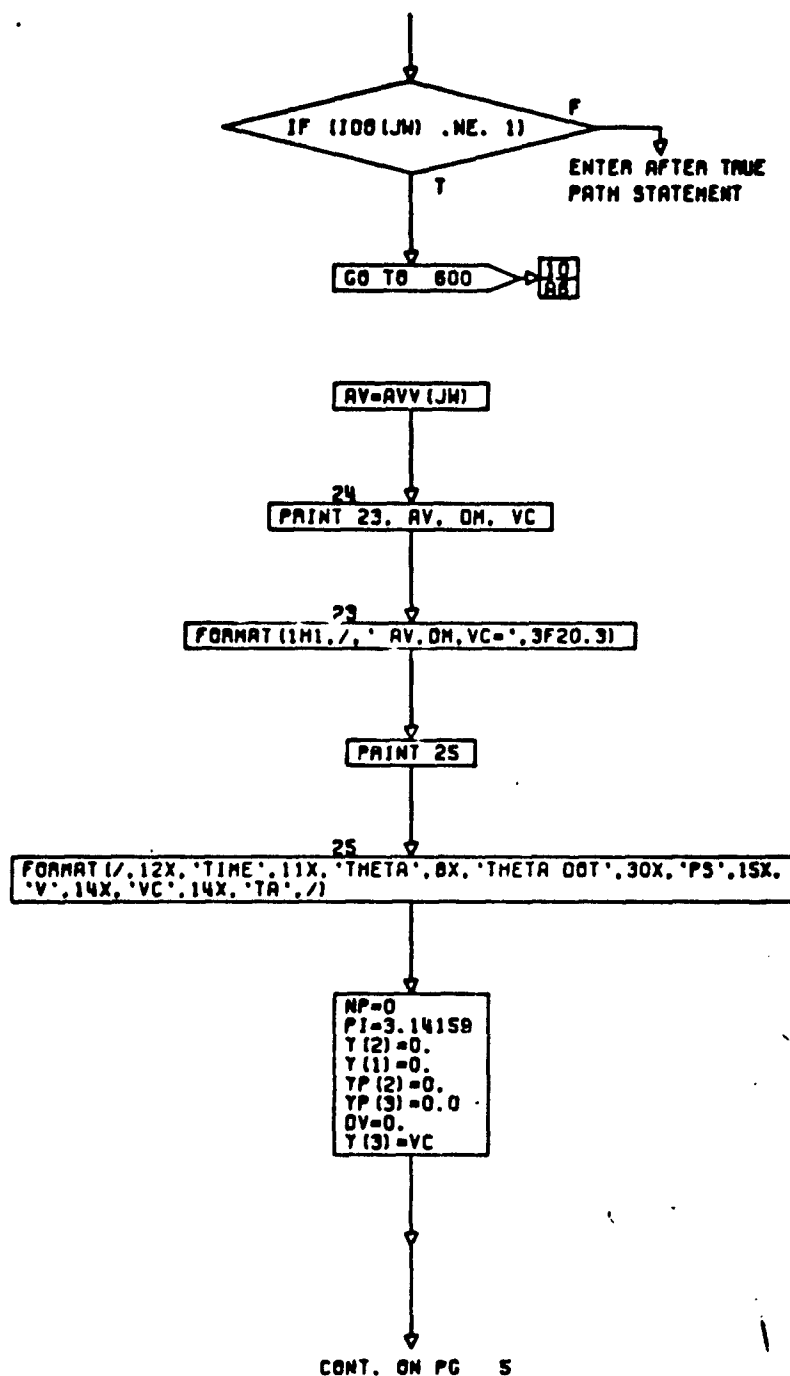
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PG 2 OF 10



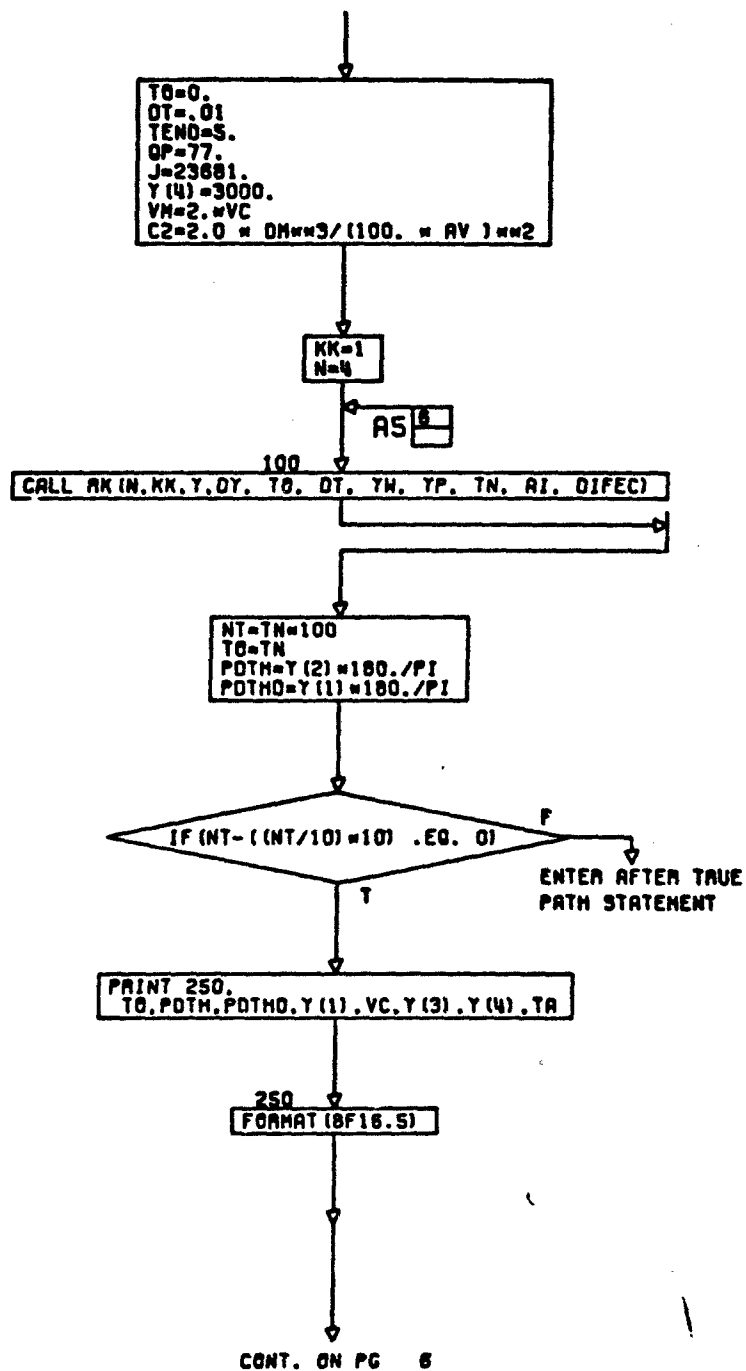
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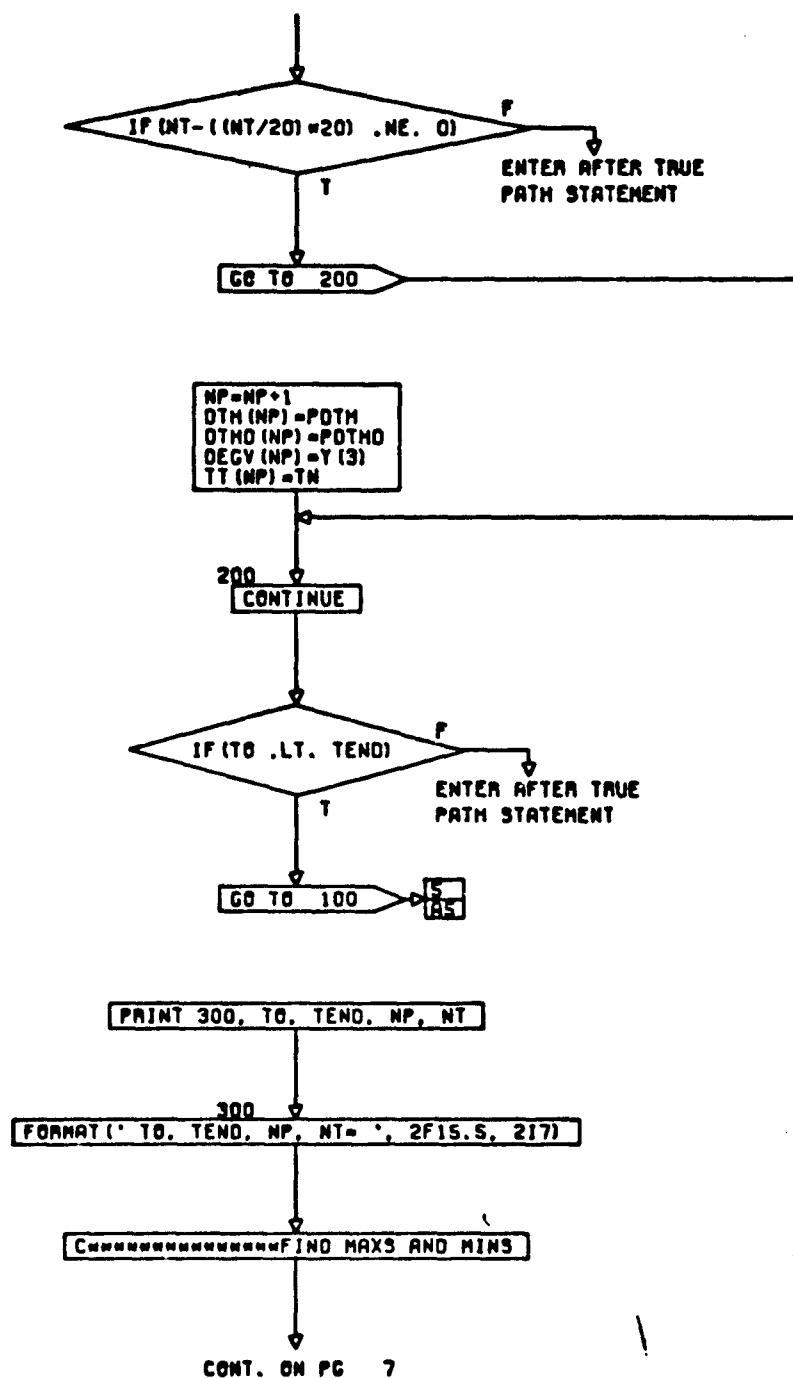


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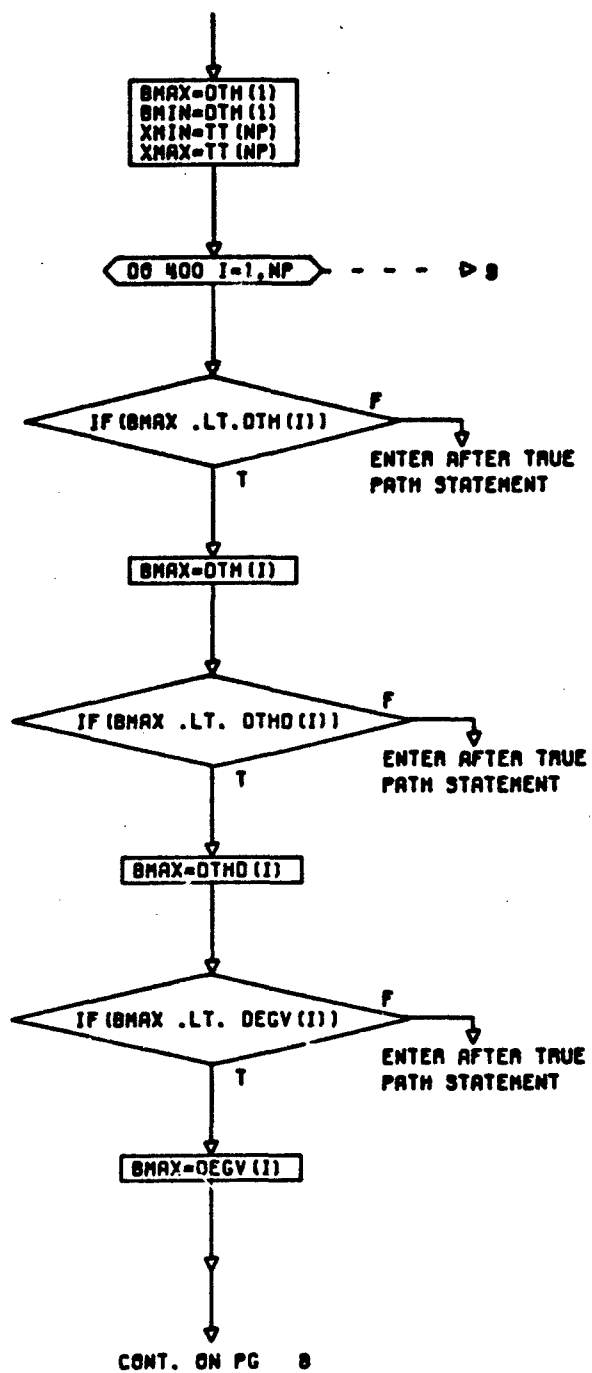


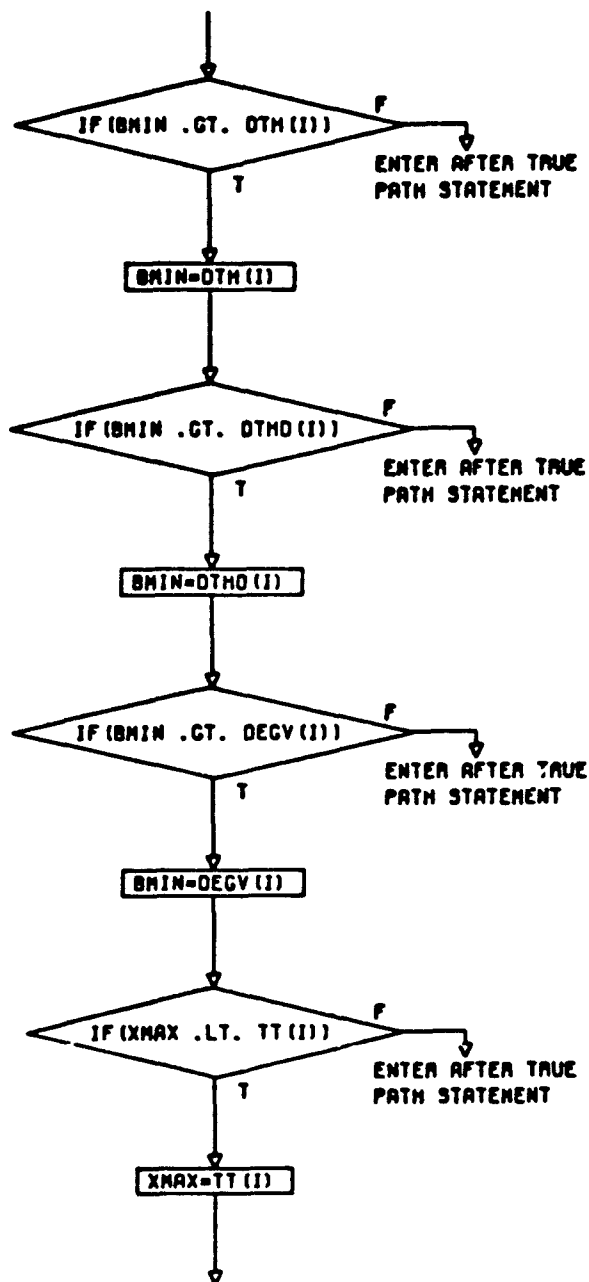


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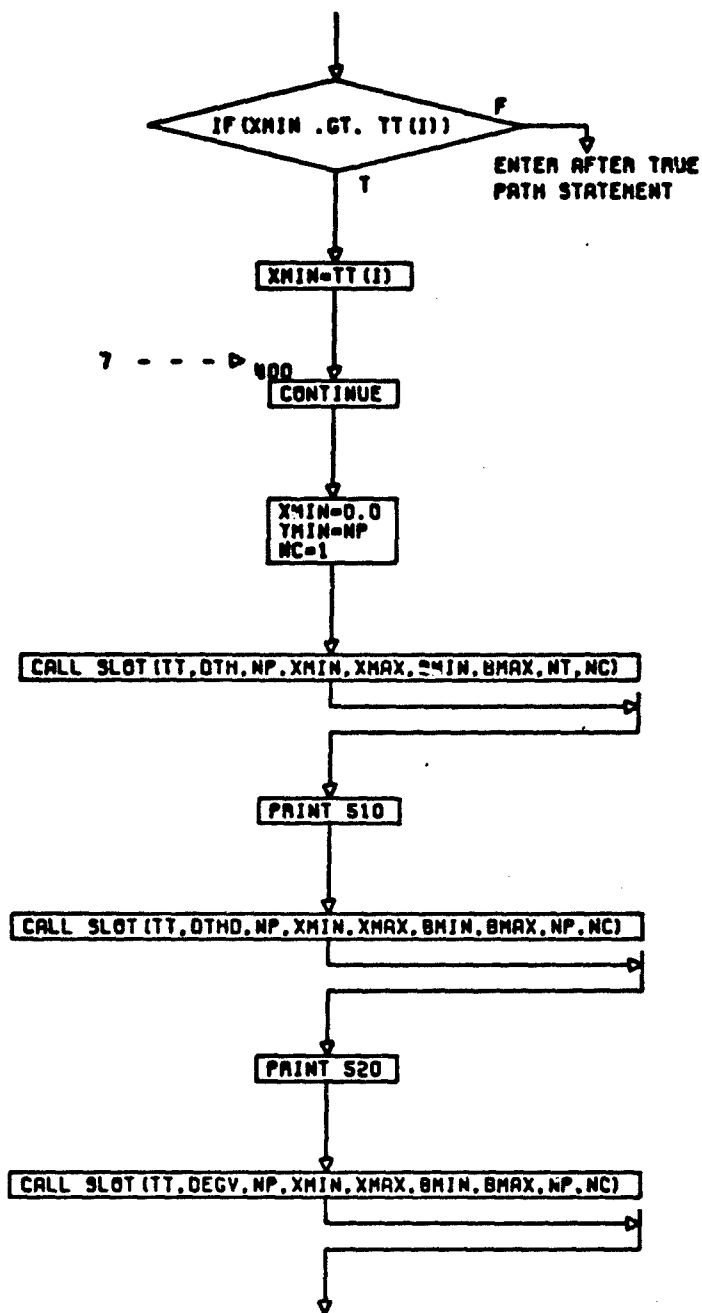
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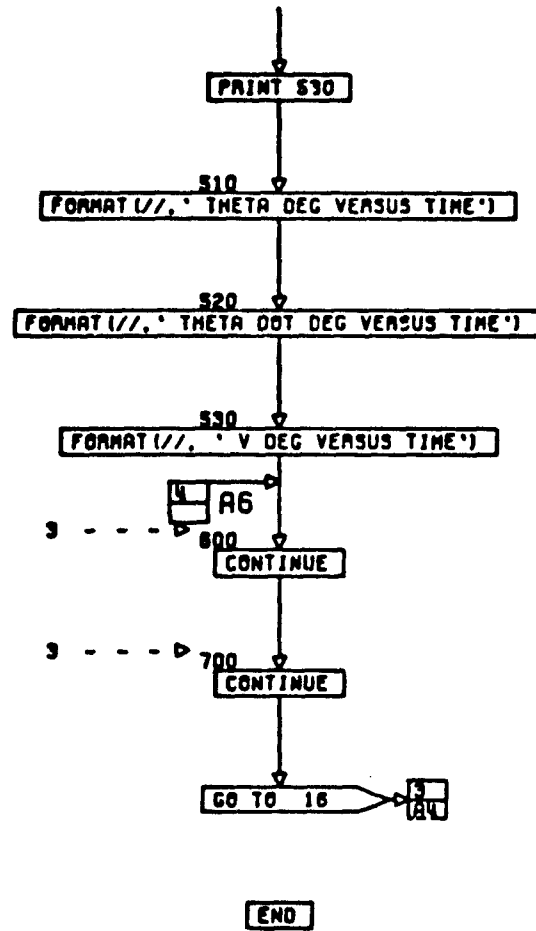
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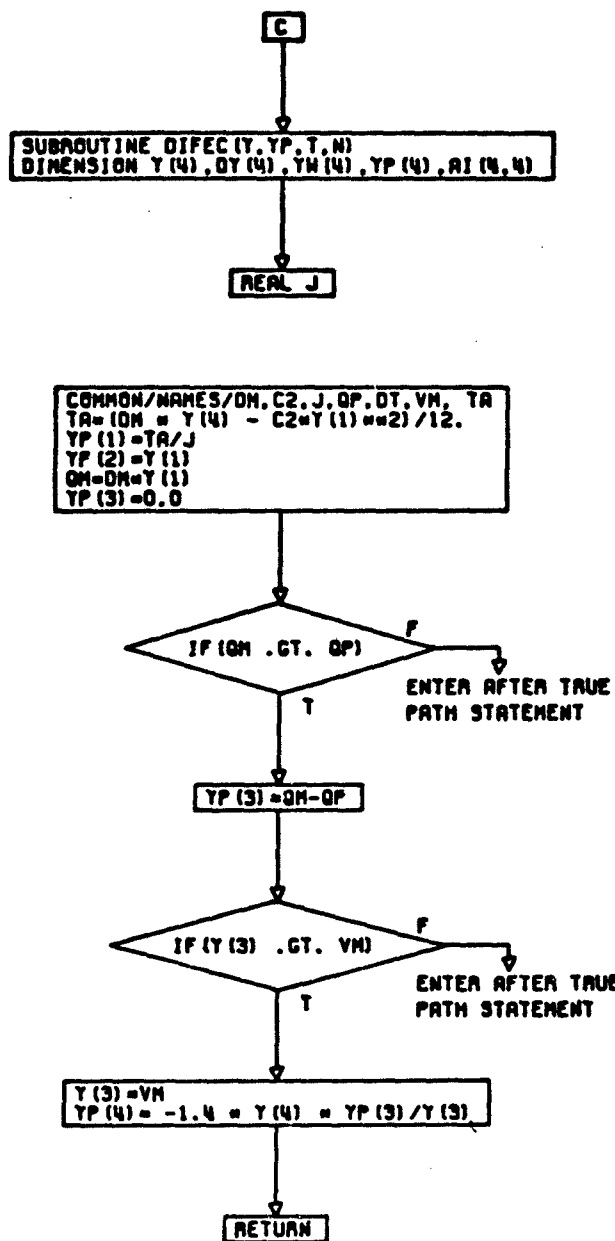


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PG 9 OF 10



PG 10 FINAL



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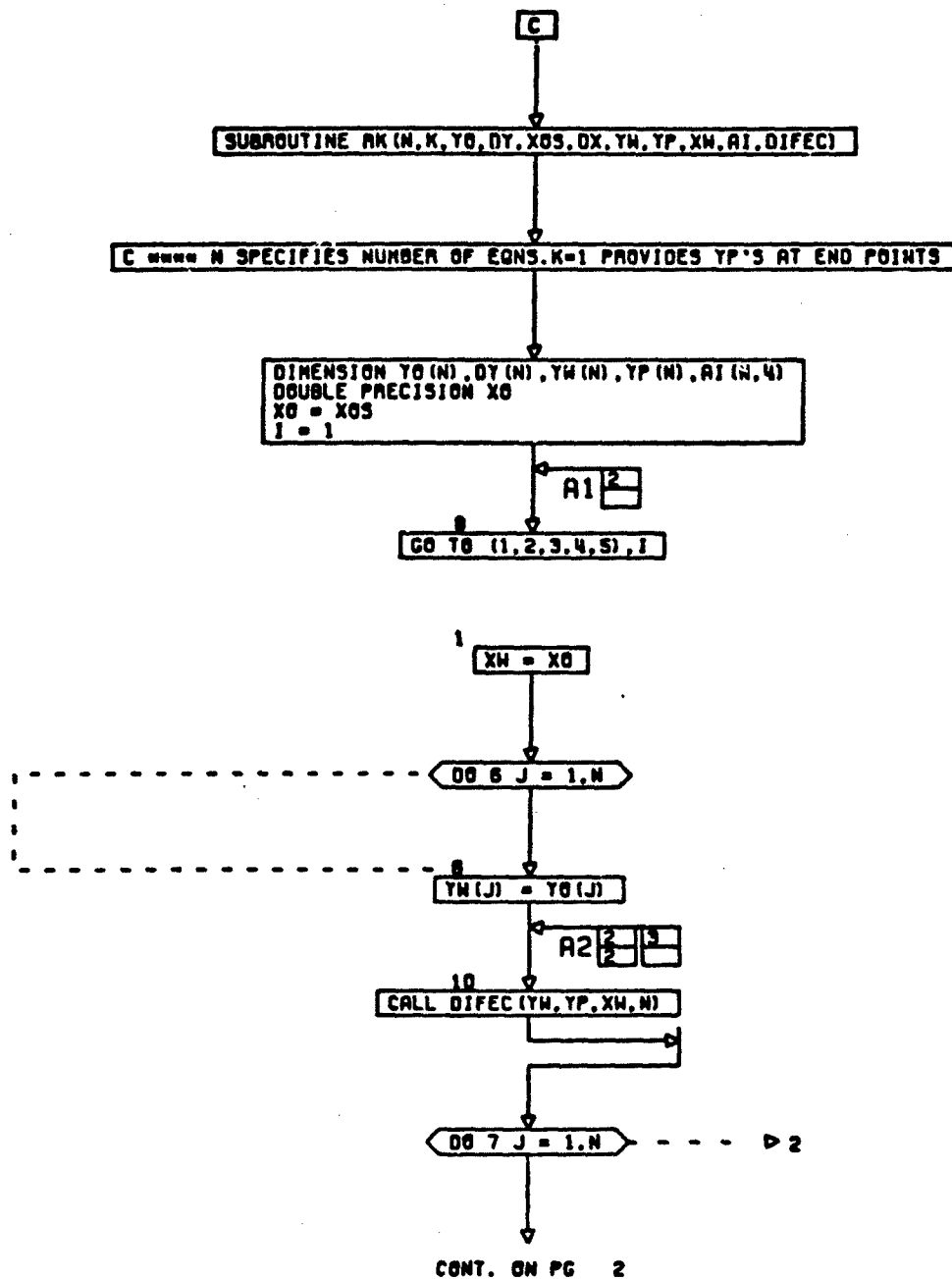
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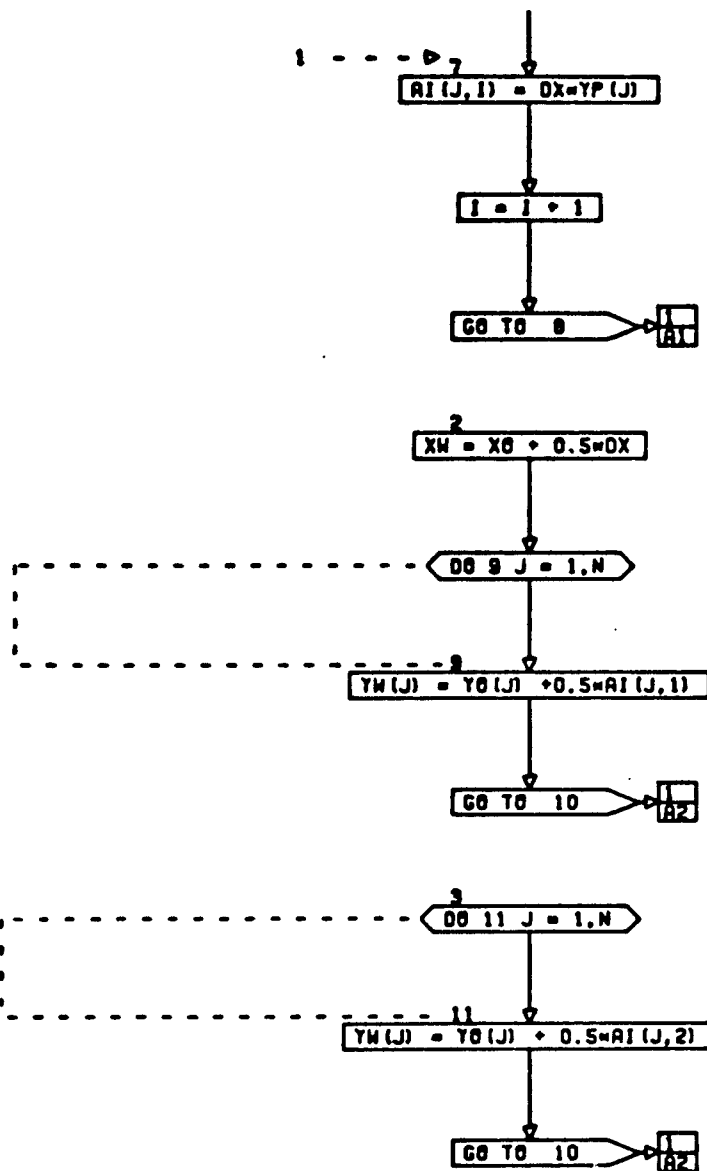
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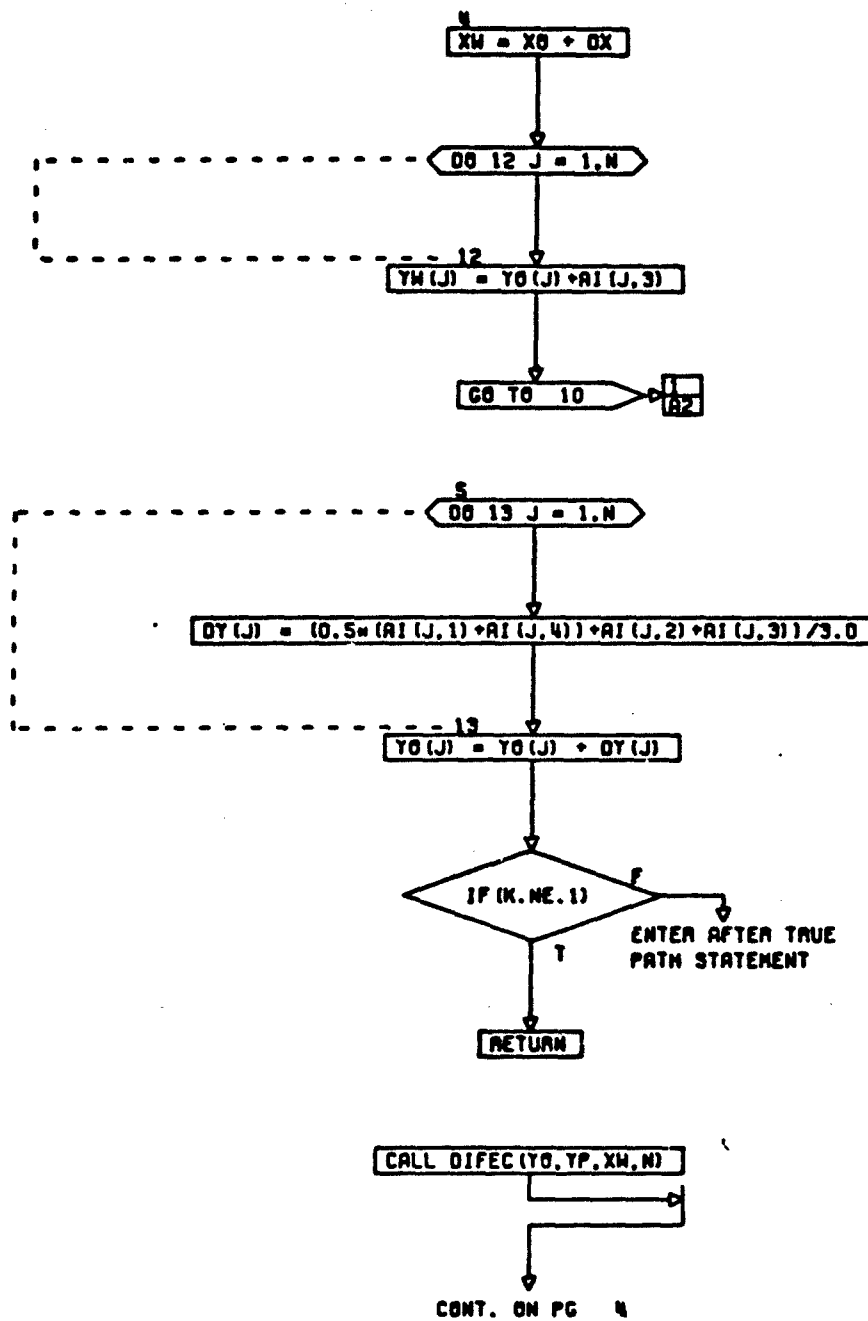






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PG 2 OF 4



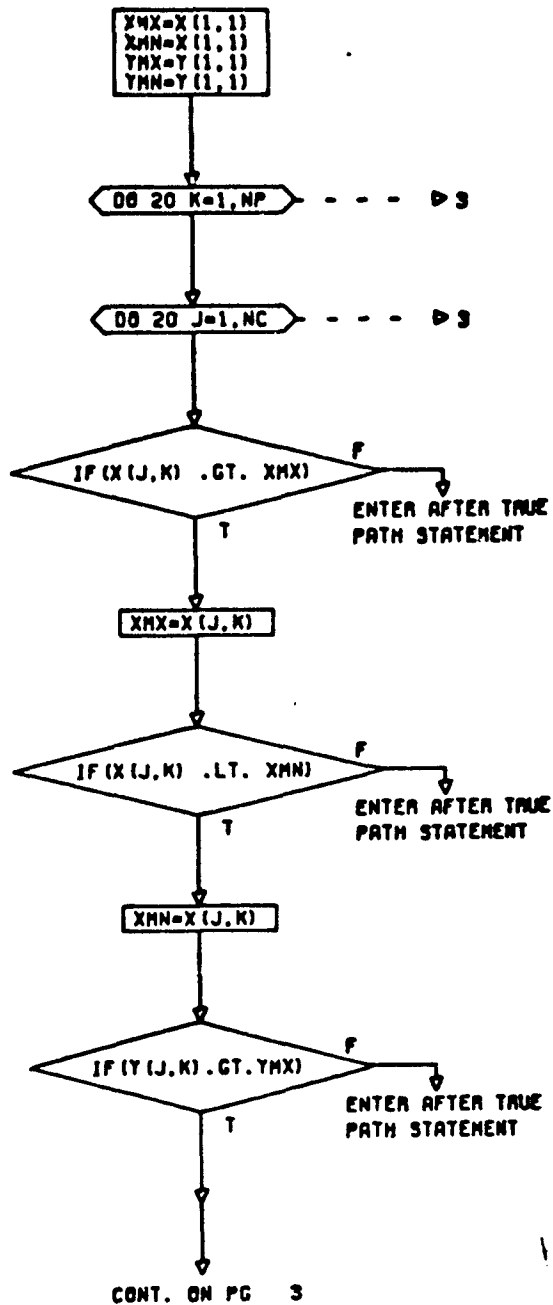
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↓  
RETURN

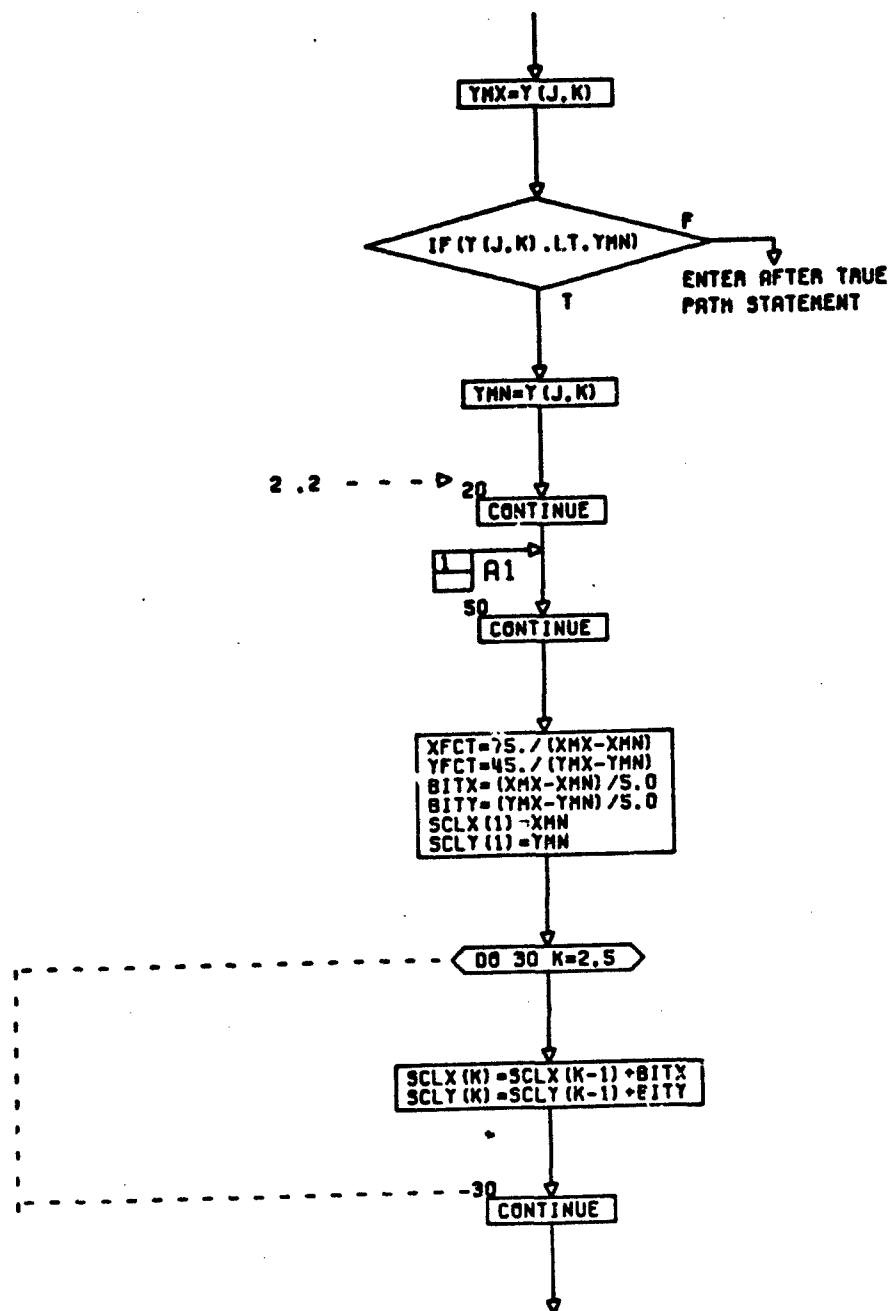
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CE 3 FINAL



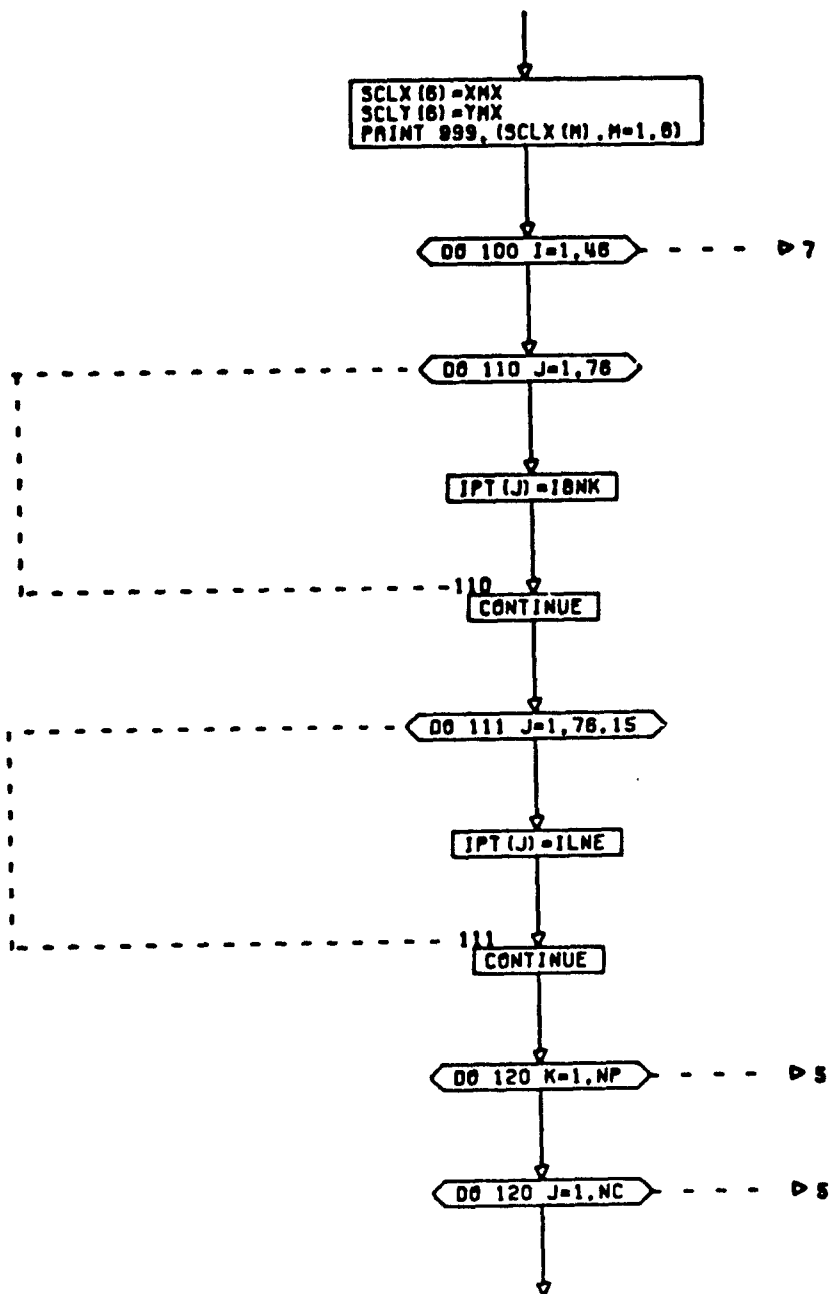


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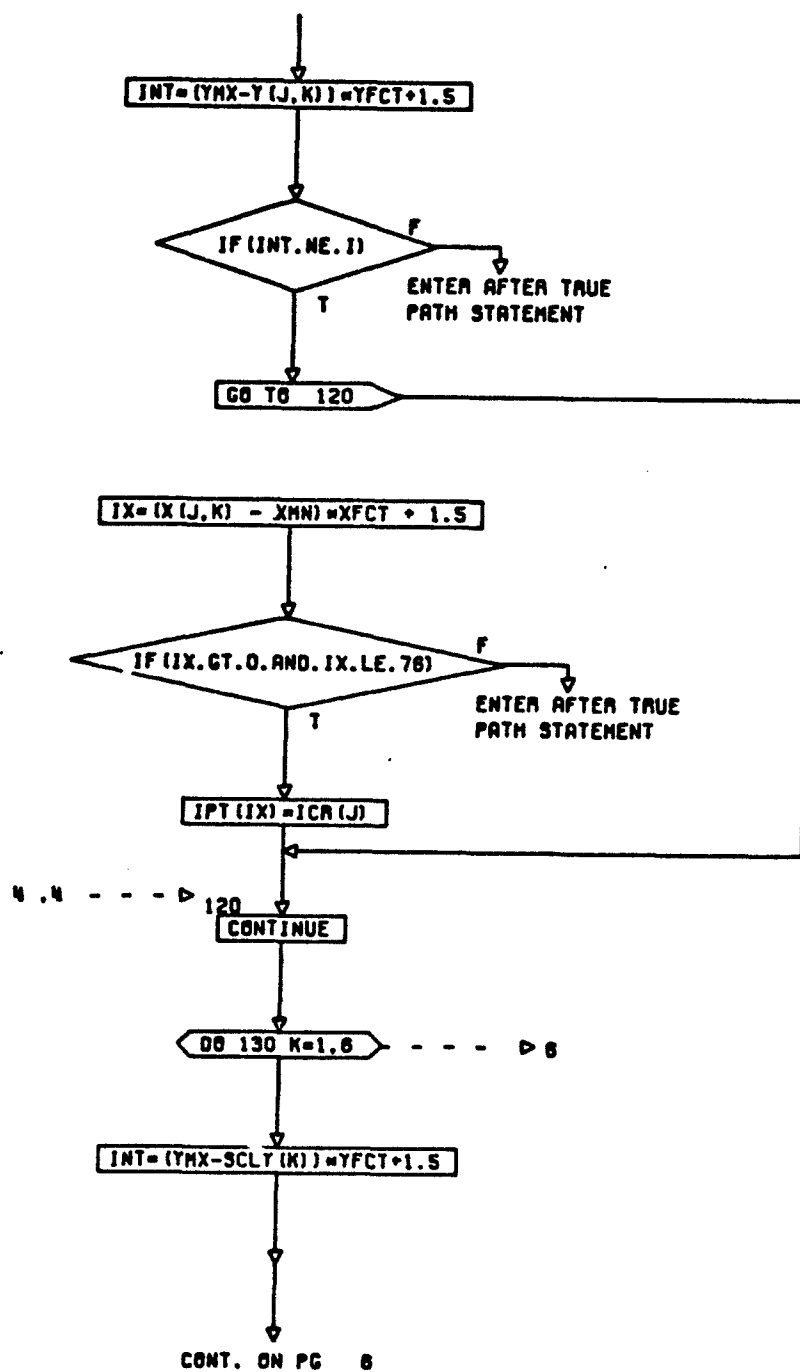
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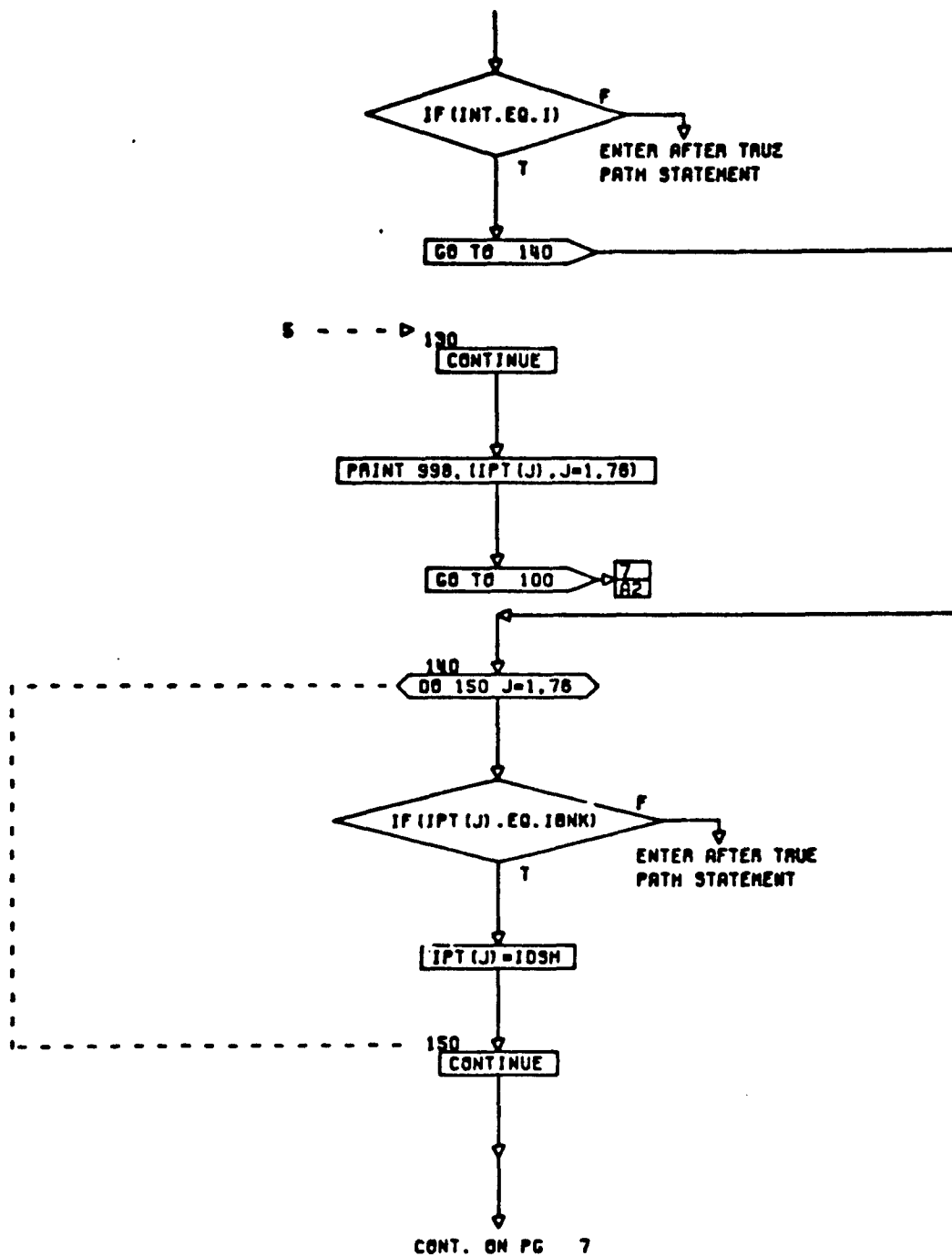
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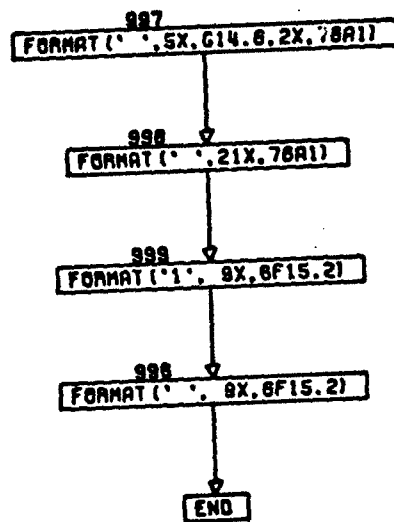
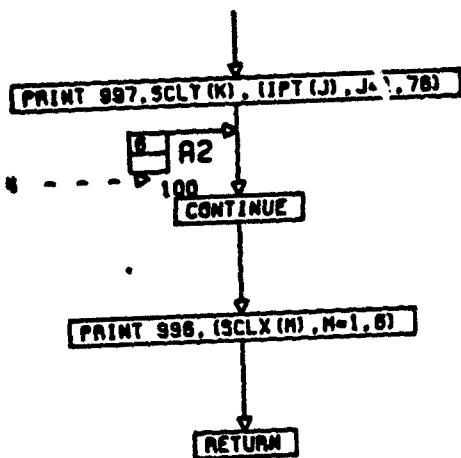




PG 5 OF 7



PG 8 OF 7



PG 7 FINBL

A-33

**PART II**  
**DIGITAL CHECK**  
**SOLUTIONS**

A-35

10/29/74 09:31:09 M3000

SYSTEMS REAL-TIME MONITOR-4.0

PAGE 1

11/15/74  
10:17:45  
SYSTEMS REAL-TIME MONITOR

00010

00060  
00070  
00080  
00090  
00100  
00120

00130

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1  EXTERNAL DIFCO
2  COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
3  COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
4  COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
5  COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
6  COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
7  COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
8  COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
9  COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
10 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
11 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
12 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
13 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
14 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
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17 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
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23 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
24 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
25 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
26 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
27 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
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32 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
33 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
34 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
35 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
36 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
37 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
38 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
39 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
40 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
41 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
42 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
43 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
44 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
45 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
46 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
47 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
48 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
49 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
50 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
51 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS
52 COMMON/INTEL/INTEL( 500), TOTL( 500), NOEL, TOLD, NON
53 COMMON/OUT/EP5, (IN, XIN), THIDT, TIGS

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12/19/74 09 01 03 M6001G

SYSTEMS 8386 FORTRAN IV COMPILER (REV. 0)

PAGE 3

IN IN

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54 MCS=MS$40 1
55 MCD=MD$40 1
56 MCH=MH$40 1
57 MOP=MP$40 1
58 TIME=0
59 LININT=1
60 ICHIT=CTP/DT *8.5
61 MEJ=21
62 DO 20 K=L,NEO
63 VPK=0
64 VCK=0
65 20 CONTINUE
66 T=0
67 PAR=0
68 CALL DIFEG(VP,T,NEO)
69 30 CONTINUE
70 IF (INGOT EQ 0) GO TO 31
71 TEMP=THO/THOUT
72 31 CONTINUE
73 LINCNT=LINCNT-1
74 IF (LINCNT LE 0) GO TO 40
75 LINCNT=50
76 PRINT 997, (NO1(J,NCS), J=L, 0), (NO2(L,NEO), L=L, 0)
77 1, (NO3(NCO,NOH,N), N=L, 6), (NO4(N,NOP), N=L, 7)
78 PRINT 997, (NO4(K), K=L, 20)
79 FORMATT(' ', 40X, 2004)
80 997 FORMATT(' ', 3104, /)
81 PPINT 999
82 PRINT 998, T, THO, EPT, TR, EPS, THOUT, THOCD, GFL XTHP
83 NPT=NPT+1
84 IF (NPT GT 350) NPT=350
85 XEL(0PT)=T
86 VPL(1,NPT)=EPT
87 IF (NCS EQ 1) GO TO 41
88 IF (NCS EQ 4) GO TO 41
89 VPL(2,NPT)=THO
90 VPL(3,NPT)=TR
91 GO TO 42
92 41 VPL(2,NPT)=THO
93 VPL(3,NPT)=TR
94 42 CONTINUE
95 FORMATT(' ', 9014, 6, /1304, 8014, 6)
96 998 IF (T GE THO) GO TO 200
97 DO 50 K=L, ICNT
98 LL=0
99 CALL RK(NEO, LL, V, DV, T, DT, VM, VP, TM, AI, DIFEG)
100 NUP=NUP+1
101 T=TM+DT
102 CALL DIFEG(V, VP, T, NEO)
103 PRINT 998, T, (V(J), J=L, NEO), (VP(L), L=L, NEO)
104 50 CONTINUE
105 GO TO 30
106 998 FORMATT(' ', F14, 3, F14, 4, F14, 5, F14, 6, 2F14, 4, F14, 4, F14, 5, F14, 3)

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00140  
00150  
0016000170  
00180

00200

00210  
00220  
0023000250  
0026000270  
00280

00300

00320

00330

00370  
00380



12/19/74 2:01 M3 NUCLEAR SYSTEMS 8506 FORTRAN IV COMPILER (REV. Q) SYSTEMS REAL-TIME MAIL (R-0) PAGE 4

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107 999 FUPR(1,1)=.9K,TIME=.6U,THRO DOT,10K,EPST=.12U,TR,
108 1 110,LP5=.6U,THIN DOT,.5K,THIN DOT,12U,TR,10K,TR...
109 203 GO TO 201,202,202,201,NC5
110 201 VNR(1)=.005
111 VNR(2)=.1
112 VNR(3)=500
113 VNR(4)=.005
114 VNR(5)=.0 1
115 VNR(6)=500,
116 NS4=1
117 GO TO 203
118 202 VNR(1)=.005
119 VNR(2)=.1
120 VNR(3)=500,
121 VNR(4)=.005
122 VNR(5)=.10
123 VNR(6)=500,
124 NS4=2
125 XPR=0 0
126 DAX=4501P
127 VNR(1)=1
128 GO 204,101,2
129 PRINT 997,(ND1(J,NC5),J=L,0),(ND2(L,ND5),L=L,0),(
130 1 CHE3(NC5,ND5,N5=L,6),(ND4(N,NDP),N=L,7)
131 PRINT 507,(HVAL(K),K=L,20)
132 CALL PLOT(XPL,VPL,NPT,MMU,XPR,VNR,NS4)
133 XPR=XPR
134 XPR=XPR+XPR
135 CONTINUE
136 GO TO 10
137 END

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SYSTEMS 250K FORTRAN IV COMPILER (REV. 0)

SYMBOL USAGE MODE STORAGE LOCATION IN MAIN

DIM	VARIABLE	REAL	PARS	C0007C
DHE	VARIABLE	REAL	PARS	C0007C
CLE	VARIABLE	REAL	PARS	C0007C
DT	VARIABLE	REAL	PARS	C0007C
DIP	VARIABLE	REAL	PARS	C0007C
DIN	VARIABLE	REAL	PARS	C0007C
UN	VARIABLE	REAL	PARS	C0007C
DITED	VARIABLE	REAL	PARS	C0007C
UL	VARIABLE	REAL	PARS	C0007C
EPOT	VARIABLE	REAL	PARS	C0007C
EPT	VARIABLE	REAL	PARS	C0007C
EXIT	VARIABLE	REAL	PARS	C0007C
F	VARIABLE	REAL	PARS	C0007C
Q OT	VARIABLE	REAL	PARS	C0007C
HO	VARIABLE	REAL	PARS	C0007C
HO1	VARIABLE	REAL	PARS	C0007C
HO2	VARIABLE	REAL	PARS	C0007C
HO3	VARIABLE	REAL	PARS	C0007C
HO4	VARIABLE	REAL	PARS	C0007C
LEAD	VARIABLE	REAL	PARS	C0007C
HOALL	VARIABLE	REAL	PARS	C0007C
I	VARIABLE	REAL	PARS	C0007C
INT	VARIABLE	REAL	PARS	C0007C
INT	VARIABLE	REAL	PARS	C0007C
J	VARIABLE	REAL	PARS	C0007C
K	VARIABLE	REAL	PARS	C0007C
L	VARIABLE	REAL	PARS	C0007C
LL	VARIABLE	REAL	PARS	C0007C
LNCONT	VARIABLE	REAL	PARS	C0007C
M	VARIABLE	REAL	PARS	C0007C
MAIN	VARIABLE	REAL	PARS	C0007C
N	VARIABLE	REAL	PARS	C0007C
NCD	VARIABLE	REAL	PARS	C0007C
NCH	VARIABLE	REAL	PARS	C0007C
NCS	VARIABLE	REAL	PARS	C0007C
NOEL	VARIABLE	REAL	PARS	C0007C
NOF	VARIABLE	REAL	PARS	C0007C
NON	VARIABLE	REAL	PARS	C0007C
NPT	VARIABLE	REAL	PARS	C0007C
NSH	VARIABLE	REAL	PARS	C0007C
NUM	VARIABLE	REAL	PARS	C0007C
O	VARIABLE	REAL	PARS	C0007C
OUT	VARIABLE	REAL	PARS	C0007C
PARS	VARIABLE	REAL	PARS	C0007C
PF	VARIABLE	REAL	PARS	C0007C
PFDT	VARIABLE	REAL	PARS	C0007C
PLOT	VARIABLE	REAL	PARS	C0007C
PH	VARIABLE	REAL	PARS	C0007C
PLOT	VARIABLE	REAL	PARS	C0007C
PS	VARIABLE	REAL	PARS	C0007C
PTE	VARIABLE	REAL	PARS	C0007C

## SYSTEMS REAL-TIME MONITOR-40

COMPILER (REV. D)

HSC-100

C-100

C-100

C-100

SYMBOL

MODE

USAGE

STORAGE

LOCATION IN MAIN

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1.3.7. 000103 10410
SYSTEMS PERL-TIME INITIATOR-4.0
SYSTEMS FORTRAN IV COMPILER (REV. 0)
PAGE 9
INPUT
1 SUBROUTINE INPUT
2 COMMON/NCDD/NC1(20,4),NC2(10,2),NC3(2,4,6),NC4(7,6),NCALL(20)
3 COMMON/PA55/PA55(55)
4 DIMENSION NAME(55)
5 DATA NAME/ K1, KR, KP, KW, PTN, ROK, J, PTE,
6 TRUC, TRUP, TRUV, TRV, C, V, B,
7 FR, PS, NCS, NCD, NCY, F, DT, TD, DTP,
8 NG, CH2, DSE, KSE, JHE, DHE, TGF, NOP, ONV,
9 ZET, GSH, CMO, TRUJ, HO, TALH, KEO, KBH, TAO,
10 TRH, TBG, TBH, EI, TRU1, TRU2, RI, TRU3, TRU4,
11 TRU5,
12 TRU6, TRU7, TRU8, TRU9, TRU10, TRU11, TRU12, TRU13, TRU14,
13 TRU15, TRU16, TRU17, TRU18, TRU19, TRU20, TRU21, TRU22, TRU23, TRU24,
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190 TRU1785, TRU1786, TRU1787, TRU1788, TRU1789, TRU1
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SYSTEMS 8536 FORTRAN IV COMPILER (REV. 0)

SYMBOL USAGE NAME STORAGE LOCATION IN INPUT

DO	LABEL	TRANSFER	LOCAL	P0015C
DO	LABEL	TRANSFER	LOCAL	P001B8
DO	LABEL	FORMAT	LOCAL	P00073
DO	LABEL	FORMAT	LOCAL	P00037
DO	LABEL	FORMAT	LOCAL	P00068
DO	VARIBLE	INTEGER FULLWORD	LOCAL	P00041
DO	VARIBLE	INTEGER FULLWORD	LOCAL	P003CC
DO	VARIBLE	INTEGER FULLWORD	LOCAL	P003C9
DO	CONSTANT	INTEGER FULLWORD	LOCAL	P003D4
C 0001	CONSTANT	REAL	LOCAL	P003E8
C 0002	CONSTANT	REAL	LOCAL	P003E4
C 0003	CONSTANT	INTEGER FULLWORD	LOCAL	P003E8
C 0004	CONSTANT	INTEGER FULLWORD	LOCAL	P003EC
C 0005	CONSTANT	INTEGER DOUBLE	LOCAL	P003D3
R IF	PROCEDURE	REAL	EXTERNAL	X0039C
C RI	PROCEDURE	REAL	EXTERNAL	X003F0
EXIT	PROCEDURE	REAL	EXTERNAL	X007C8
DO1	ARRAY	REAL	HEAD	C0000A
DO2	ARRAY	REAL	HEAD	C00148
DO3	ARRAY	REAL	HEAD	C00798
DO4	ARRAY	REAL	HEAD	C00258
DO	COMMON BLOCK	REAL	HEAD	S00248
DO	VARIBLE	INTEGER FULLWORD	LOCAL	C002F8
ID	PROCEDURE	INTEGER FULLWORD	LOCAL	P0004C
INPUT	VARIBLE	INTEGER FULLWORD	LOCAL	P00158
J	VARIBLE	INTEGER FULLWORD	LOCAL	P00059
K	VARIBLE	INTEGER FULLWORD	LOCAL	P00054
L	VARIBLE	INTEGER FULLWORD	LOCAL	P00058
N	VARIBLE	INTEGER FULLWORD	LOCAL	P0005C
N	VARIBLE	INTEGER FULLWORD	LOCAL	P00068
N	VARIBLE	INTEGER FULLWORD	LOCAL	P00064
N	VARIBLE	INTEGER FULLWORD	LOCAL	P00148
N	VARIBLE	INTEGER FULLWORD	LOCAL	P00144
N	VARIBLE	INTEGER FULLWORD	LOCAL	P00148
N	VARIBLE	INTEGER FULLWORD	LOCAL	P0014C
N	VARIBLE	INTEGER FULLWORD	LOCAL	C00008
N	COMMON BLOCK	REAL	HEAD	S0000C
N	PROCEDURE	REAL	EXTERNAL	X0015C
N	PROCEDURE	REAL	EXTERNAL	X00154
N	PROCEDURE	REAL	EXTERNAL	X00198
N	PROCEDURE	REAL	EXTERNAL	X001C8
N	PROCEDURE	REAL	LOCAL	P00158
N	PROCEDURE	REAL	EXTERNAL	X00368









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160 NTRP=OR*(THSDT-THSD)/214EPT
161 V(19)=EPS
162 RETURN
163 C*****
164 C IF NAL INFLU PLUM MODEL - POSITION CONTROL
165 C*****
166 500 EPI=THS-THD
167 V(10)=EPI
168 EPI=0.0
169 EPI=J14EPT-3.0*THSD-3.0*THSDT
170 GO TO 301
171 IF (PI) GT PS1PH=PS
172 IF (PI) LT -PS1PH=-PS
173 TACH=PI*PI
174 THSDT=(THSD-THSDT)/214EPT
175 TEMP=0.0
176 IF (THSDT GT 0.0) TEMP=THSDT-0.001
177 IF (THSDT LT -0.001) TEMP=THSDT+0.001
178 TGP3D=V(19)*BG*TEMP -TGP3D/100
179 VP(13)=TGP3D
180 PH500=(TGP3D+TGP3D+TGP3D+TGP3D)/100
181 VP(20)=PH500
182 THSDC=(THSD+PH500+PH500-THSD)/100
183 VP(18)=THSDC
184 VP(17)=V(18)
185 EPSPR=V(18)*(THS-THSD)+V(18)*(THSDT-THSD)
186 EPSD=V(18)*(THSDT-THSD)+V(18)*(THSDT-THSD)
187 EPSD=(THSD/100)*(THSD+PH500+PH500-THSD)/100
188 VP(19)=EPSD
189 GO TO (101,102,103,104),ICS
190 TGP3D=10.0*V(18)*(THSD-THSD)-2.0*ZET+V(18)*TGP3D-10.0*V(18)*TGP3D
191 VP(13)=TGP3D
192 PH500=(TGP3D+TGP3D+TGP3D+TGP3D)/100
193 TH500=(THSD+PH500+PH500-THSD)/100
194 VP(20)=PH500
195 VP(18)=TH500
196 VP(17)=V(18)
197 EPS=V(18)*(THS-THSD)+V(18)*(THSDT-THSD)
198 GO TO 301
199 PHD=THSD
200 PHD=0.0
201 IF (PHD GT 0.0) PHD=PHD-0.001
202 IF (PHD LT -0.001) PHD=PHD+0.001
203 TH500=(PHD-THSD)/100
204 VP(21)=TH500
205 PHD=TH500
206 PHD=0.0
207 IF (PHD GT 0.0) PHD=PHD-0.001
208 IF (PHD LT -0.001) PHD=PHD+0.001
209 PH500=(PHD+PHD+PHD+PHD)/100
210 VP(20)=PH500
211 TH500=(THSD+PH500+PH500-THSD)/100
212 VP(18)=TH500

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SCIENTIFIC REAL-TIME MONITOR-4.0

MONITOR

09.01.03

12/19/74

SYSTEMS 83/86 FORTRAN IV COMPILER (REV 0)

DIEO

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213 VP(17)=V(18)
214 LPS=V(18)*THS-THGS)*X8*(THSDT-THGSD)-X8H*THGSD
215 GO TO 203
216 STOP C.M
217
218 250 TOSPO=MIN(MIN*THGD-2 *ZET+MIN*TOGPO-MIN*THGSD)
219 THSP=MIN(MIN*THGD-2 *ZET+MIN*THGSD-MIN*THGSD)
220 THSD=THSPD
221 THGSD=0
222 IF (THSD GT 180) THSD=THSPD-D8H
223 IF (THSD LT (-180)) THSD=THSPD+D8H
224 THGSD=0
225 IF (1/THSD GT D8G) THGSD=TOGSD-D8G
226 IF (TOGSD LT (-D8G)) THGSD=TOGSD+D8G
227 V(18)=THSD
228 THSDT=(THSD+THGSD+THSS )/THUI
229 TOSG=THG-THH+THSS
230 LPS=X1*(THS-THGS)-XCR*THGSD-XCR*THGSD
231 THSDT=0
232 THGSD=(THUI*TOGSD+TOGSD-THGSD)/THUI
233 VP(18)=THGSD
234 VP(12)=TOGSD
235 VP(13)=TOGSD
236 VP(14)=THSPD
237 VP(15)=THSPD
238 VP(16)=THGSD
239 VP(17)=THGSD
240 GO TO (101,202,307,501),NCS
241 EPS=X1*(THS-THGS)+XCR*(THSDT-THGSD)-XCR*THGSD
242 GO TO 201
243 EPS=X1*(THS-THGS)+XCR*(THSDT-THGSD)-XCR*THGSD
244 GO TO 303
245 EPT=THS-THG
246 EPT=0
247 GO TO 381
248 END

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02040

SYSTEMS 8306 FORTAN IV COMPILER (REV. 0)

SYMBOL USAGE NAME STORAGE LOCATION IN DIFED

110	LABEL	TRANSFER	LOCAL	P00480
1100	LABEL	TRANSFER	LOCAL	P00480
1101	LABEL	TRANSFER	LOCAL	P00480
111	LABEL	TRANSFER	LOCAL	P00484
112	LABEL	TRANSFER	LOCAL	P00488
1123	LABEL	TRANSFER	LOCAL	P00614
1124	LABEL	TRANSFER	LOCAL	P00618
113	LABEL	TRANSFER	LOCAL	P00618
114	LABEL	TRANSFER	LOCAL	P00618
1150	LABEL	TRANSFER	LOCAL	P00664
118	LABEL	TRANSFER	LOCAL	P00664
1208	LABEL	TRANSFER	LOCAL	P00700
1209	LABEL	TRANSFER	LOCAL	P00700
1201	LABEL	TRANSFER	LOCAL	P00700
1202	LABEL	TRANSFER	LOCAL	P00700
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128	LABEL	TRANSFER	LOCAL	P00700
129	LABEL	TRANSFER	LOCAL	P00700
1301	LABEL	TRANSFER	LOCAL	P00700
1303	LABEL	TRANSFER	LOCAL	P00700
1307	LABEL	TRANSFER	LOCAL	P00700
1321	LABEL	TRANSFER	LOCAL	P00700
1322	LABEL	TRANSFER	LOCAL	P00700
1350	LABEL	TRANSFER	LOCAL	P00700
1351	LABEL	TRANSFER	LOCAL	P00700
1400	LABEL	TRANSFER	LOCAL	P00700
1458	LABEL	TRANSFER	LOCAL	P00700
150	LABEL	TRANSFER	LOCAL	P00700
1508	LABEL	TRANSFER	LOCAL	P00700
1501	LABEL	TRANSFER	LOCAL	P00700
1530	LABEL	TRANSFER	LOCAL	P00700
1550	LABEL	TRANSFER	LOCAL	P00700
1558	LABEL	TRANSFER	LOCAL	P00700
1630	VARIABLE	INTEGER FULLWORD	LOCAL	P01164
1631	VARIABLE	INTEGER FULLWORD	LOCAL	P01164
1632	VARIABLE	INTEGER FULLWORD	LOCAL	P01164
1633	VARIABLE	INTEGER FULLWORD	LOCAL	P01164
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1679	VARIABLE	INTEGER FULLWORD	LOCAL	P01164
1680	VARIABLE	INTEGER FULLWORD	LOCAL	P01164
1681	VARIABLE	INTEGER FULLWORD	LOCAL	P01164

SYMBOL	USAGE	MODE	STATUS	LOCATION IN DIFER
DBJ	VARIABLE	INTEGER FULLWORD	LOCAL	P013E4
DBA	VARIABLE	INTEGER FULLWORD	LOCAL	P017E8
DBA	VARIABLE	INTEGER FULLWORD	LOCAL	P017E9
DBA	VARIABLE	INTEGER FULLWORD	LOCAL	P017F0
DBA	VARIABLE	INTEGER FULLWORD	LOCAL	P017F4
DBA	VARIABLE	INTEGER FULLWORD	LOCAL	P017F8
DBA	VARIABLE	INTEGER FULLWORD	LOCAL	P013FC
DBA	VARIABLE	INTEGER FULLWORD	LOCAL	P01400
DBA	VARIABLE	INTEGER FULLWORD	LOCAL	P01404
DBA	VARIABLE	INTEGER FULLWORD	LOCAL	P01409
DBA	CONSTANT	REAL	LOCAL	P01413
DBA	CONSTANT	REAL	LOCAL	P0141C
DBA	CONSTANT	REAL	LOCAL	P01420
DBA	CONSTANT	REAL	LOCAL	P01424
DBA	CONSTANT	REAL	LOCAL	P01428
DBA	CONSTANT	REAL	LOCAL	P0142C
DBA	CONSTANT	REAL	LOCAL	P01434
DBA	CONSTANT	REAL	LOCAL	P01438
DBA	CONSTANT	REAL	LOCAL	P0147C
DBA	CONSTANT	REAL	LOCAL	P01440
DBA	CONSTANT	REAL	LOCAL	P01444
DBA	CONSTANT	REAL	LOCAL	P01448
DBA	VARIABLE	INTEGER DOUBLE	PARS	P00008
DBA	VARIABLE	REAL	LOCAL	P00009
DBA	VARIABLE	REAL	LOCAL	P0000A
DBA	VARIABLE	REAL	PARS	P00030
DBA	VARIABLE	REAL	PARS	P00040
DBA	VARIABLE	REAL	PARS	P00038
DBA	COMMON BLOCK	REAL	PARS	P0006C
DBA	VARIABLE	REAL	PARS	P00094
DBA	VARIABLE	REAL	PARS	P00099
DBA	VARIABLE	REAL	PARS	P0009C
DBA	VARIABLE	REAL	PARS	P0007C
DBA	VARIABLE	REAL	PARS	P00070
DBA	VARIABLE	REAL	PARS	P0005C
DBA	VARIABLE	REAL	PARS	P00064
DBA	PROCEDURE	REAL	ENTRY	P0009C
DBA	VARIABLE	REAL	PARS	P0008C
DBA	VARIABLE	REAL	LOCAL	P00009
DBA	VARIABLE	REAL	LOCAL	P0000C
DBA	VARIABLE	REAL	OPUT	P00009
DBA	VARIABLE	REAL	LOCAL	P00010
DBA	PROCEDURE	REAL	EXTERNAL	X01394
DBA	VARIABLE	REAL	LOCAL	P00014
DBA	VARIABLE	REAL	LOCAL	P00018
DBA	VARIABLE	REAL	LOCAL	P0001C
DBA	VARIABLE	REAL	PARS	P00059
DBA	PROCEDURE	REAL	EXTERNAL	X000E0
DBA	PROCEDURE	REAL	EXTERNAL	X012D8
DBA	VARIABLE	REAL	PARS	P0009C
DBA	VARIABLE	INTEGER FULLWORD	PARS	P00004





SYSTEMS REL-TIME JITTER-40

SYSTEMS REL-TIME JITTER-40

SYSTEMS REL-TIME JITTER-40

SYSTEMS REL-TIME JITTER-40

SYSTEMS REL-TIME JITTER-40

SYSTEMS REL-TIME JITTER-40

SYSTEMS REL-TIME JITTER-40

SYSTEMS REL-TIME JITTER-40



[illegible]

SYSTEMS 10/3/6 FORTRAN IV COMPILER (DEV G)

LOCATION IN PLOT

SOURCE	NAME	TYPE	LOCATION IN PLOT
10000	TRANSFER	TRANSFER	10000
10001	TRANSFER	TRANSFER	10000
10002	TRANSFER	TRANSFER	10000
10003	TRANSFER	TRANSFER	10000
10004	TRANSFER	TRANSFER	10000
10005	TRANSFER	TRANSFER	10000
10006	TRANSFER	TRANSFER	10000
10007	TRANSFER	TRANSFER	10000
10008	TRANSFER	TRANSFER	10000
10009	TRANSFER	TRANSFER	10000
10010	TRANSFER	TRANSFER	10000
10011	TRANSFER	TRANSFER	10000
10012	TRANSFER	TRANSFER	10000
10013	TRANSFER	TRANSFER	10000
10014	TRANSFER	TRANSFER	10000
10015	TRANSFER	TRANSFER	10000
10016	TRANSFER	TRANSFER	10000
10017	TRANSFER	TRANSFER	10000
10018	TRANSFER	TRANSFER	10000
10019	TRANSFER	TRANSFER	10000
10020	TRANSFER	TRANSFER	10000
10021	TRANSFER	TRANSFER	10000
10022	TRANSFER	TRANSFER	10000
10023	TRANSFER	TRANSFER	10000
10024	TRANSFER	TRANSFER	10000
10025	TRANSFER	TRANSFER	10000
10026	TRANSFER	TRANSFER	10000
10027	TRANSFER	TRANSFER	10000
10028	TRANSFER	TRANSFER	10000
10029	TRANSFER	TRANSFER	10000
10030	TRANSFER	TRANSFER	10000
10031	TRANSFER	TRANSFER	10000
10032	TRANSFER	TRANSFER	10000
10033	TRANSFER	TRANSFER	10000
10034	TRANSFER	TRANSFER	10000
10035	TRANSFER	TRANSFER	10000
10036	TRANSFER	TRANSFER	10000
10037	TRANSFER	TRANSFER	10000
10038	TRANSFER	TRANSFER	10000
10039	TRANSFER	TRANSFER	10000
10040	TRANSFER	TRANSFER	10000
10041	TRANSFER	TRANSFER	10000
10042	TRANSFER	TRANSFER	10000
10043	TRANSFER	TRANSFER	10000
10044	TRANSFER	TRANSFER	10000
10045	TRANSFER	TRANSFER	10000
10046	TRANSFER	TRANSFER	10000
10047	TRANSFER	TRANSFER	10000
10048	TRANSFER	TRANSFER	10000
10049	TRANSFER	TRANSFER	10000
10050	TRANSFER	TRANSFER	10000
10051	TRANSFER	TRANSFER	10000
10052	TRANSFER	TRANSFER	10000
10053	TRANSFER	TRANSFER	10000
10054	TRANSFER	TRANSFER	10000
10055	TRANSFER	TRANSFER	10000
10056	TRANSFER	TRANSFER	10000
10057	TRANSFER	TRANSFER	10000
10058	TRANSFER	TRANSFER	10000
10059	TRANSFER	TRANSFER	10000
10060	TRANSFER	TRANSFER	10000
10061	TRANSFER	TRANSFER	10000
10062	TRANSFER	TRANSFER	10000
10063	TRANSFER	TRANSFER	10000
10064	TRANSFER	TRANSFER	10000
10065	TRANSFER	TRANSFER	10000
10066	TRANSFER	TRANSFER	10000
10067	TRANSFER	TRANSFER	10000
10068	TRANSFER	TRANSFER	10000
10069	TRANSFER	TRANSFER	10000
10070	TRANSFER	TRANSFER	10000
10071	TRANSFER	TRANSFER	10000
10072	TRANSFER	TRANSFER	10000
10073	TRANSFER	TRANSFER	10000
10074	TRANSFER	TRANSFER	10000
10075	TRANSFER	TRANSFER	10000
10076	TRANSFER	TRANSFER	10000
10077	TRANSFER	TRANSFER	10000
10078	TRANSFER	TRANSFER	10000
10079	TRANSFER	TRANSFER	10000
10080	TRANSFER	TRANSFER	10000
10081	TRANSFER	TRANSFER	10000
10082	TRANSFER	TRANSFER	10000
10083	TRANSFER	TRANSFER	10000
10084	TRANSFER	TRANSFER	10000
10085	TRANSFER	TRANSFER	10000
10086	TRANSFER	TRANSFER	10000
10087	TRANSFER	TRANSFER	10000
10088	TRANSFER	TRANSFER	10000
10089	TRANSFER	TRANSFER	10000
10090	TRANSFER	TRANSFER	10000
10091	TRANSFER	TRANSFER	10000
10092	TRANSFER	TRANSFER	10000
10093	TRANSFER	TRANSFER	10000
10094	TRANSFER	TRANSFER	10000
10095	TRANSFER	TRANSFER	10000
10096	TRANSFER	TRANSFER	10000
10097	TRANSFER	TRANSFER	10000
10098	TRANSFER	TRANSFER	10000
10099	TRANSFER	TRANSFER	10000
10100	TRANSFER	TRANSFER	10000



10/19/74 00 01 03 110010 SYSTEMS PUL-TIME 1-011010-4 0  
SYSTEMS 8006 FORTRAN IV COMPILER (REV 0)

NAME

```

1  SUBROUTINE ZONE(Y,NP,YMR,YMR(NC))
2  DIMENSION YMR(NP),YMR(NC),YMR(NC)
3  GO TO 1,1,NC
4  YMR=0
5  DO 20 J=1,NP
6  YMR=ABS(YMR(J))
7  IF(YMR.GT.YMR)YMR=YMR
8  CONTINUE
9  YMR=0
10 INT=YMR
11 IF(INT.NE.0)GO TO 40
12 YMR=YMR*10
13 YMR=YMR*10
14 GO TO 20
15 IF(YMR-INT).GT.1E-5)INT=INT+1
16 IF(INT.NE.1)INT=10
17 YMR=YMR-INT*10
18 YMR=YMR*10
19 CONTINUE
20 RETURN
21 END

```

12/19/74 09:01:23 M00016  
 SYSTENS 3586 FURTAN IV COMPILER (REV. G)  
 STATUS PNL-TIME MONITOR-40

SYMBOL	USAGE	MODE	STORAGE	LOCATION IN SCALE
00	LABEL	TRANSFER	IMMEDIATE	100000
01	LABEL	TRANSFER	IMMEDIATE	100000
02	LABEL	TRANSFER	LOCAL	100000
03	LABEL	TRANSFER	LOCAL	100000
04	VARIBLE	INTEGER FULLWORD	LOCAL	100000
05	VARIBLE	INTEGER FULLWORD	LOCAL	100000
06	VARIBLE	INTEGER FULLWORD	LOCAL	100000
07	VARIBLE	INTEGER FULLWORD	LOCAL	100000
08	CONSTANT	REAL	LOCAL	100000
09	CONSTANT	REAL	LOCAL	100000
10	PROCEDURE	REAL	EXTERNAL	100000
11	PROCEDURE	INTEGER FULLWORD	EXTERNAL	100000
12	PROCEDURE	REAL	EXTERNAL	100000
13	PROCEDURE	REAL	EXTERNAL	100000
14	VARIBLE	INTEGER FULLWORD	LOCAL	100000
15	VARIBLE	INTEGER FULLWORD	LOCAL	100000
16	VARIBLE	INTEGER FULLWORD	LOCAL	100000
17	VARIBLE	INTEGER FULLWORD	LOCAL	100000
18	VARIBLE	INTEGER FULLWORD	LOCAL	100000
19	VARIBLE	INTEGER FULLWORD	LOCAL	100000
20	VARIBLE	INTEGER FULLWORD	LOCAL	100000
21	VARIBLE	INTEGER FULLWORD	LOCAL	100000
22	VARIBLE	INTEGER FULLWORD	LOCAL	100000
23	VARIBLE	INTEGER FULLWORD	LOCAL	100000
24	VARIBLE	INTEGER FULLWORD	LOCAL	100000
25	VARIBLE	INTEGER FULLWORD	LOCAL	100000
26	VARIBLE	INTEGER FULLWORD	LOCAL	100000
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28	VARIBLE	INTEGER FULLWORD	LOCAL	100000
29	VARIBLE	INTEGER FULLWORD	LOCAL	100000
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37	VARIBLE	INTEGER FULLWORD	LOCAL	100000
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89	VARIBLE	INTEGER FULLWORD	LOCAL	100000
90	VARIBLE	INTEGER FULLWORD	LOCAL	100000
91	VARIBLE	INTEGER FULLWORD	LOCAL	100000
92	VARIBLE	INTEGER FULLWORD	LOCAL	100000
93	VARIBLE	INTEGER FULLWORD	LOCAL	100000
94	VARIBLE	INTEGER FULLWORD	LOCAL	100000
95	VARIBLE	INTEGER FULLWORD	LOCAL	100000
96	VARIBLE	INTEGER FULLWORD	LOCAL	100000
97	VARIBLE	INTEGER FULLWORD	LOCAL	100000
98	VARIBLE	INTEGER FULLWORD	LOCAL	100000
99	VARIBLE	INTEGER FULLWORD	LOCAL	100000
100	VARIBLE	INTEGER FULLWORD	LOCAL	100000

```

SYSTENS 0.5.96 FORTPHN IV COMPILER (REV Q)
RK
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35
SUBROUTINE F4(N,K,VO,DV,XD,XV,VP,XM,AL,DIFEC)
C ***** N SPECIFIES NUMBER OF EQNS. K=1 PROVIDES VP'S AT END POINTS
C DIMENSION YOUR(DV,N),VX(N),VP(N),AL(N,4)
DOUBLE PRECISION XO
XO = XD5
I = 1
0 DO 10 (1,2,3,4,5), I
1 XM = XO
2 DO 6 J = 1,N
3 VXC(J) = VXC(J)
4 CALL DIFEC(VX,VP,XM,N)
5 DO 7 J = 1,N
6 ATC(J,1) = DV*VP(J)
7 I = I + 1
8 GO TO 8
9 XM = XO + 0.5*DX
10 DO 9 J = 1,N
11 VXC(J) = VXC(J) + 0.5*ATC(J,1)
12 GO TO 10
13 DO 11 J = 1,N
14 VXC(J) = VXC(J) + 0.5*ATC(J,2)
15 GO TO 10
16 XM = XO + DX
17 DO 12 J = 1,N
18 VXC(J) = VXC(J) + ATC(J,3)
19 GO TO 10
20 DO 13 J = 1,N
21 DV(J) = (0.5*ATC(J,1)+ATC(J,4))+ATC(J,2)+ATC(J,3))/2.0
22 VXC(J) = VXC(J) + DV(J)
23 IF(K.NE.1) RETURN
24 DO 16 J = 1,N
25 VXC(J) = VXC(J)
26 CALL DIFEC(VX,VP,XM,N)
27 RETURN
28 END

```

## SYSTEMS 3.5.8.6 FORTRAN IV COMPILER (REV G)

LINE	USAGE	MODE	STORAGE	LOCATION IN PK
>1	LABEL	TRANSFER	LOCAL	P00090
>10	LABEL	TRANSFER	LOCAL	P00098
>11	LABEL	TRANSFER	INACTIVE	P00100
>12	LABEL	TRANSFER	INACTIVE	P00100
>13	LABEL	TRANSFER	INACTIVE	P00100
>16	LABEL	TRANSFER	INACTIVE	P00100
>2	LABEL	TRANSFER	LOCAL	P00140
>3	LABEL	TRANSFER	LOCAL	P001C8
>4	LABEL	TRANSFER	LOCAL	P00258
>5	LABEL	TRANSFER	LOCAL	P00258
>6	LABEL	TRANSFER	INACTIVE	P00300
>7	LABEL	TRANSFER	INACTIVE	P00300
>8	LABEL	TRANSFER	INACTIVE	P00300
>9	LABEL	TRANSFER	INACTIVE	P00300
>A0	VARIBLE	INTEGER FULLWORD	LOCAL	P00300
>A1	VARIBLE	INTEGER FULLWORD	LOCAL	P00400
>A2	VARIBLE	INTEGER FULLWORD	LOCAL	P00400
>A3	VARIBLE	INTEGER FULLWORD	LOCAL	P00410
>A4	VARIBLE	INTEGER FULLWORD	LOCAL	P00414
>A5	VARIBLE	INTEGER FULLWORD	LOCAL	P00418
>A6	VARIBLE	INTEGER FULLWORD	LOCAL	P0041C
>A7	VARIBLE	INTEGER FULLWORD	LOCAL	P00420
>A8	VARIBLE	INTEGER FULLWORD	LOCAL	P00424
>A9	VARIBLE	INTEGER FULLWORD	LOCAL	P0042C
>B0	VARIBLE	INTEGER FULLWORD	LOCAL	P00430
>B1	CONSTANT	REAL	LOCAL	P00434
C.0001	CONSTANT	REAL	LOCAL	P00438
C.0002	CONSTANT	REAL	LOCAL	P0043C
A1	ARRAY	REAL	DUMMY	P00044
DX	VARIBLE	REAL	DUMMY	P00034
DY	VARIBLE	REAL	DUMMY	P00034
DIFEC	PROCEDURE	REAL	DUMMY	P0003C
EXIT	PROCEDURE	REAL	DUMMY	P00048
F.PR	PROCEDURE	REAL	EXTERNAL	X00404
G.OT	PROCEDURE	REAL	EXTERNAL	X00418
I	VARIBLE	REAL	EXTERNAL	X00464
J	VARIBLE	INTEGER FULLWORD	LOCAL	P00038
K	VARIBLE	INTEGER FULLWORD	LOCAL	P0003C
N	VARIBLE	INTEGER FULLWORD	DUMMY	P00024
RET.	VARIBLE	INTEGER FULLWORD	DUMMY	P00028
RK	PROCEDURE	REAL	DUMMY	P00010
X0	VARIBLE	DOUBLE	ENTRY	P00014
X0S	VARIBLE	REAL	LOCAL	P00000
X1	VARIBLE	REAL	DUMMY	P00030
Y0	ARRAY	REAL	DUMMY	P00040
VP	ARRAY	REAL	DUMMY	P00020
WM	ARRAY	REAL	DUMMY	P0003C
			DUMMY	P00038

EXTERNAL DIFED  
COMMON/TABLES/THMOL ( 500) ,TCTOL ( 500) ,MOEL,TOLD,MOM  
COMMON/OPUT/EP5,QH,XTMP,THIDT,THGS

COMMON/PARS/XKI,XKA,XKP,XKH,PTN,ROM,XJ,PTE,TAUC,TAUP,TAUV,OGH,  
AVMX,XL,C,V,B,XKA,PS,XNCS,XNCO,XNCN,F,DT,TNO,OTP,XKB  
.CH2,OSE,XKSE,XJHE,OME,TGF  
.XNOP,WN,ZET,OBH,OBG,TAUJ,MO,TALM  
.XK8G,XK8H,TAG,TAH,T8G,T8H  
.E1,TAU1,TAU2,A1,TAU3,TAU4,TD,TAUS

COMMON/CASE/NCS,NCO,NCN,NOP  
DIMENSION XPL (350),YPL (3,350),YMX (3),YMN (3)  
DIMENSION Y (21),YL (21),YP (21),Z (21),O (21),OY (21),YH (21),AI (4,21)  
COMMON/HEAD/HO1 (20,4),HO2 (10,2),HO3 (2,4,6),HO4 (7,6),HOALL (20)

EQUIVALENCE (Y (1),THG) , (YP (1),THGOT) , (Y (8),AV ) ,  
(Y (2),THGO) , (YP (2),THGOT) , (YP (8),AVOT) ,  
(Y (3),TA ) , (YP (3),TAOT ) ,  
(Y (4),PF ) , (YP (4),PFOT ) ,  
(Y (5),PH ) , (YP (5),PHOT ) ,  
(Y (6),EPT ) , (YP (6),EPOT ) ,  
(Y (7),THS ) , (YP (7),THSOT ) ,  
, (Y (9),THH) , (YP (9),THHOT) , (Y (10),THHG) , (YP (10),THHGO) ,

(Y (11),THMGO) , (YP (11),THMGOT)  
, (YP (12),TCSPOT) , (Y (12),TCSPOT)  
, (YP (13),TCSP30) , (Y (13),TCSP30)  
, (YP (14),THSPOT) , (Y (14),THSPOT)  
, (YP (15),THSP30) , (Y (15),THSP30)  
, (YP (16),THSSOT) , (Y (16),THSS)

READ 995, (MDALL (K),K=1,20)

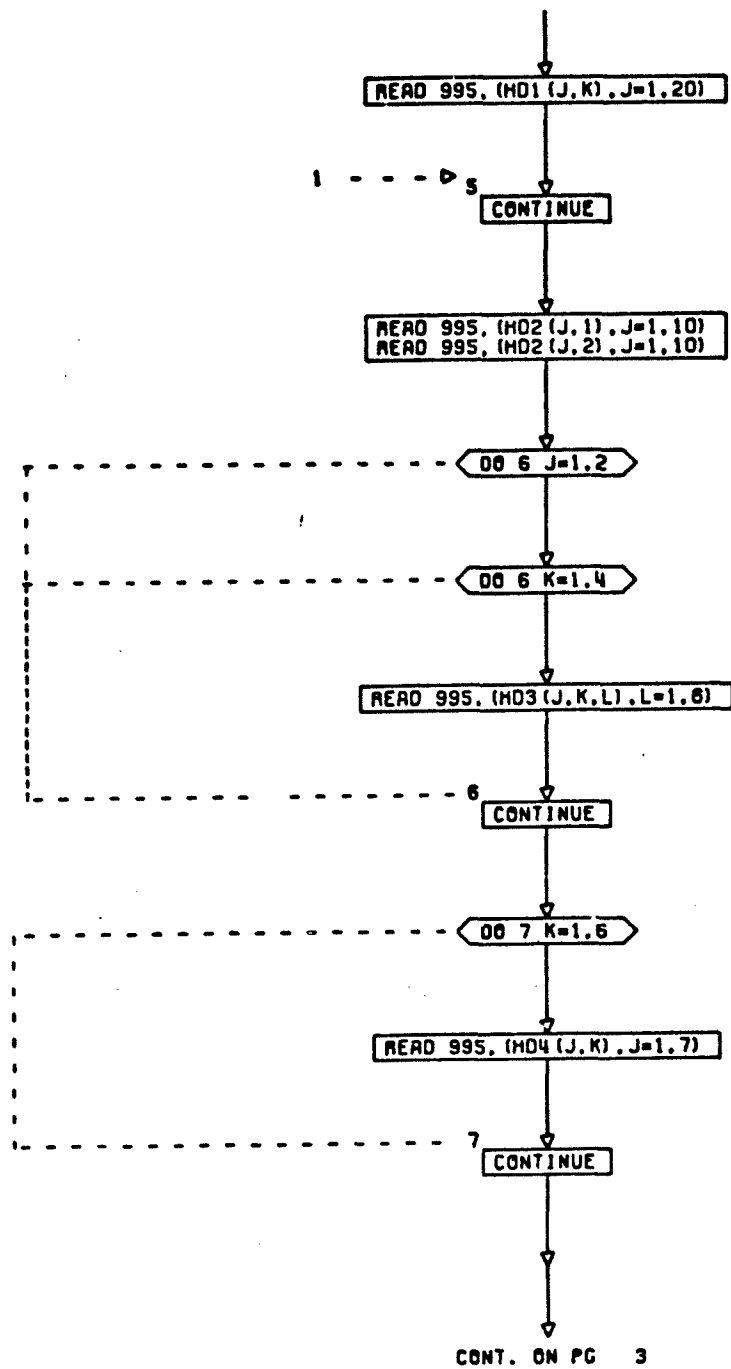
DO 5 K=1,4 - - - - > 2

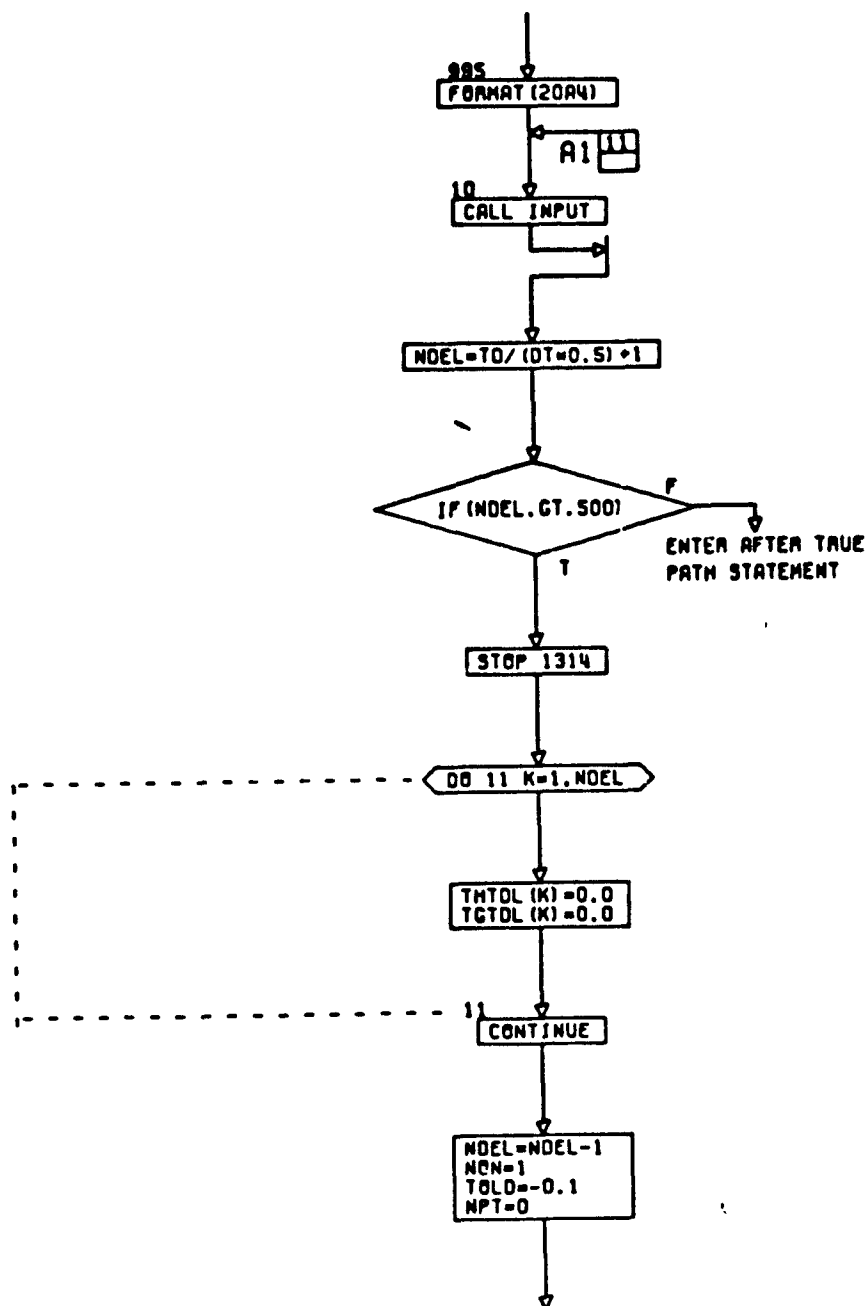
CONT. ON PG 2

PG 1 OF 11

A-63



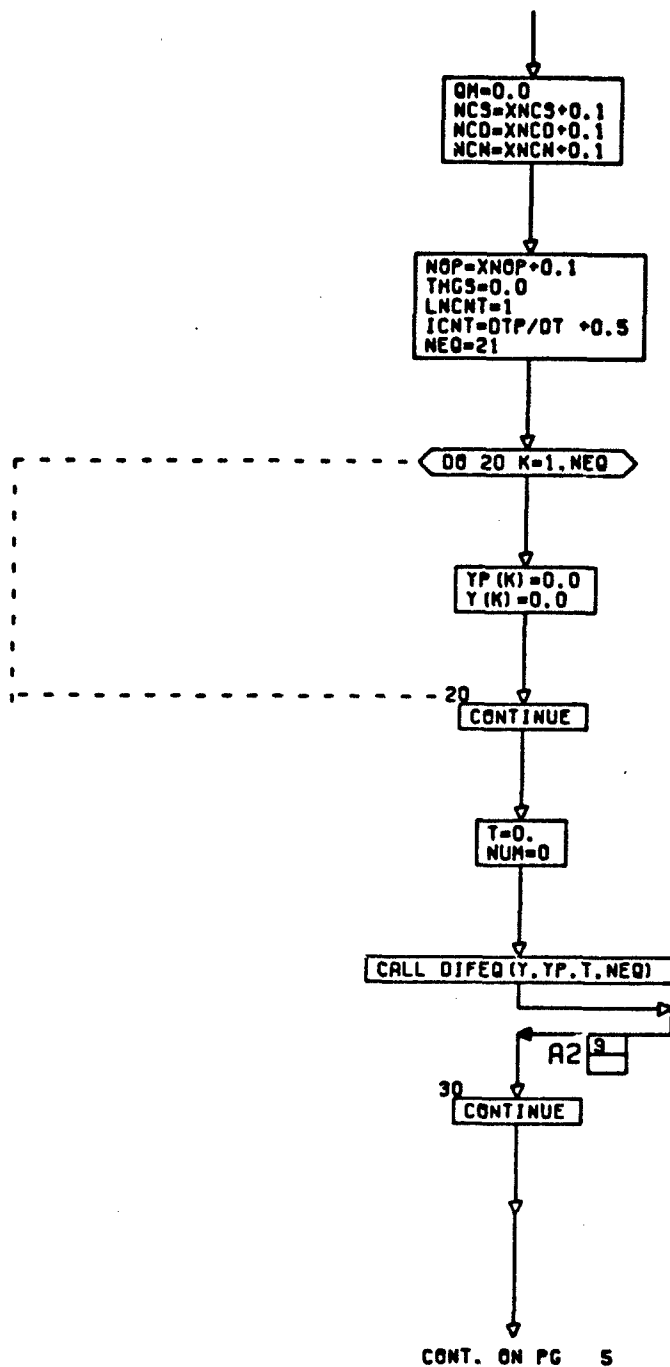


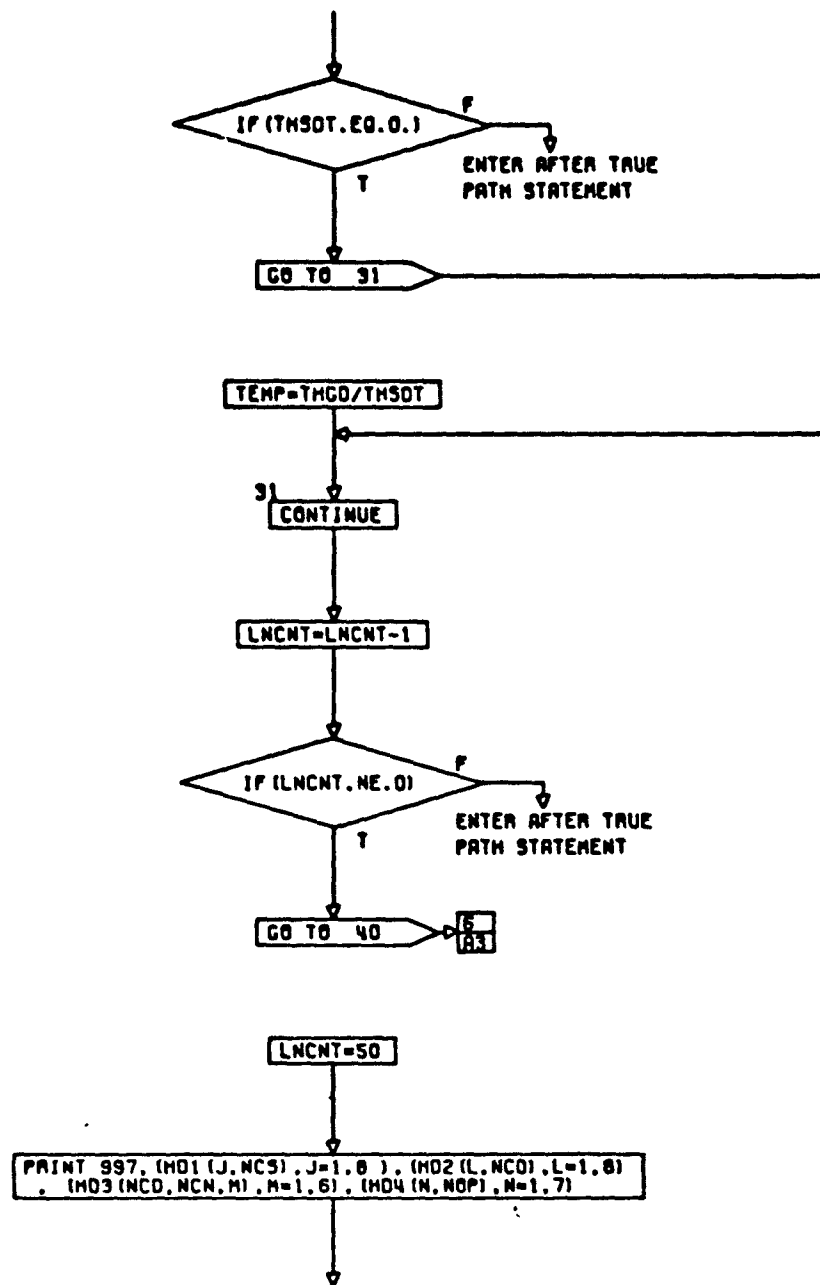


CONT. ON PG 4

PG 3 OF 11

A-65

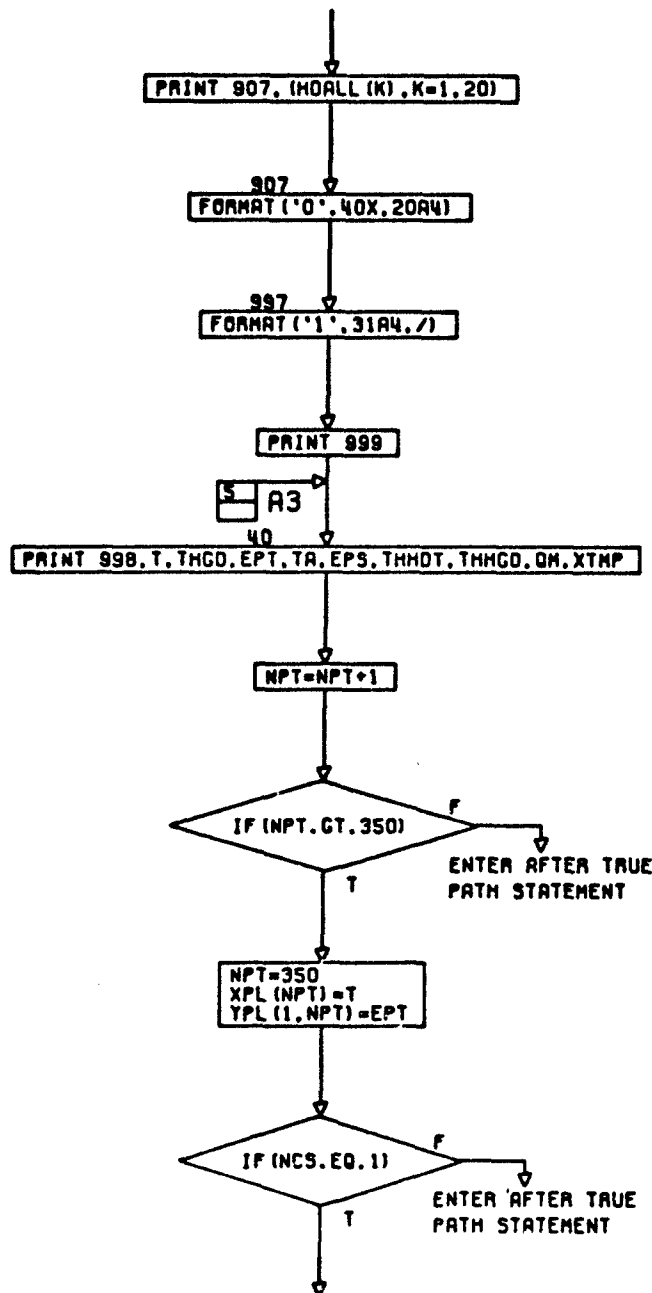




CONT. ON PG 8

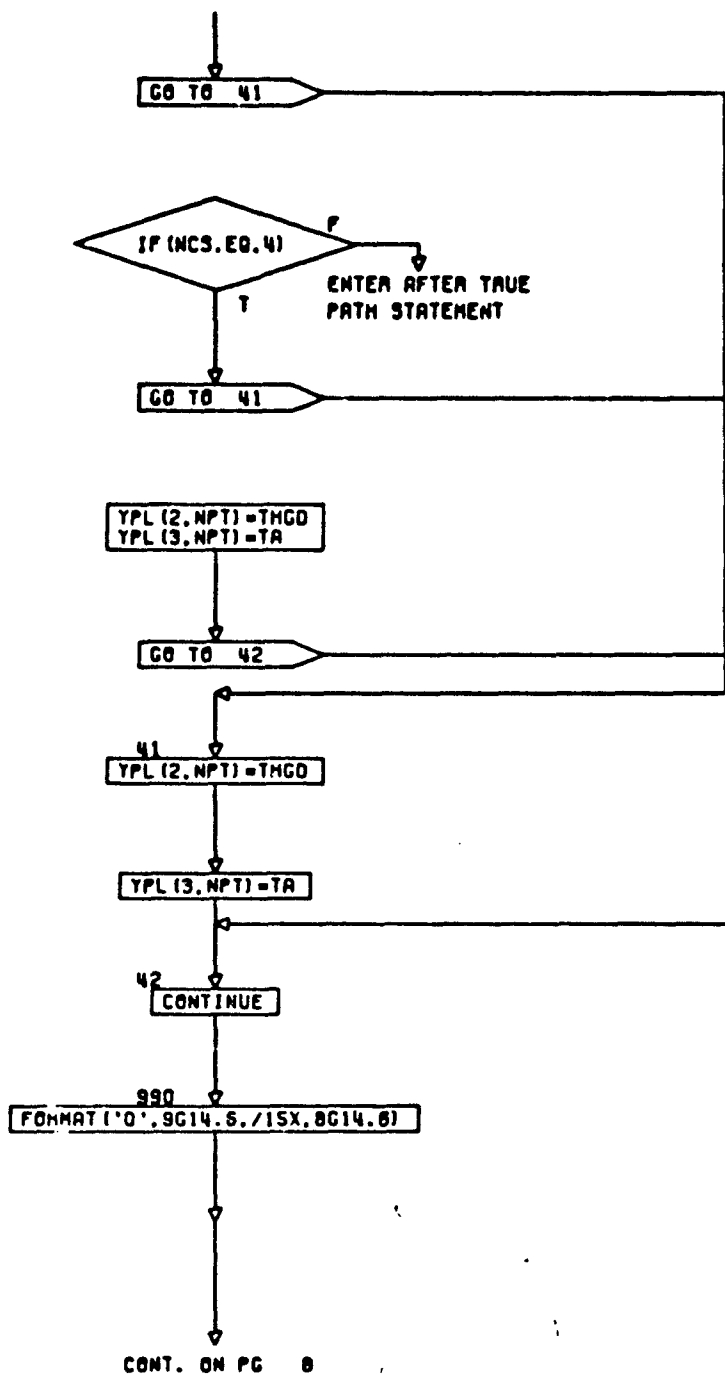
PG 5 OF 11

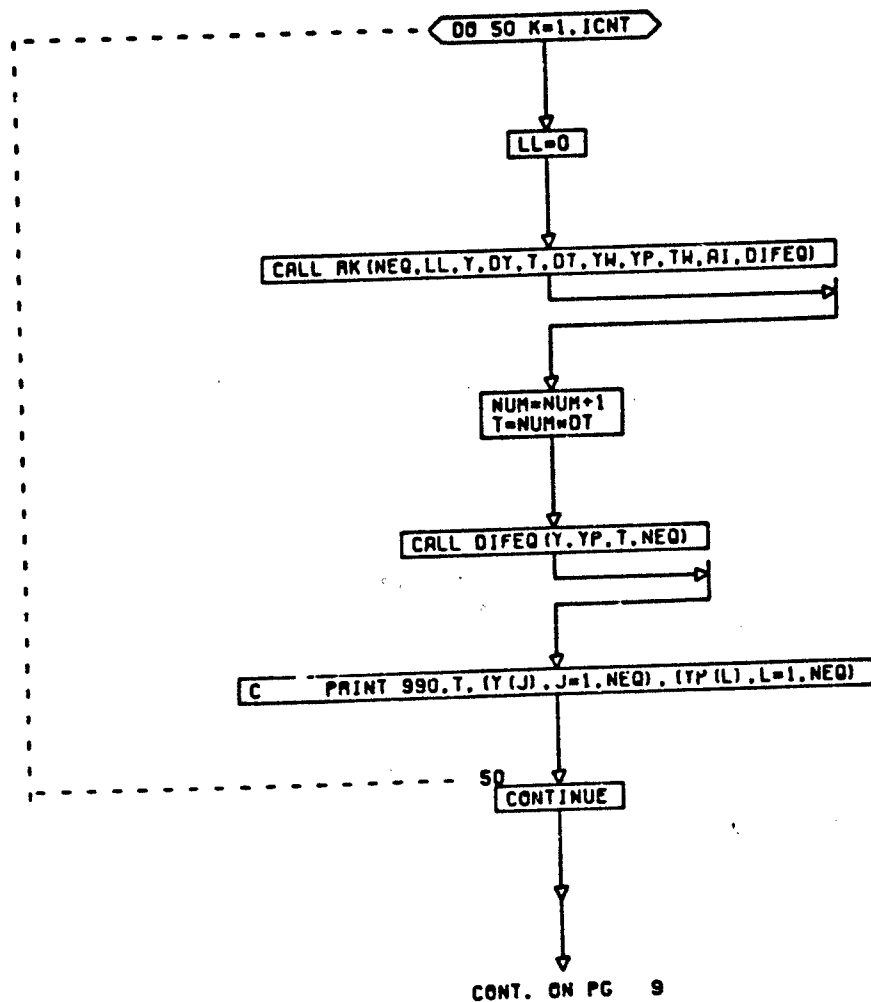
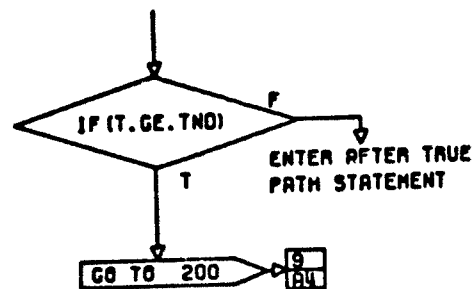
A-67

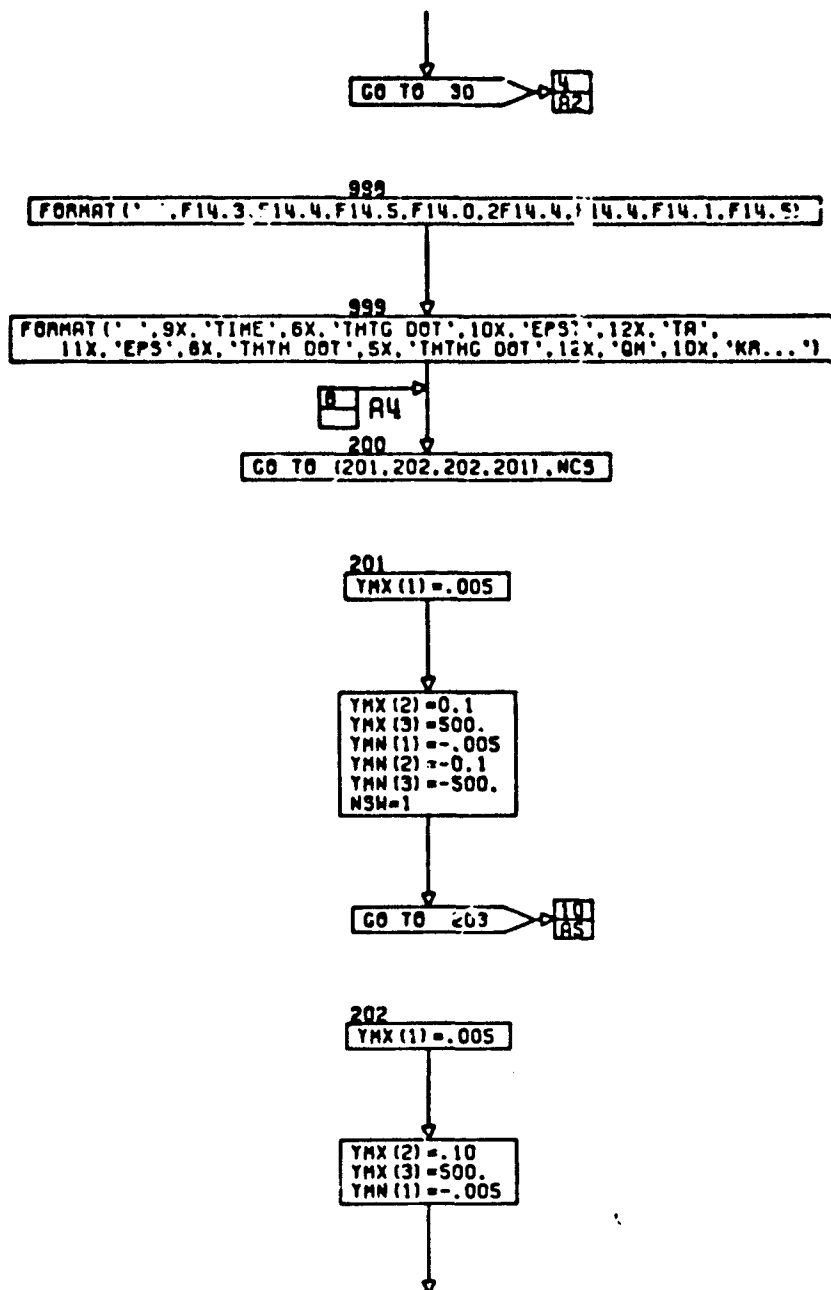


CONT. ON PG 7

PG 6 OF 11



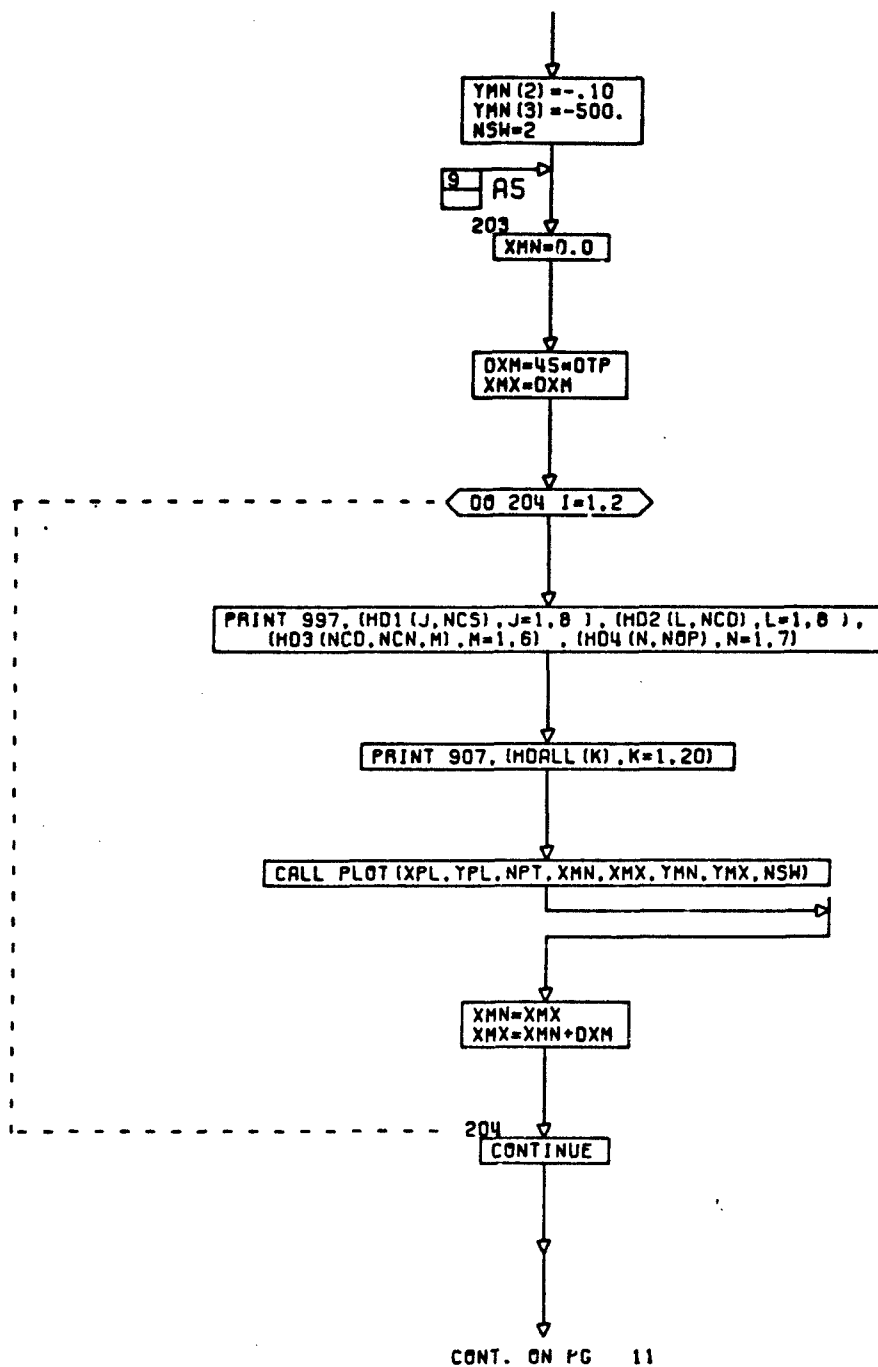


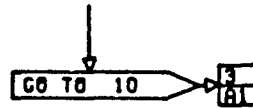


CONT. ON PG 10

PG 9 OF 11







END

PG 11 FINAL

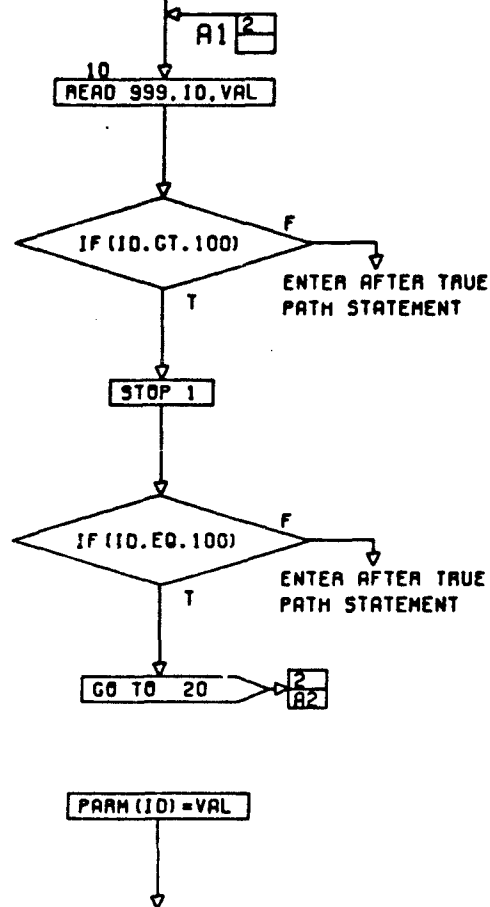
A-73

```

SUBROUTINE INPUT
COMMON/HEAD/HD1 (20,4),HD2 (10,2),HD3 (2,4,6),HD4 (7,6),HDALL (20)
COMMON/PARS/PA1 (55)
DIMENSION NAME (55)

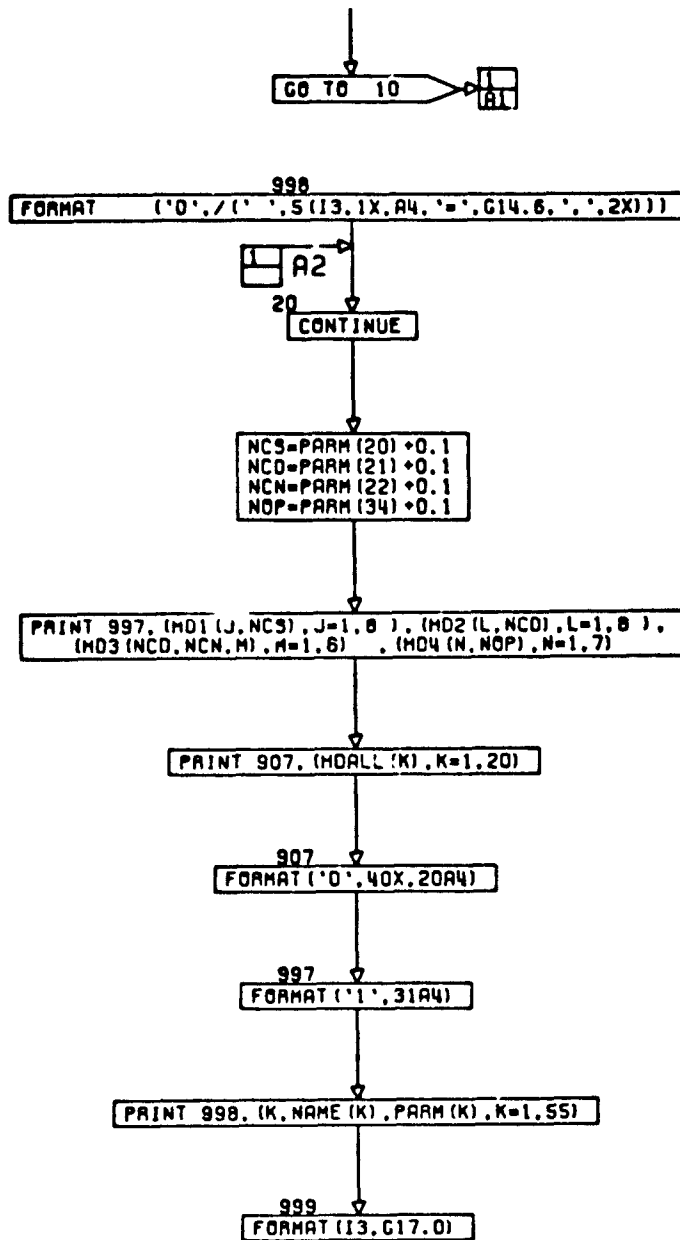
```

DATA NAME	KI	KR	KP	KH	PTN	ADM	J	PTE
TAUC	TAUP	TAUV	OGH	AVMX	L	C	V	B
KA	PS	NCS	NCO	NCN	F	OT	TND	DTP
KB	CH2	DSE	KSE	JME	OME	TCF	NOP	OMN
ZET	DBH	DBG	TAU1	HD	TALM	KBG	KBH	TAC
TAM	T8C	T8H	E1	TAU1	TAU2	A1	TAU3	TAU4
TD	TAUS							



CONT. ON PG 2

PG 1 OF 3



CONT. ON PG 3

PG 2 OF 3

A-75

↓  
RETURN

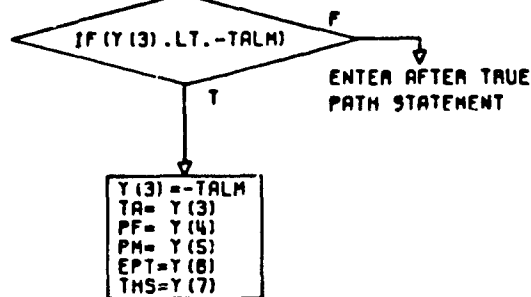
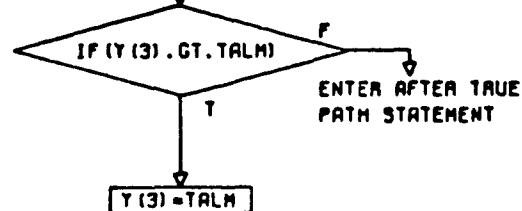
END

PG 3 FINAL

SUBROUTINE DIFEQ(Y,YP,T,NEQ)  
 COMMON/TABLES/TMTOL( 500),TCTOL( 500),NOEL,TOLD,NON  
 COMMON/OPUT:EPS,QH,XTMP,THIDT,THGS

COMMON/PARS/XKI,XKA,XKP,XKH,PTN,ADM,XJ,PTE,TAUC,TAUP,TAUV,OGM,  
 RVMX,XL,C,V,B,XKA,PS,XNCS,XNCO,XNCN,F,DT,TND,DTP,XKB  
 .CH2,OSE,XKSE,XJHE,OHE,TGF  
 .XNOP,WN,ZET,OBH,OBG,TAUI,MD,TALM  
 .XK8G,XK8H,TAG,TAM,T8G,T8H  
 .E1,TAUI,TAU2,A1,TAU3,TAU4,ZD,TAUS

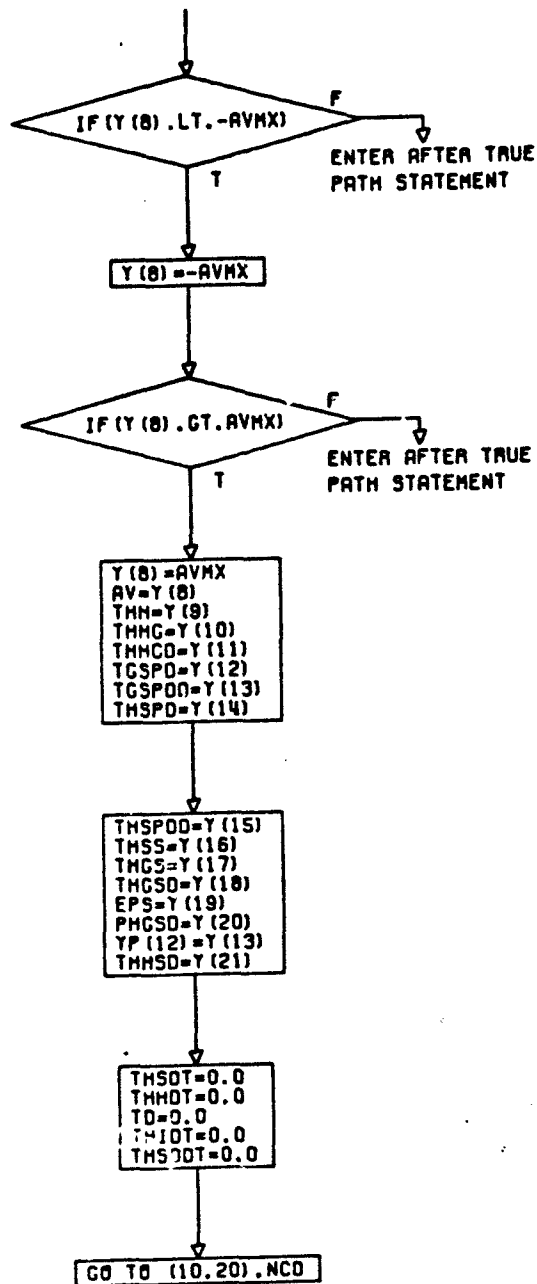
COMMON/CASE/NCS,NCO,NCN,NOP  
 DIMENSION Y(NEQ),YP(NEQ)  
 TWOPI=6.283185  
 THG=Y(1)  
 THGD=Y(2)



CONT. ON PG 2

PG 1 OF 17

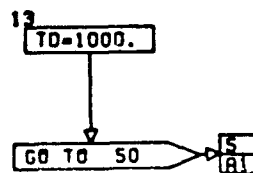
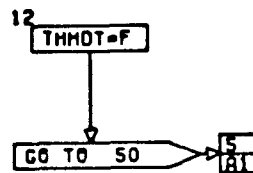
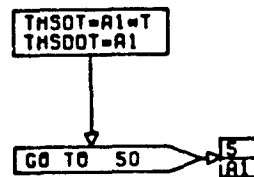
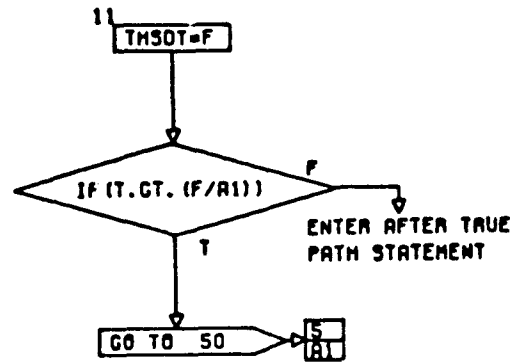
A-77



CONT. ON PG 3

PG 2 OF 17

10  
GO TO (11,12,13,14),NCN

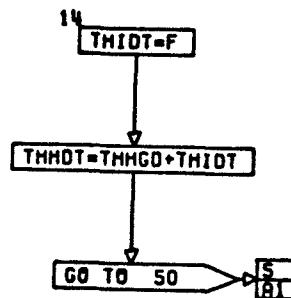


CONT. ON PG 4

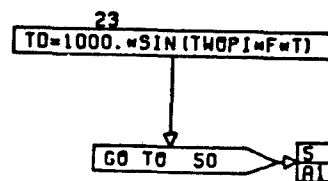
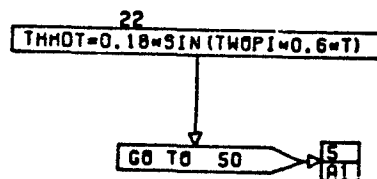
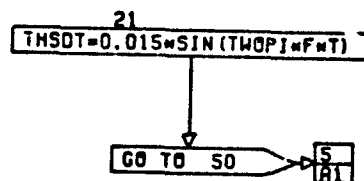
PG 3 OF 17

A-79



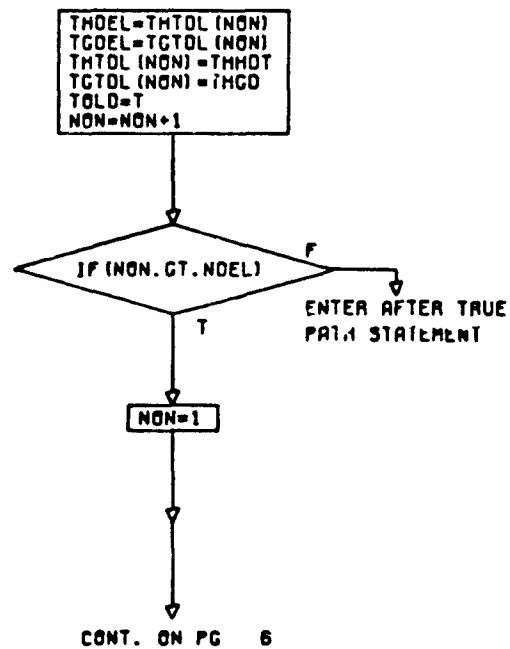
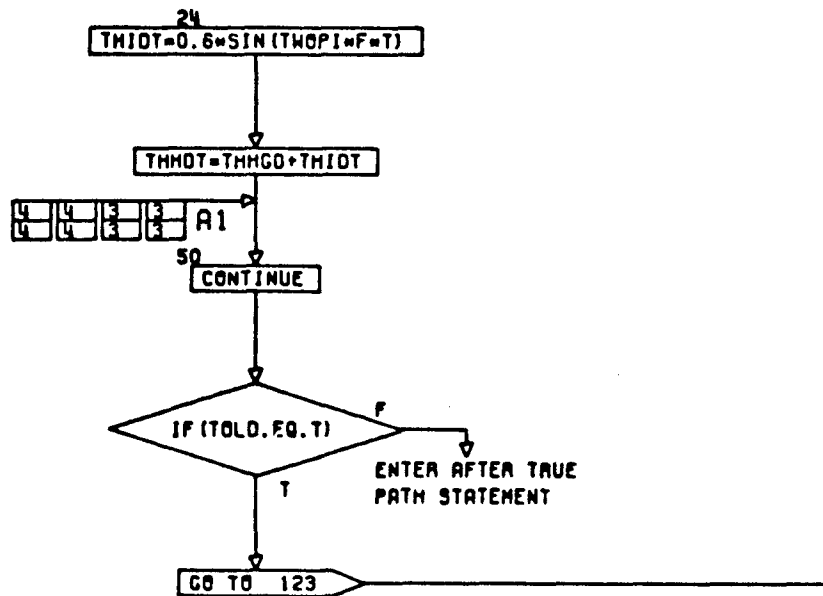


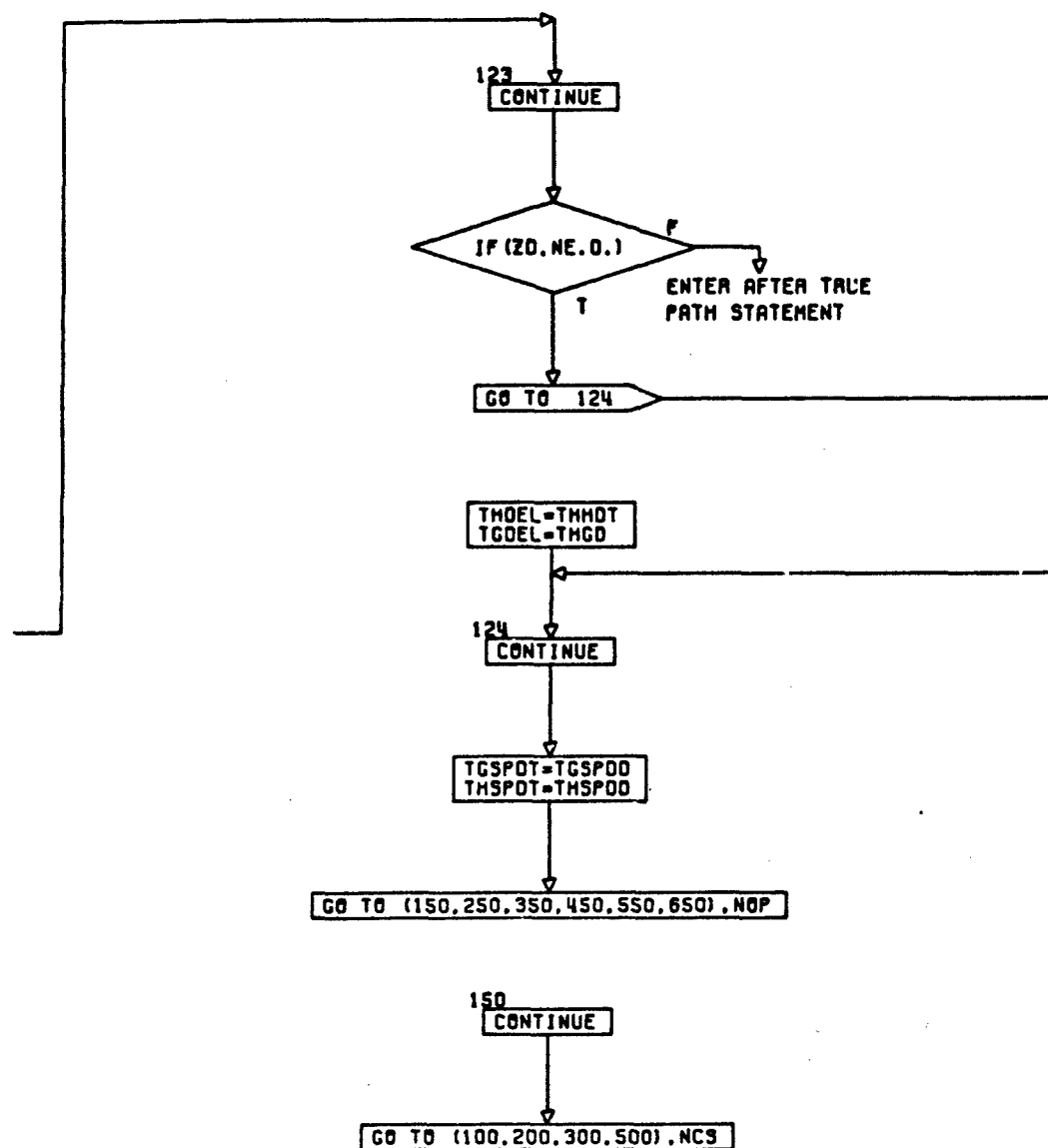
20  
GO TO (21,22,23,24),NCN



CONT. ON PG 5

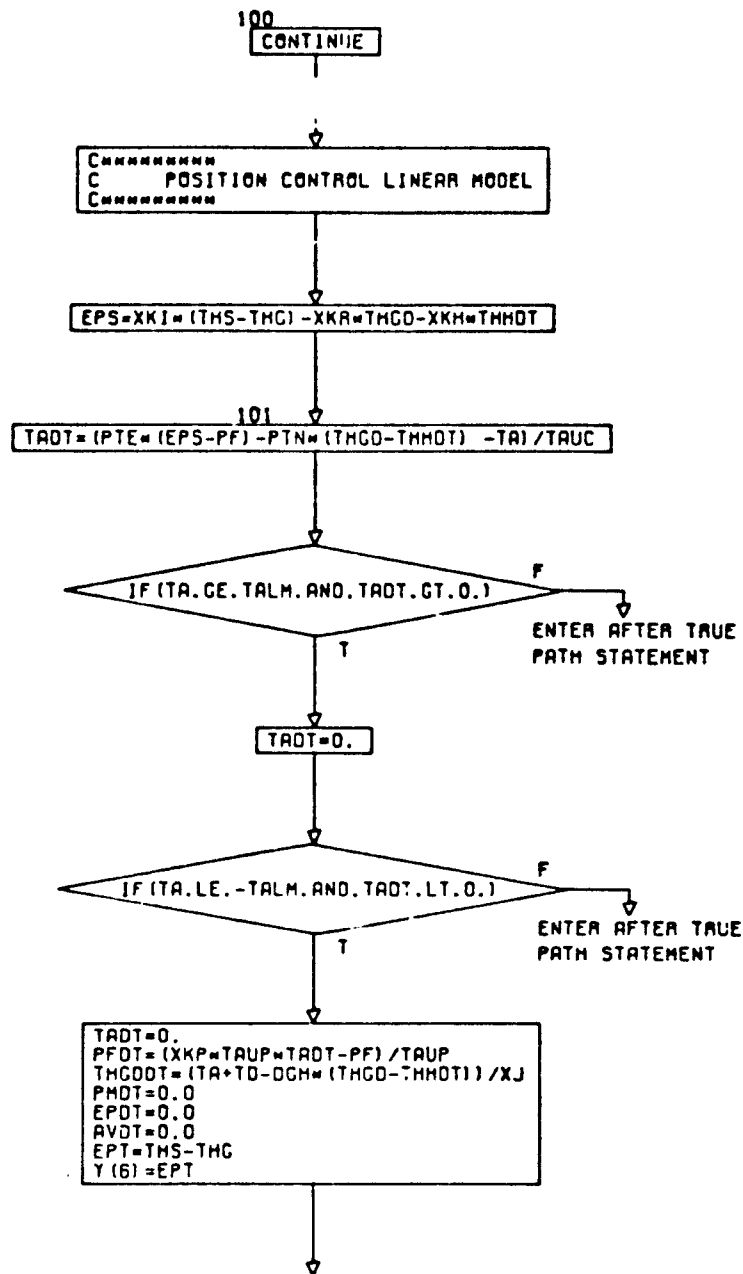
PG 4 OF 17





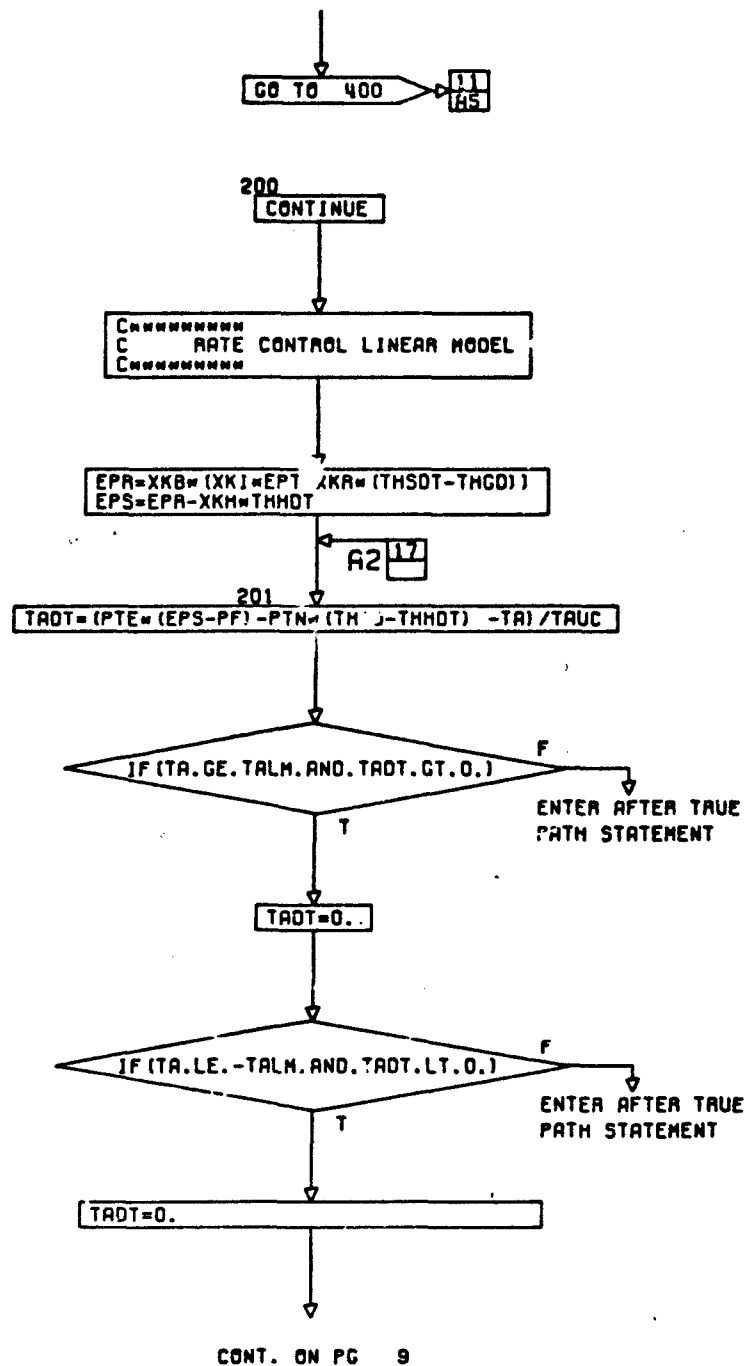
CONT. ON PG 7

PG 6 OF 17



CONT. ON PG 8

PG 7 OF 17



$PFDT = (XKP \cdot TAUP \cdot TADT - PF) / TAUP$   
 $EPDT = THSDT - THGD$   
 $PHDT = 0.0$   
 $AVDT = 0.0$   
 $THGDOOT = (TA \cdot TD - DGH \cdot (THGD - THHDT)) / XJ$

GO TO 400 11  
A5

300  
CONTINUE

C\*\*\*\*\*  
 C   NONLINEAR FLOW MODEL - RATE CONTROL  
 C\*\*\*\*\*

$EPA = XKB \cdot (XKI \cdot EPT + XKA \cdot (THSDT - THGD))$   
 $EPS = EPA - XKH \cdot THHDT$

A3 14  
16  
17

303  
CONTINUE

$EPDT = THSDT - THGD$

A4 12  
17

301  
CONTINUE

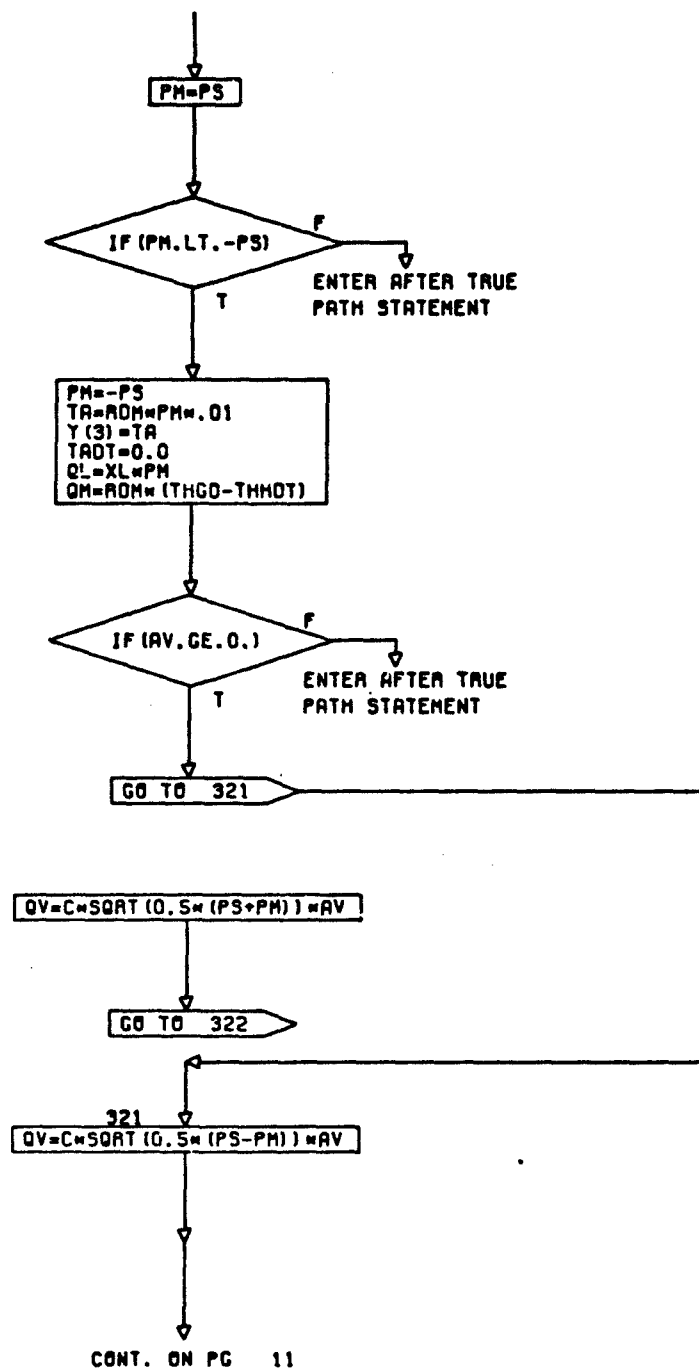
IF (PM.GT.PS) F  
ENTER AFTER TRUE  
PATH STATEMENT

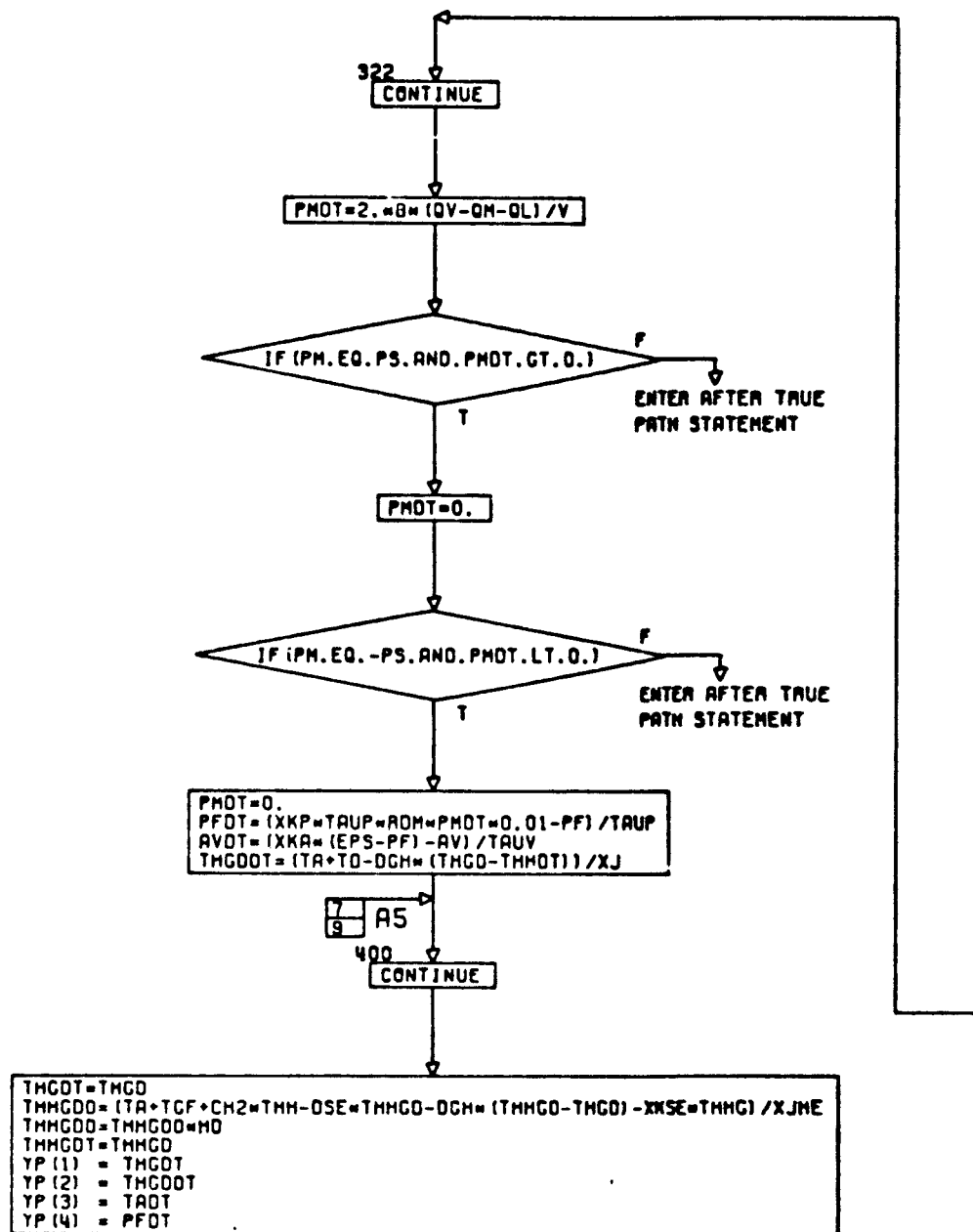
T

CONT. ON PG 10

PG 9 OF 17

A-85





CONT. ON PG 12

PG 11 OF 17

A-87



YP (5) = PHDT  
 YP (6) = EPDT  
 YP (7) = THSDT  
 YP (8) = AVDT  
 YP (9) = THHDT  
 YP (10) = THHGD  
 YP (11) = THHGD  
 XTMP=XKA= (THSDT-THGD) \*XKI=EPT

Y (19) = EPS

RETURN

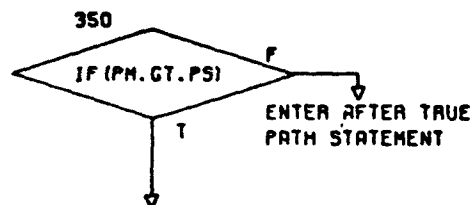
C\*\*\*\*\*  
 C NONLINEAR FLOW MODEL - POSITION CONTROL  
 C\*\*\*\*\*

500  
 EPT=THS-THG

Y (6) = EPT  
 EPDT=0.0  
 EPS=XKI=EPT-XKA=THGD-XKH=THHDT

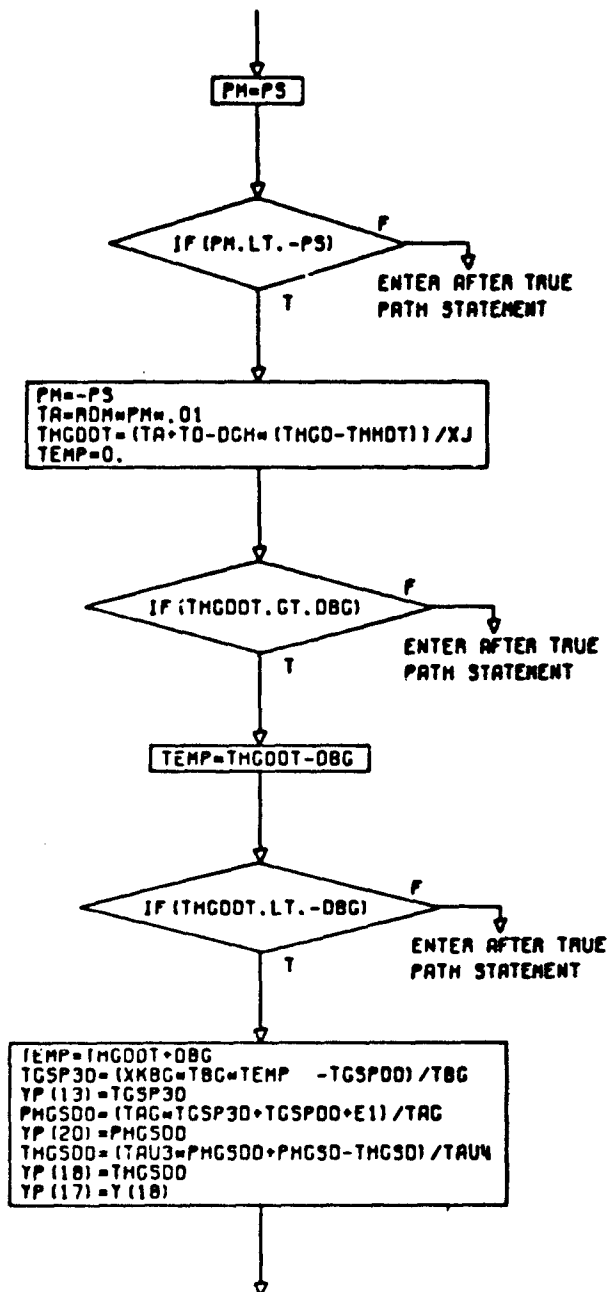
GO TO 301

9  
R4



CONT. ON PG 13

PG 12 OF 17



CONT. ON PG 14

PG 13 OF 17

A-89

$EPSPA = XKI * (THS - THGS) + XKA * (THSOT - THGSD)$   
 $EPSPO = XKI * (THSOT - THGSD) + XKA * (THSODT - THGSDO)$   
 $EPSOT = ((TAU2 / TAU1) * (TAU1 * EPSPO + EPSPA) - EPS) / TAU2$   
 $YP(19) = EPSOT$

GO TO (101, 202, 303, 501), NC9

450  
 $TCSP30 = 10. * WN * WN * (THGO - PHGSD) - 2. * ZET * WN * TCSPDO - WN * WN * TCSPD$

$YP(13) = TCSP30$   
 $PHGSOD = (1.035 * TCSPDO + TCSPD - PHGSD) / 0.007$   
 $THGSOD = (TAU3 * PHGSOD + PHGSD - THGSD) / TAU4$   
 $YP(20) = PHGSOD$   
 $YP(18) = THGSOD$   
 $YP(17) = Y(18)$   
 $EPS = XKI * (THS - THGS) + XKA * (THSOT - THGSD)$

GO TO 303 → 9  
103

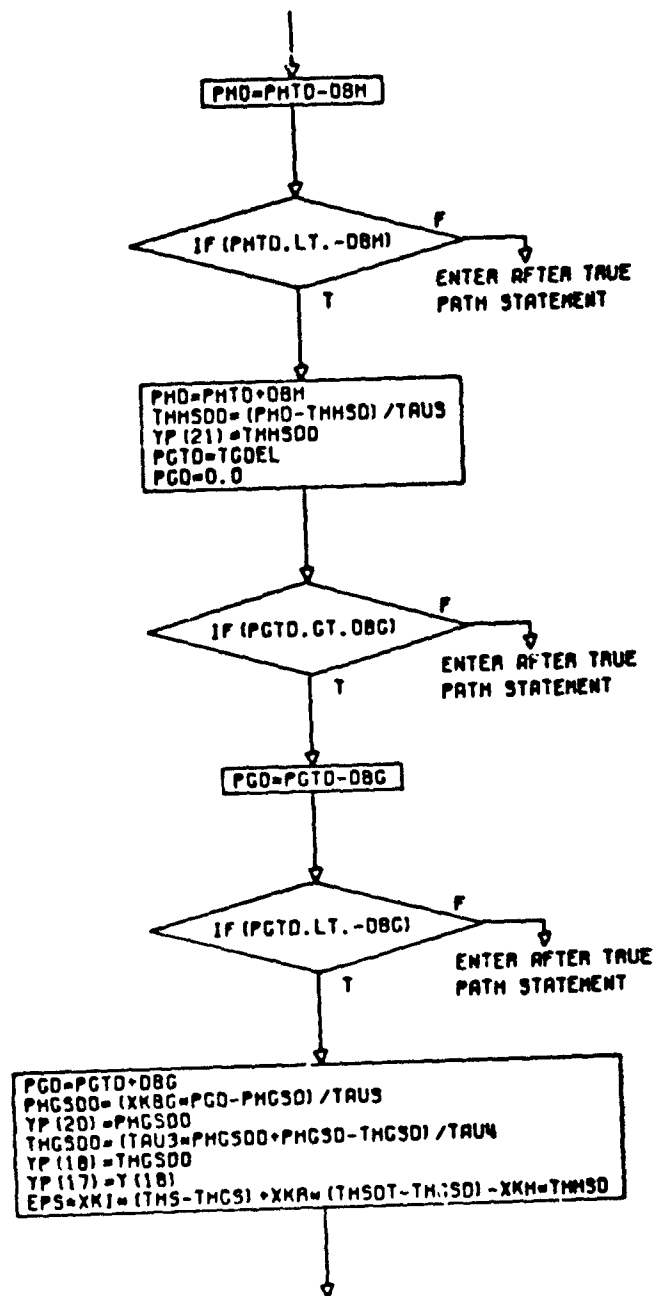
550  
 $PMTD = THDEL$

$PHD = 0.0$

IF (PMTD.GT.DBM)  
 F  
 ENTER AFTER TRUE  
 PATH STATEMENT  
 T

CONT. ON PG 15

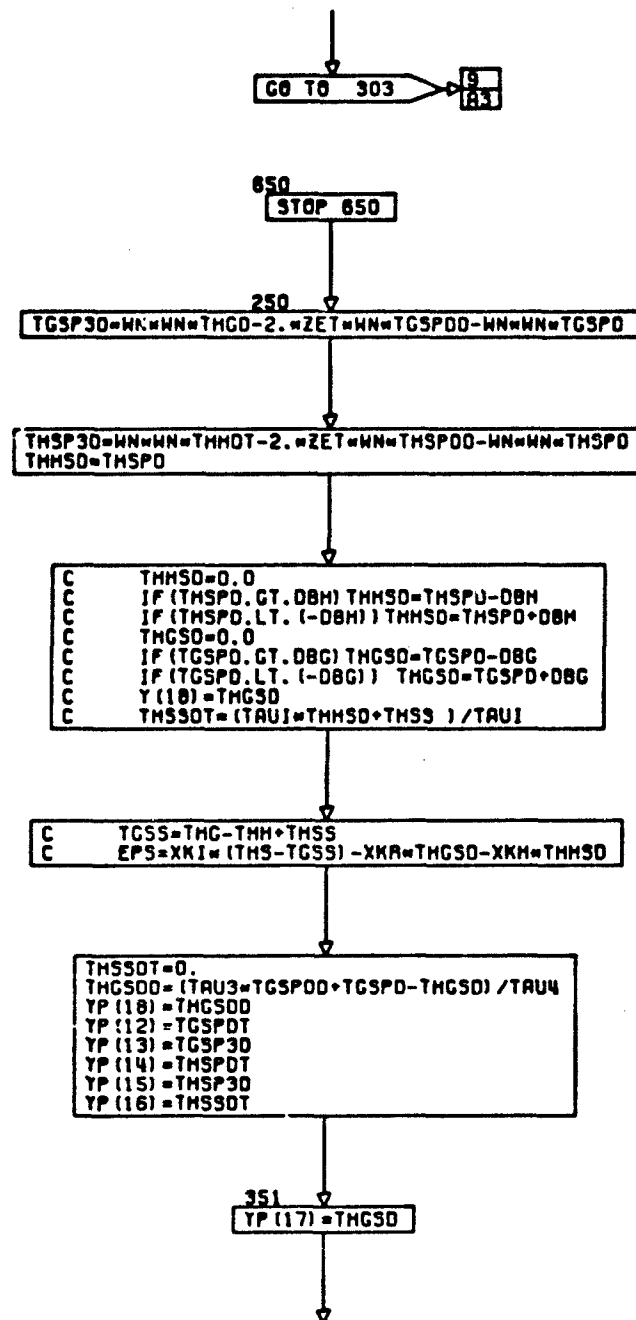
PG 14 OF 17



CONT. ON PG 16

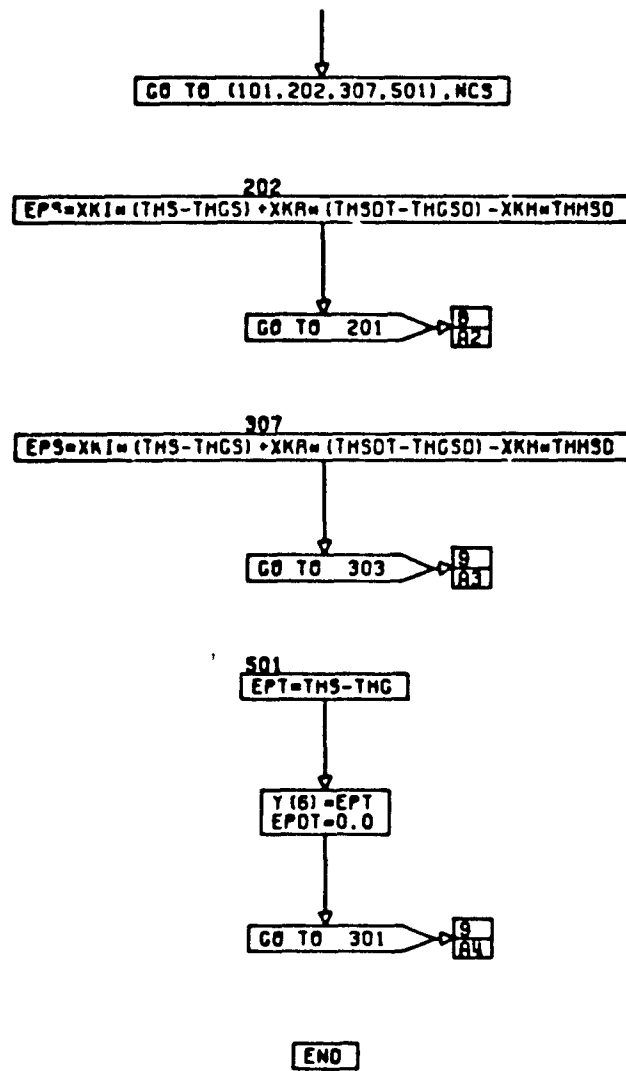
PG 15 OF 17

A-91

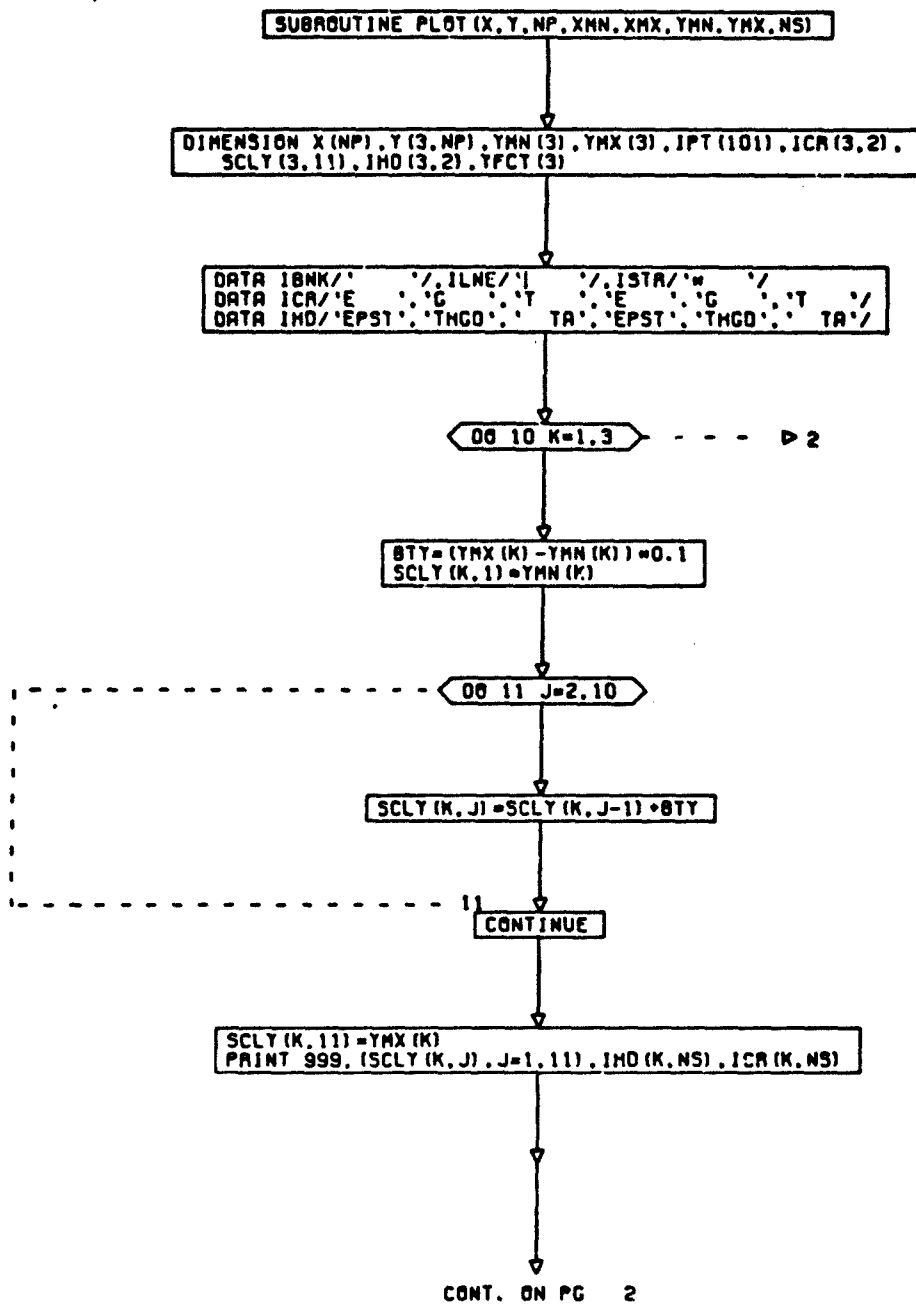


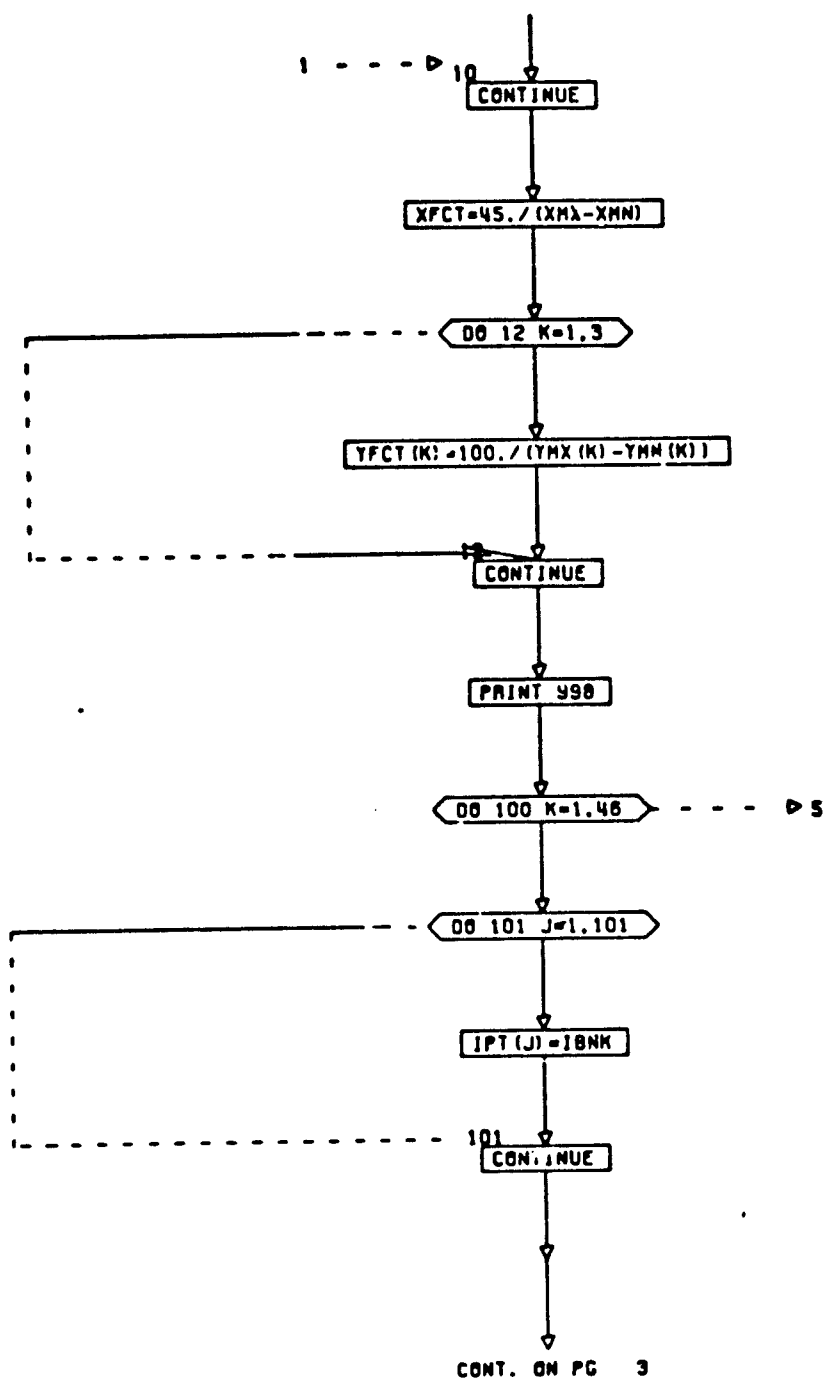
CONT. ON PG 17

PG 16 OF 17



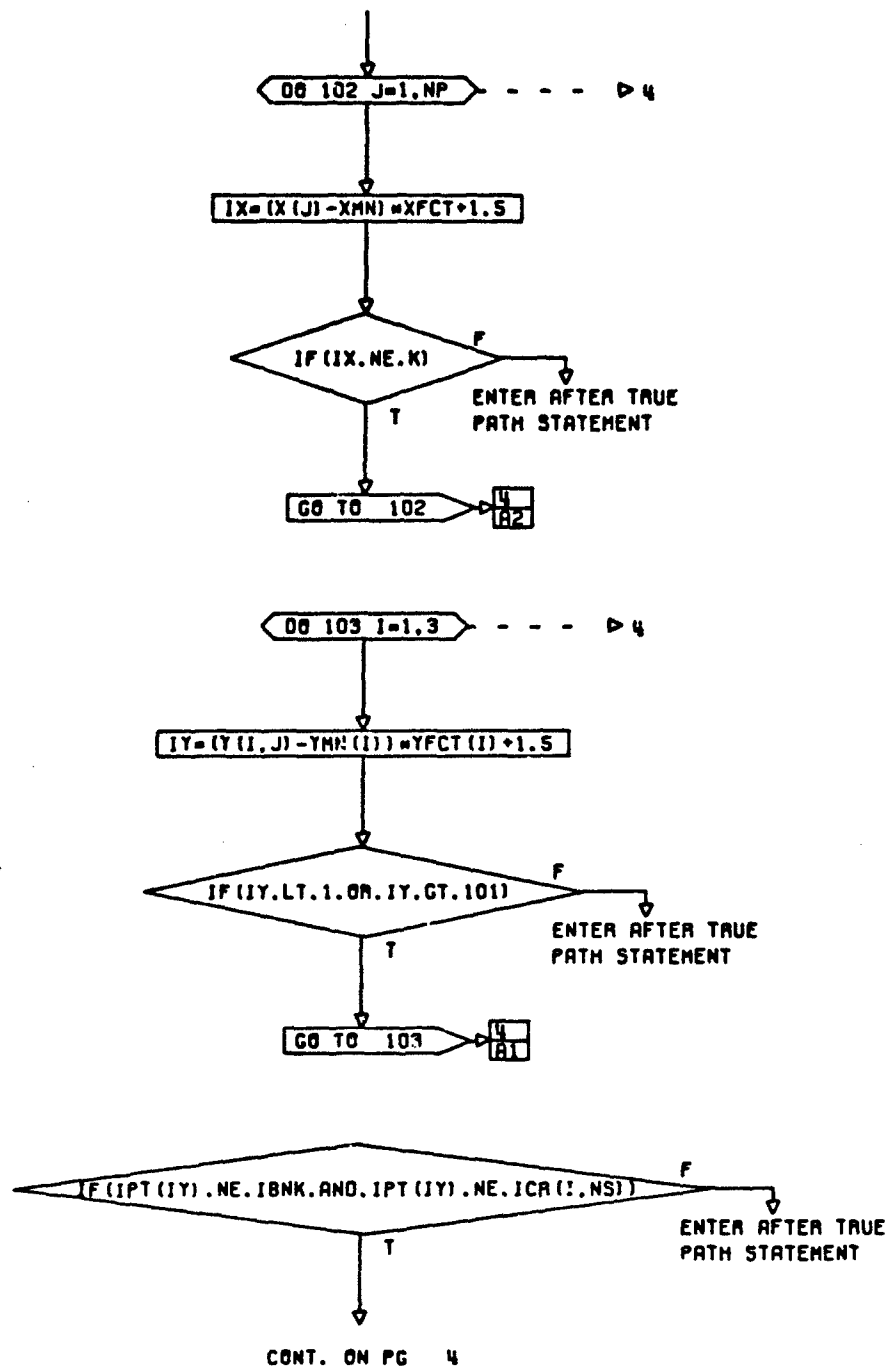
PG 17 FINAL

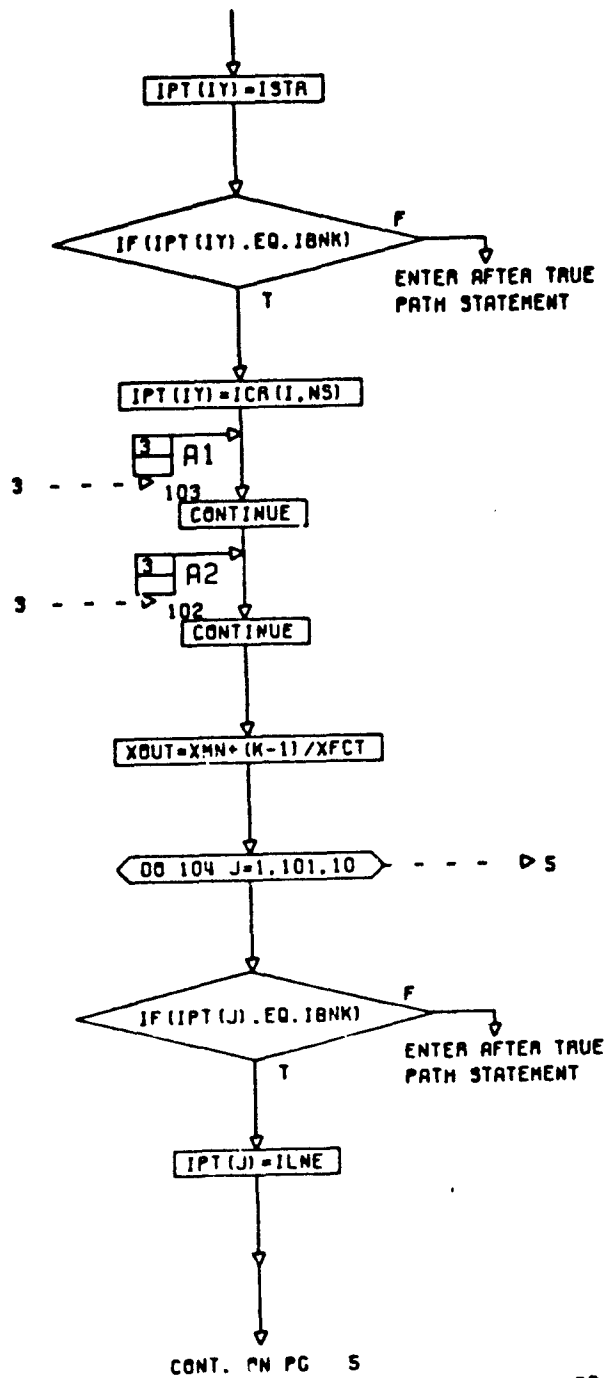




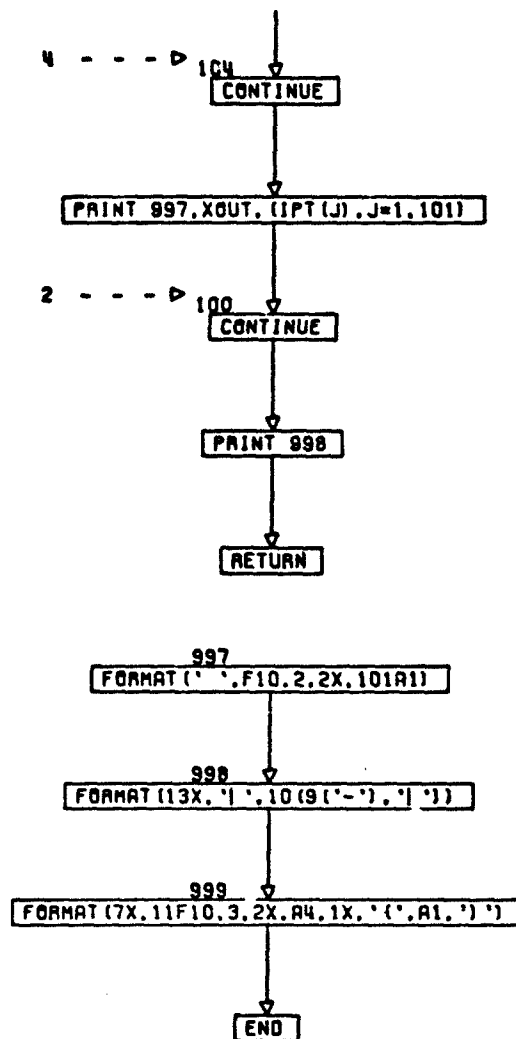
PG 2 OF 5







PG 4 OF 5



SUBROUTINE SCALE (Y, NP, YMN, YMX, NC)  
 DIMENSION Y (NC, NP), YMN (NC), YMX (NC)

00 10 K=1, NC - - - - -> 3

YMAX=0.0

00 20 J=1, NP

TMP=ABS (Y (K, J))

IF (TMP.GT. YMAX)

ENTER AFTER TRUE  
 PATH STATEMENT

YMAX=TMP

20 CONTINUE

NPOW=0

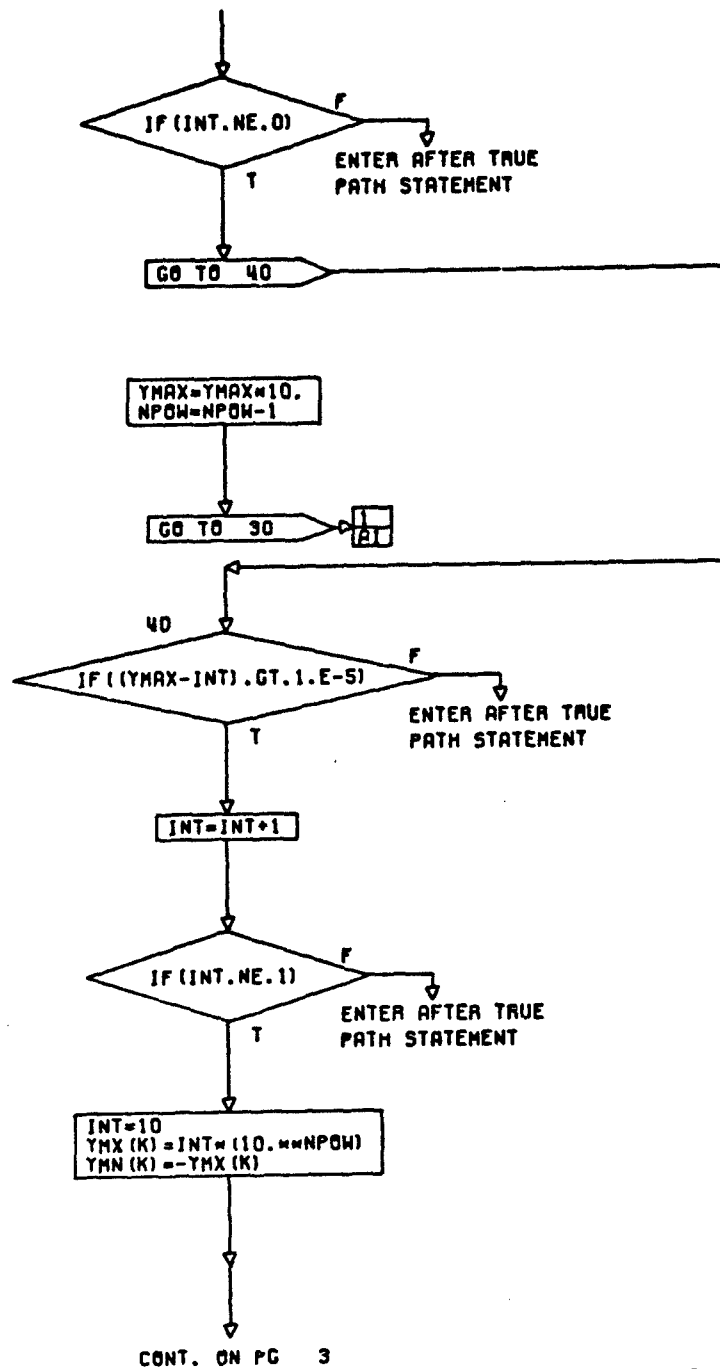
A1 2

30 INT=YMAX

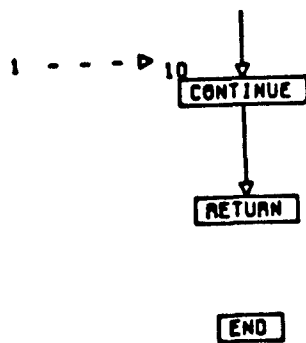
CONT. ON PG 2

PG 1 OF 3

A-99

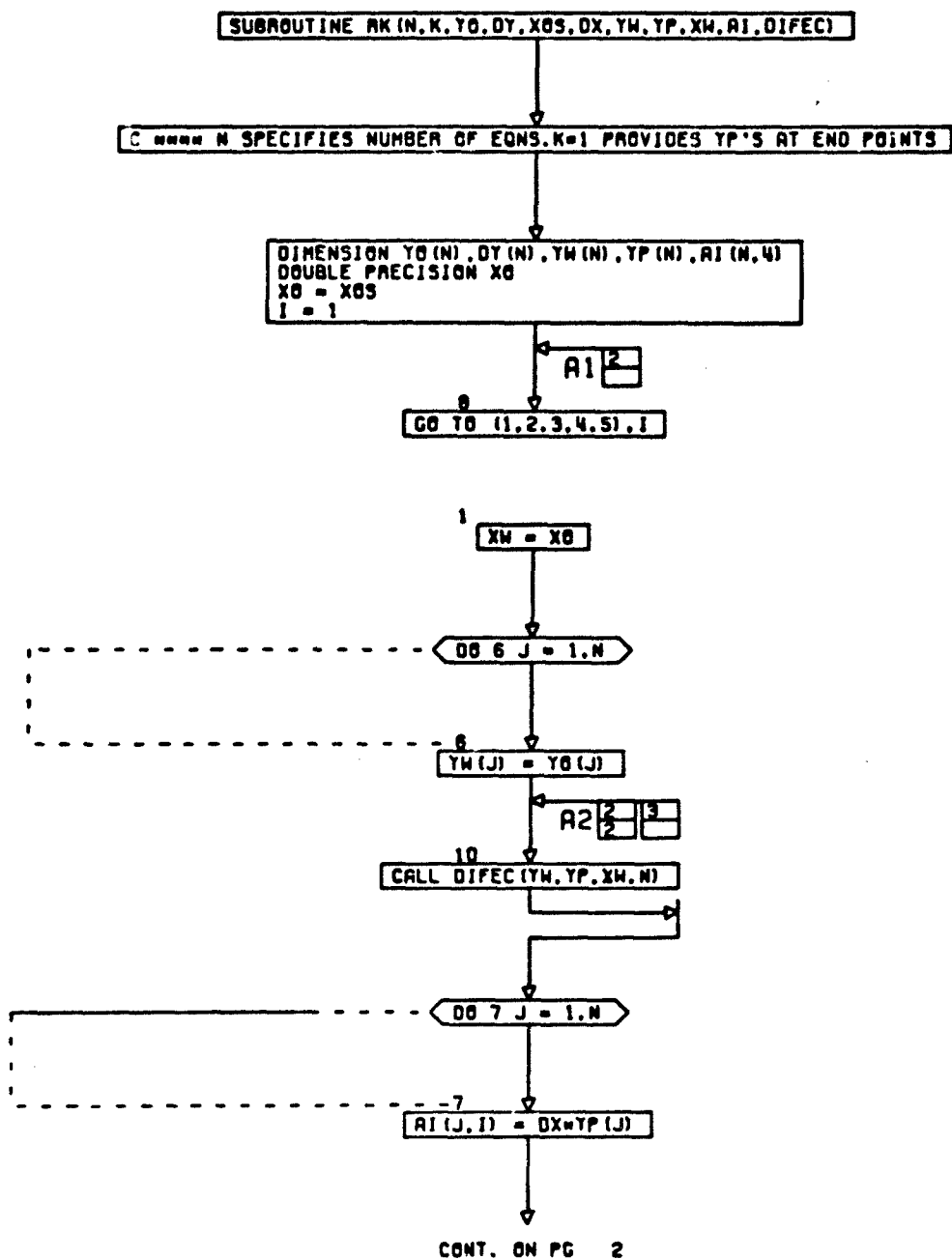


PG 2 OF 3



PG 3 FINAL

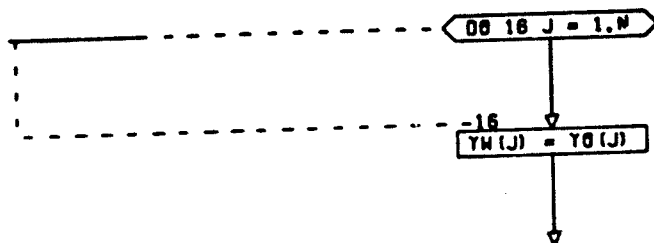
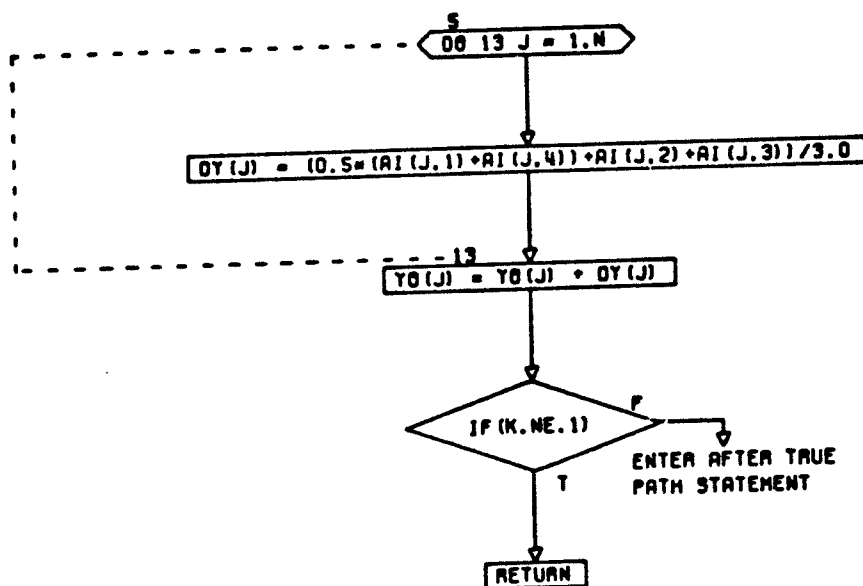
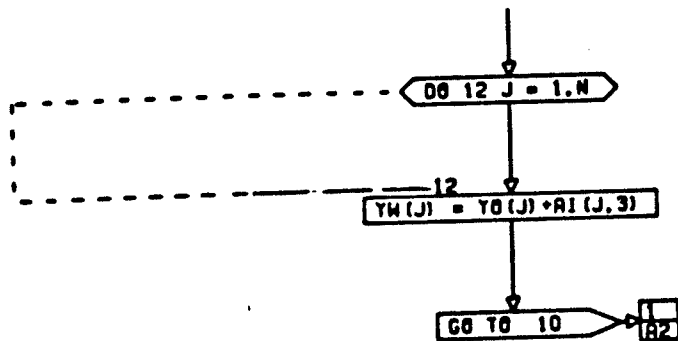
A-101





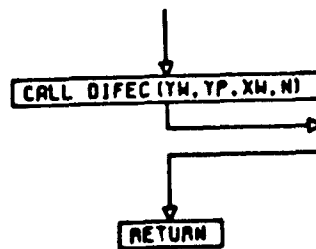
2 of 4





CONT. ON PG 4

PG 3 OF 4



END

PG 1

A-105

**APPENDIX B**  
**ANALOG AND HYBRID WIRING DIAGRAMS**

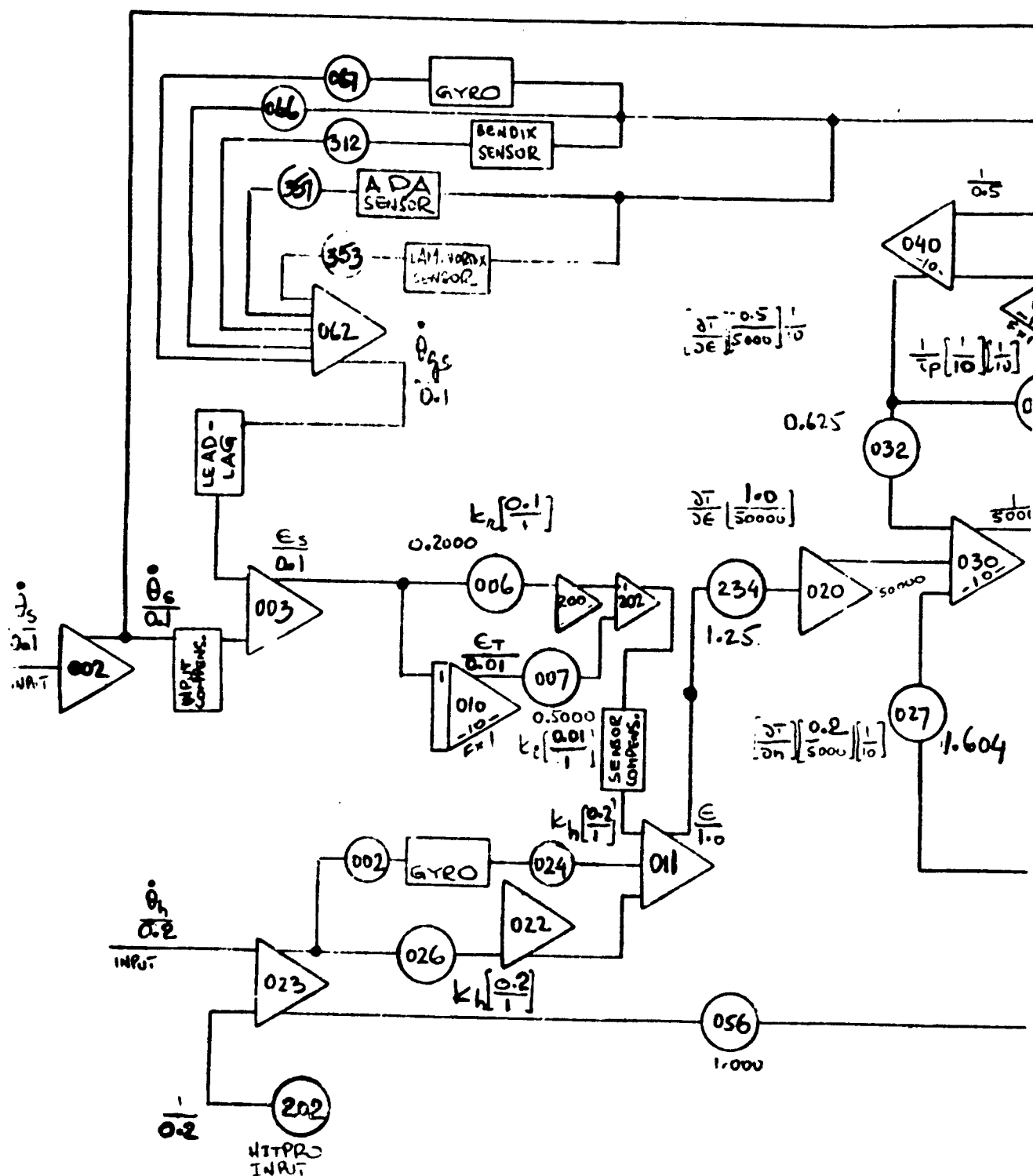
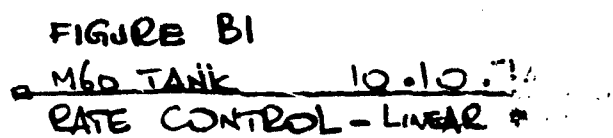


Figure B-1. Rate Control Sy



72

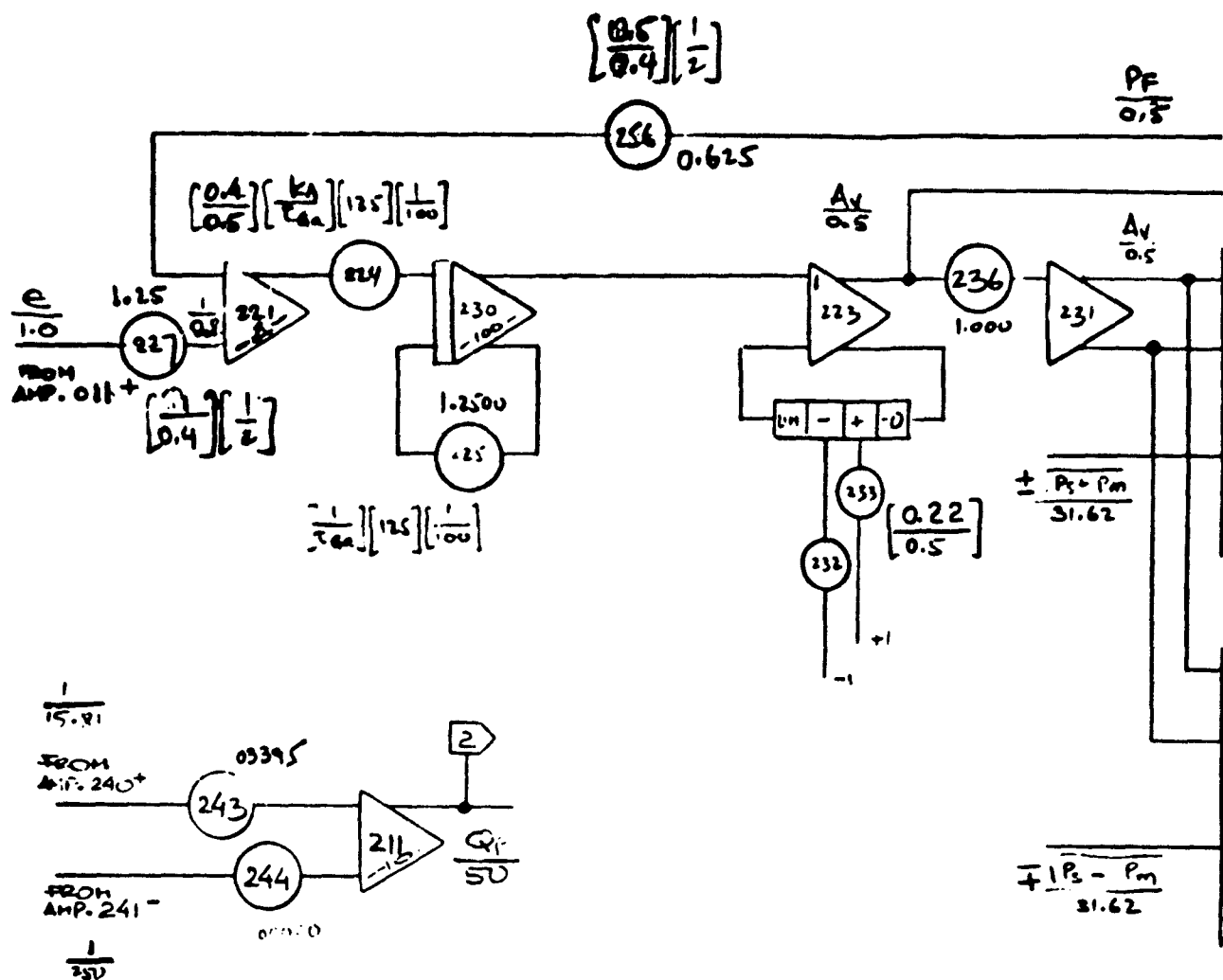


FIGURE B2

M60 TANK  $8.12 \times 10^{-3}$   
 NONLINEAR FLOW MODEL  
 RATE CONTROL

DUAL  
 SQUARE ROOT  
 CIRCUIT

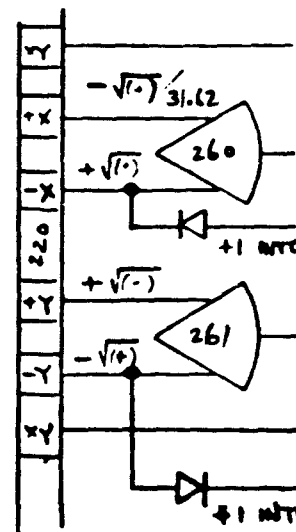
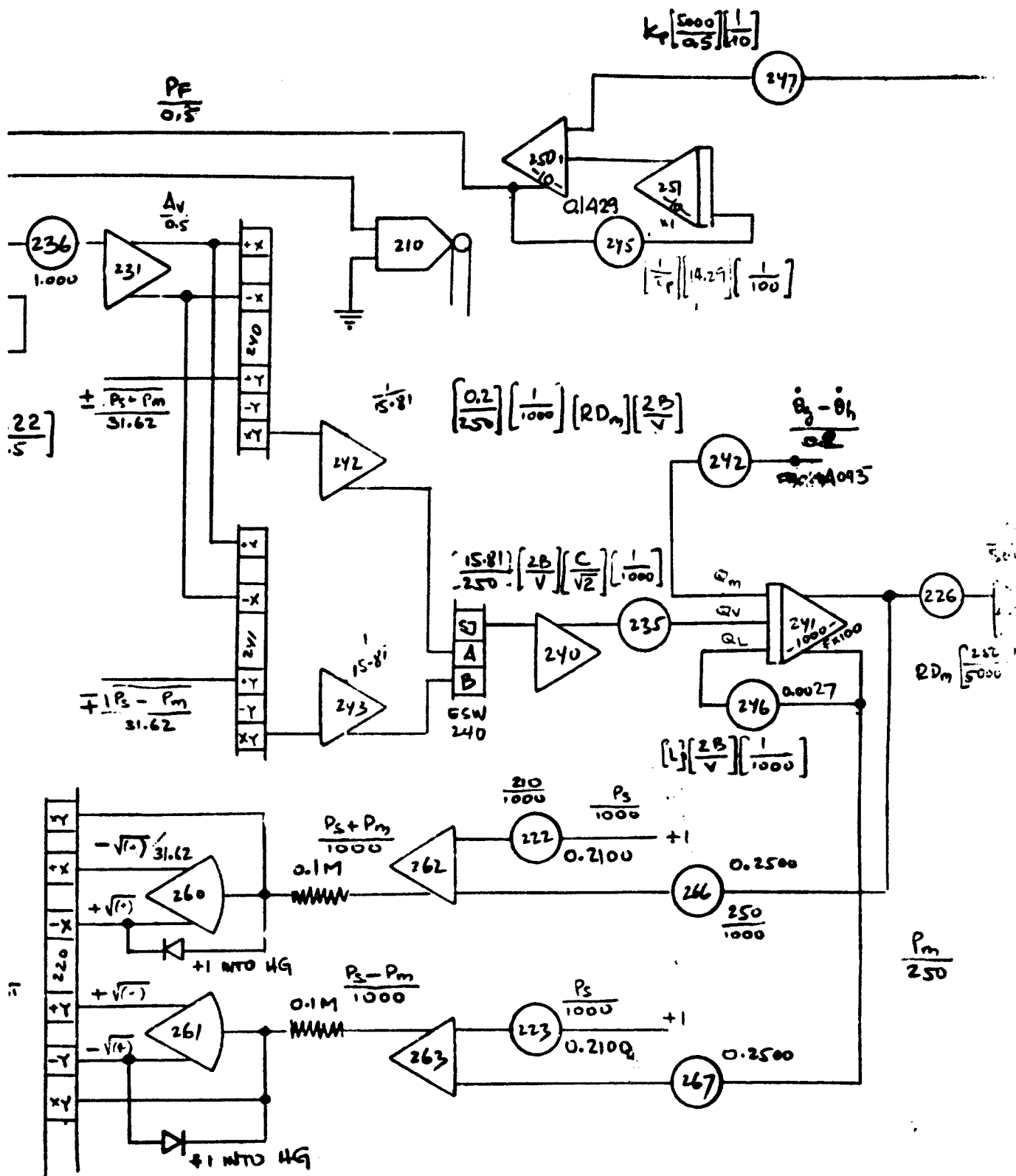
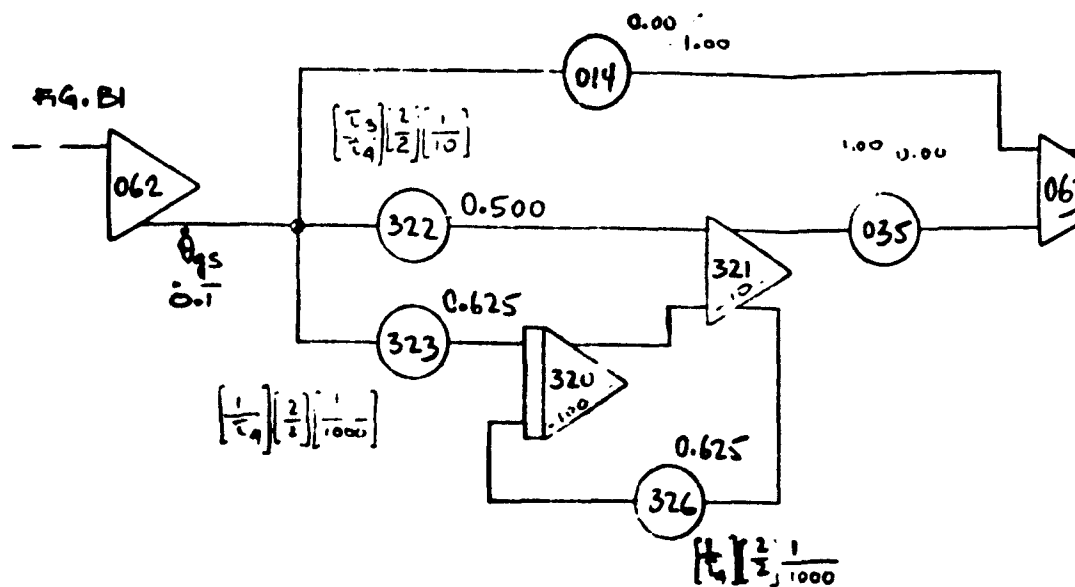


Figure B-2. Nonlinear Flow Model in Rate



linear Flow Model in Rate Control System

LEAD-LAG



Coulomb Friction  
Stiction

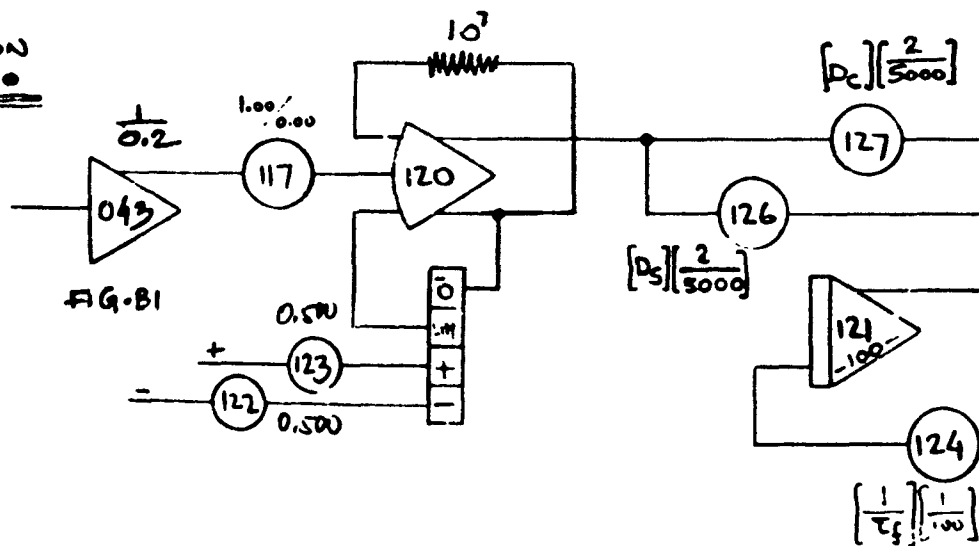
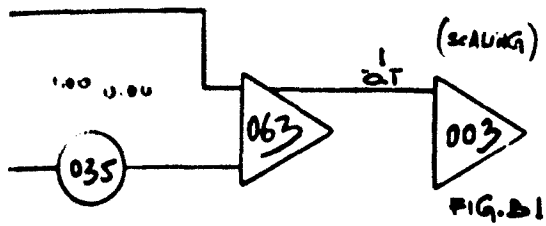


Figure B-3. Compensation Circuit





100

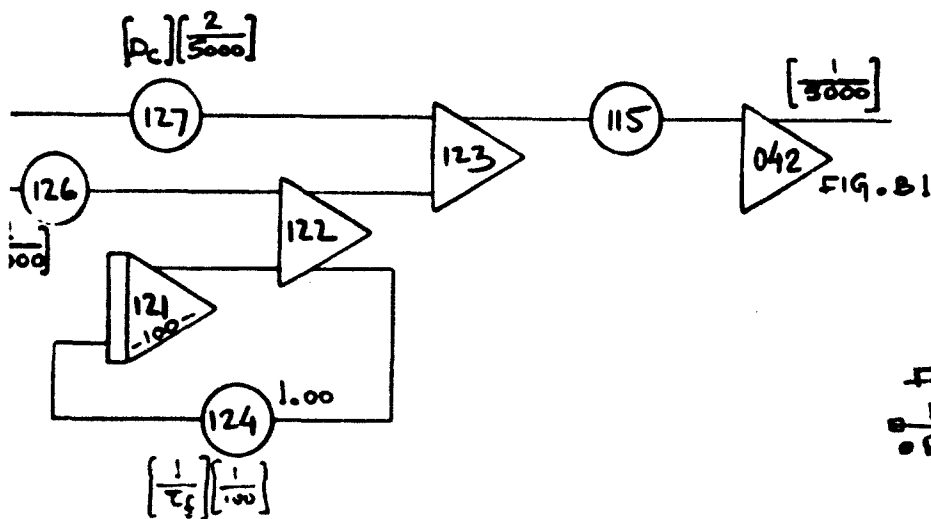


FIG. B3  
M60 TANK 10-22-74  
• RATE CONTROL • COMPENSATION CIRCUIT

Compensation Circuits in Rate Control System

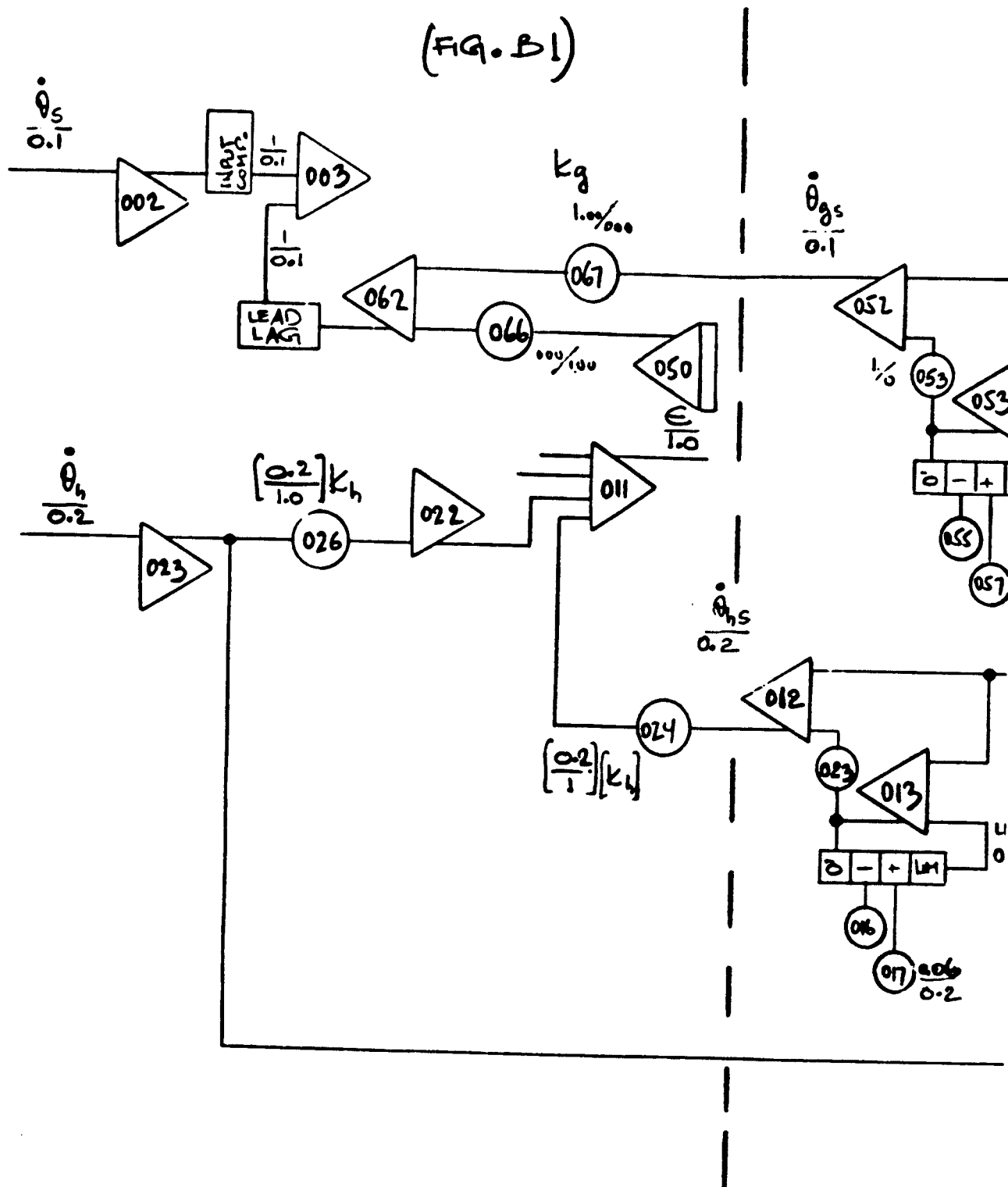


Figure B-4. Gyros in Rate Cont

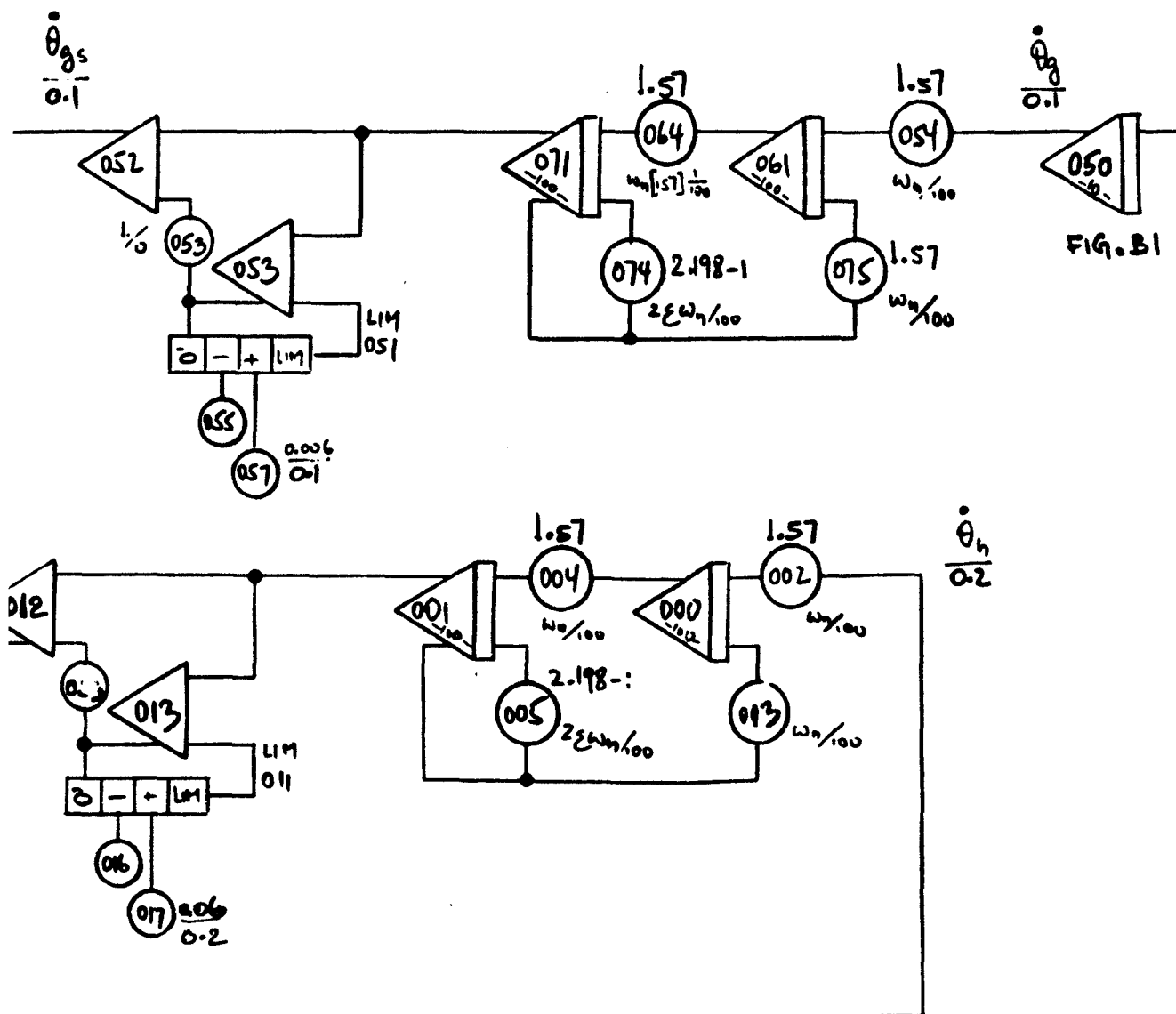


FIGURE B4  
M60 • RATE CONTROL  
GYROS • 10 • 4 • 74

-4. Gyros in Rate Control System

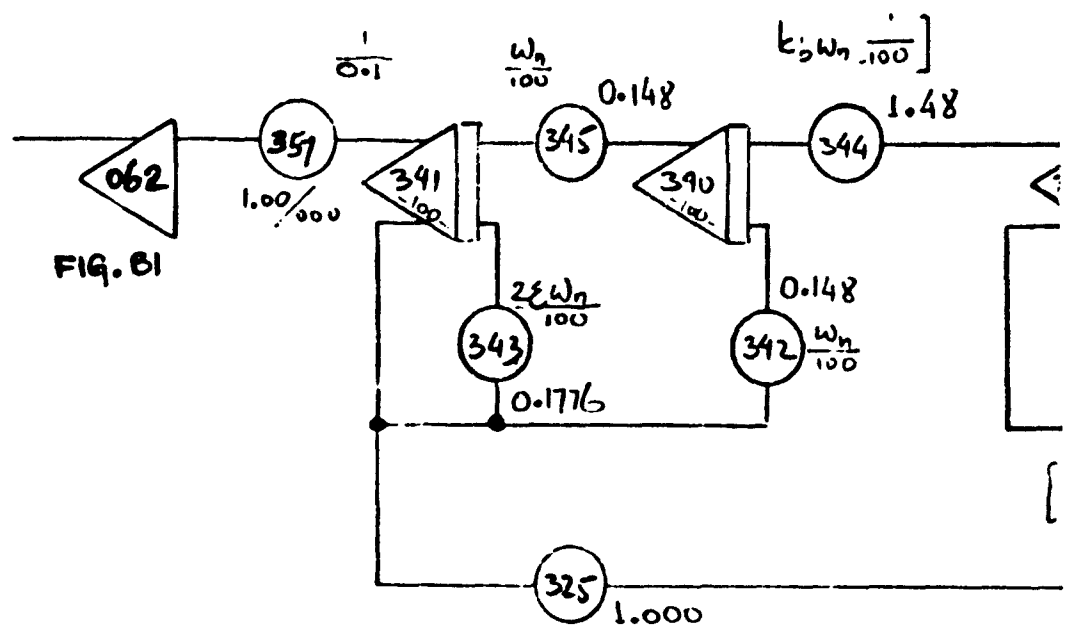


FIG. B5  
 M60 TANK • RATE CONTROL  
 AIRESEARCH PNEUMATIC ACCELEROMETER  
 11-12-74

Figure B-5. Airesearch Pneumatic Acce



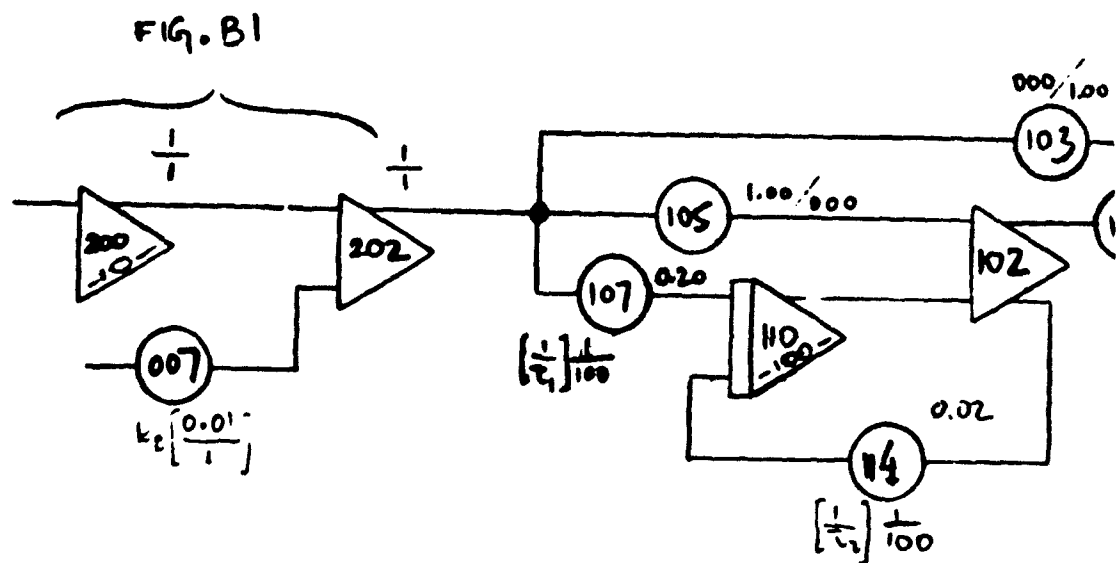
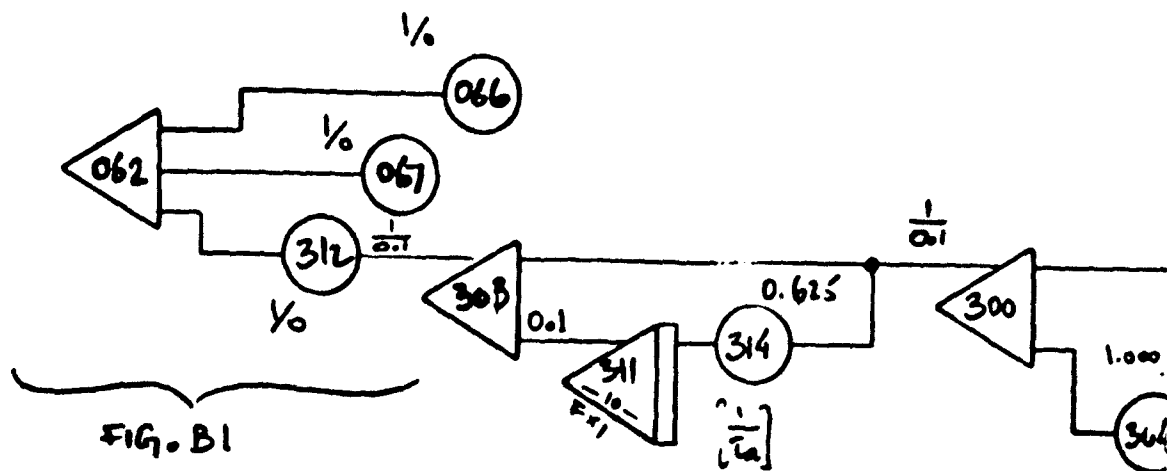


Figure B-6. Bendix Sensor in R

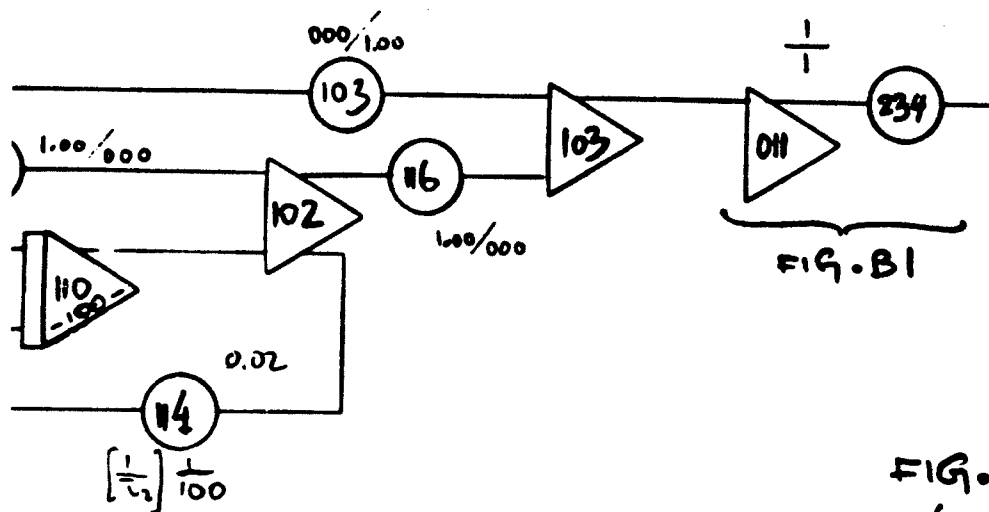
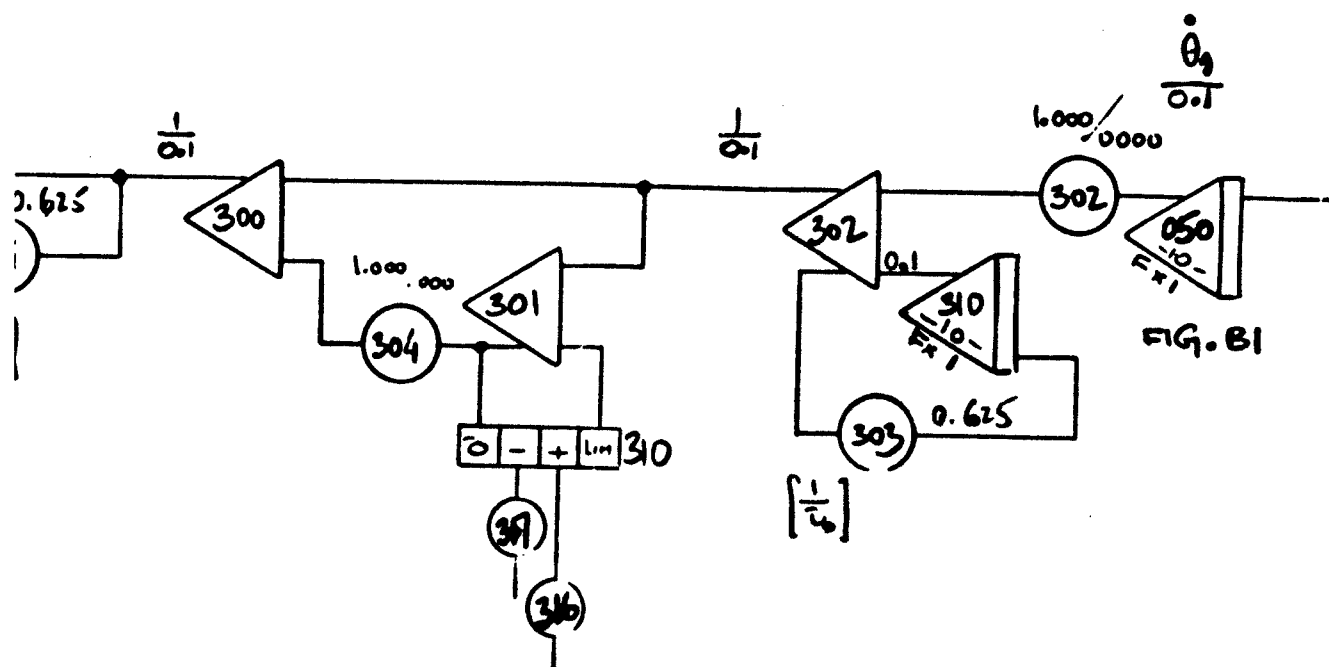


FIG. B6  
M60 TANK • RATE CONTROL  
BENDIX SENSOR

11-7-74

Figure B-6. Bendix Sensor in Rate Control System

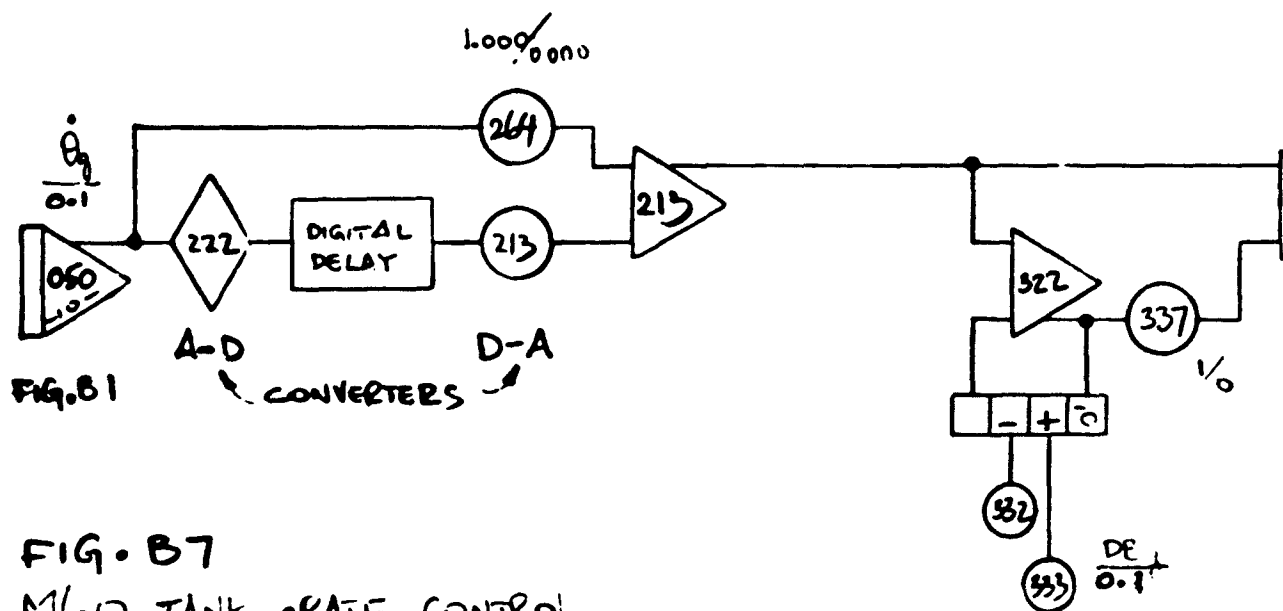
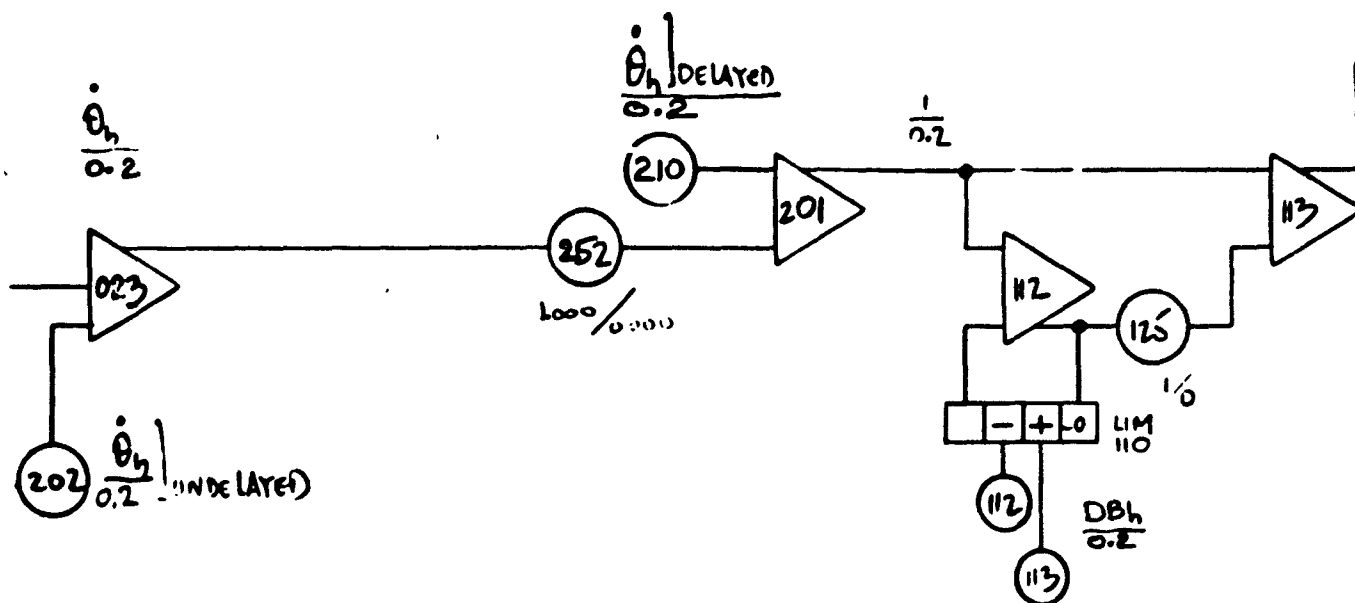


FIG. B7  
M60 TANK RATE CONTROL  
LAMINAR VORTEX SENSOR 11-19-74

Figure B-7. Laminar Vortex Sensor in R





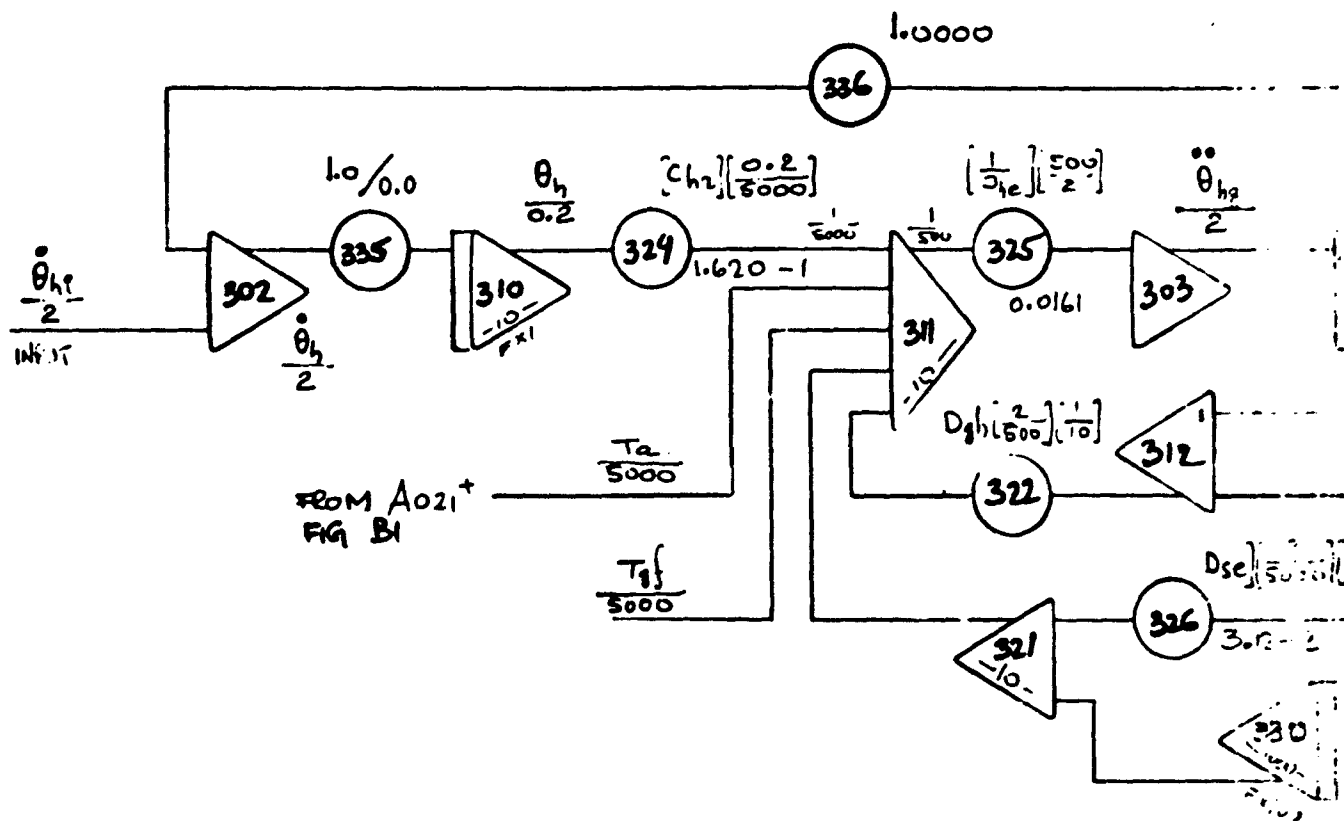


FIG. B8

MGO TANK • 10.8.74.  
 RATE CONTROL • LINEAR FLOW  
 HULL DYNAMICS

Figure B-8. Hull Dynamics in Rate Control System

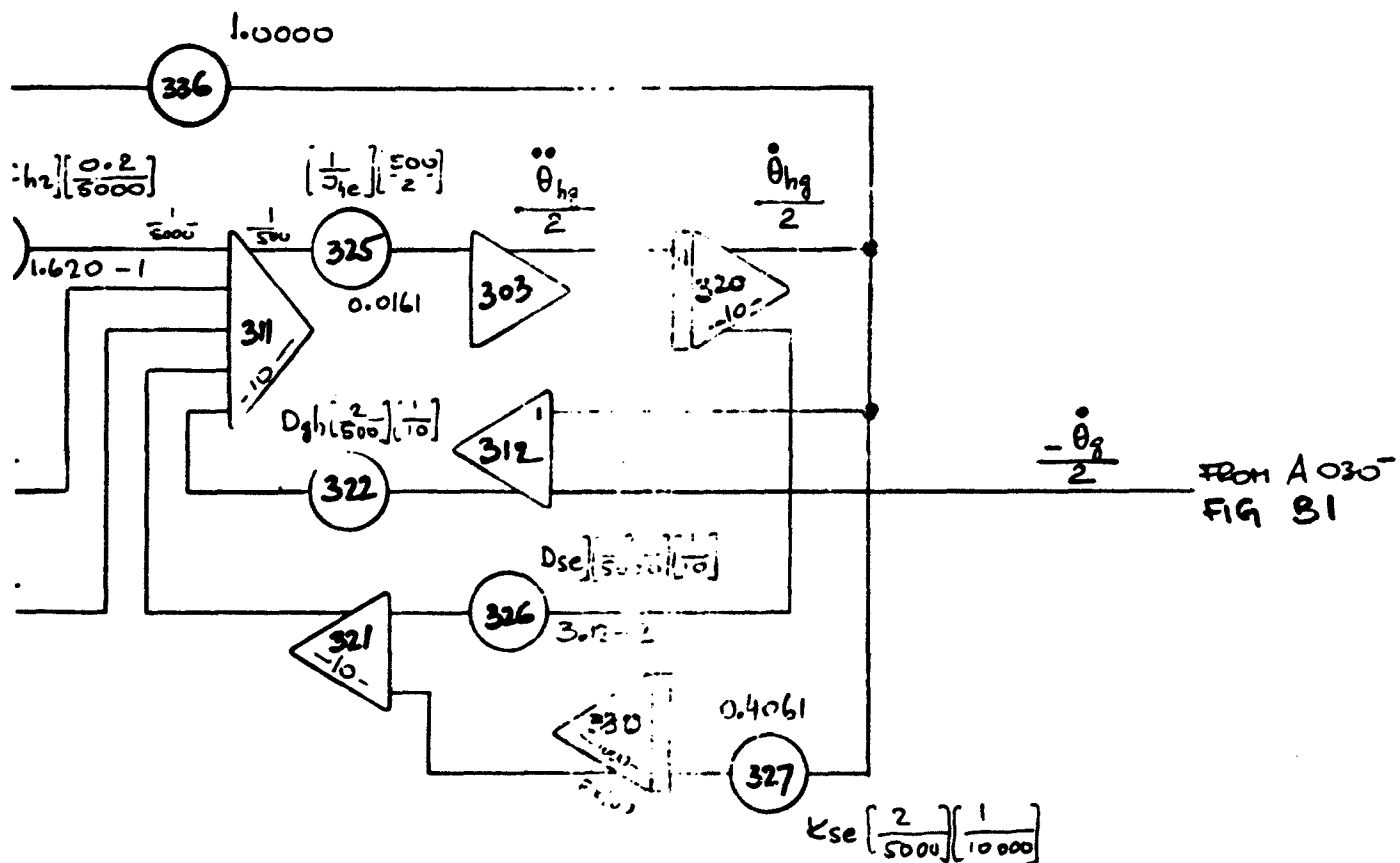


Figure B-8. Hull Dynamics in Rate Control System

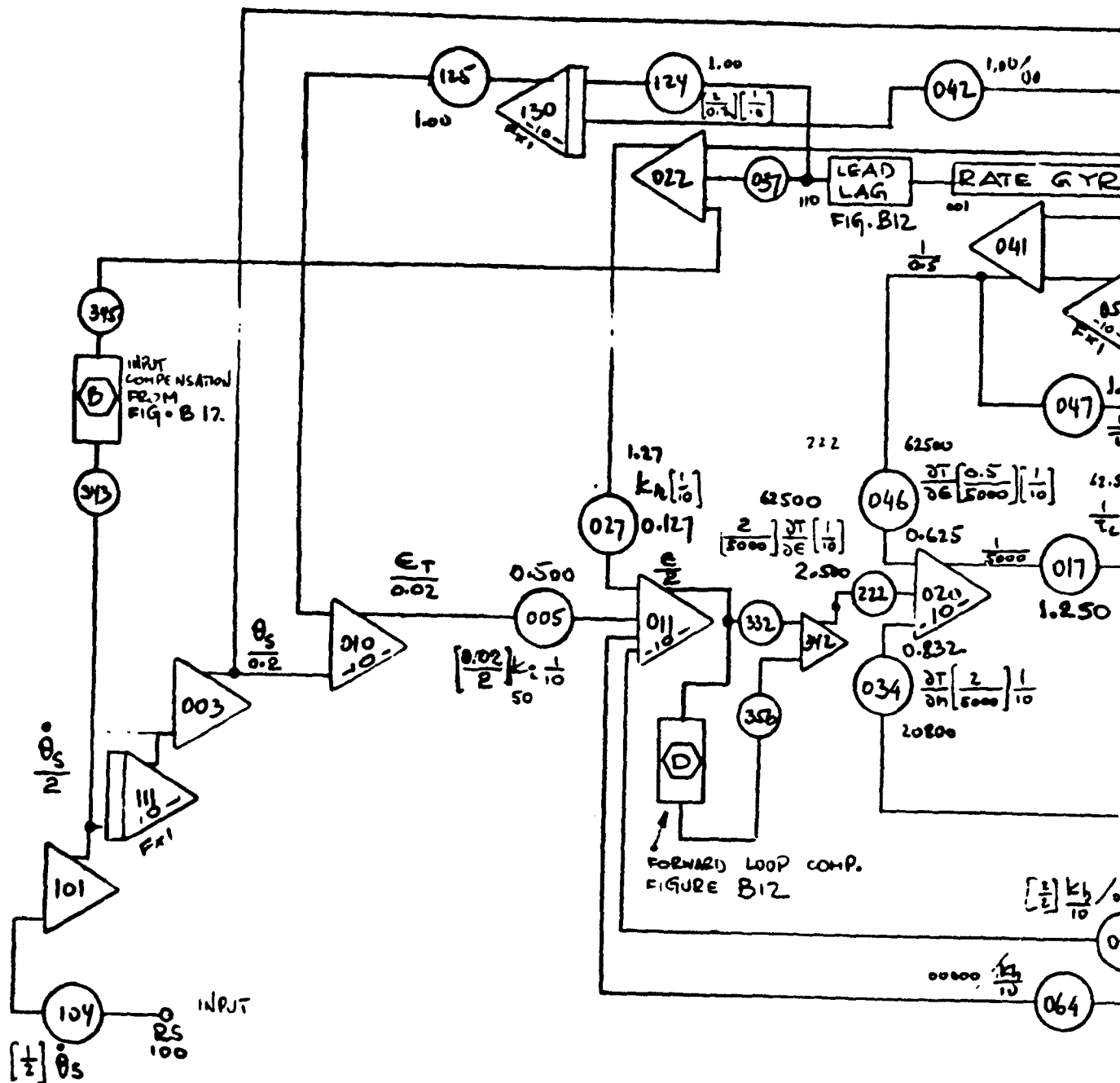
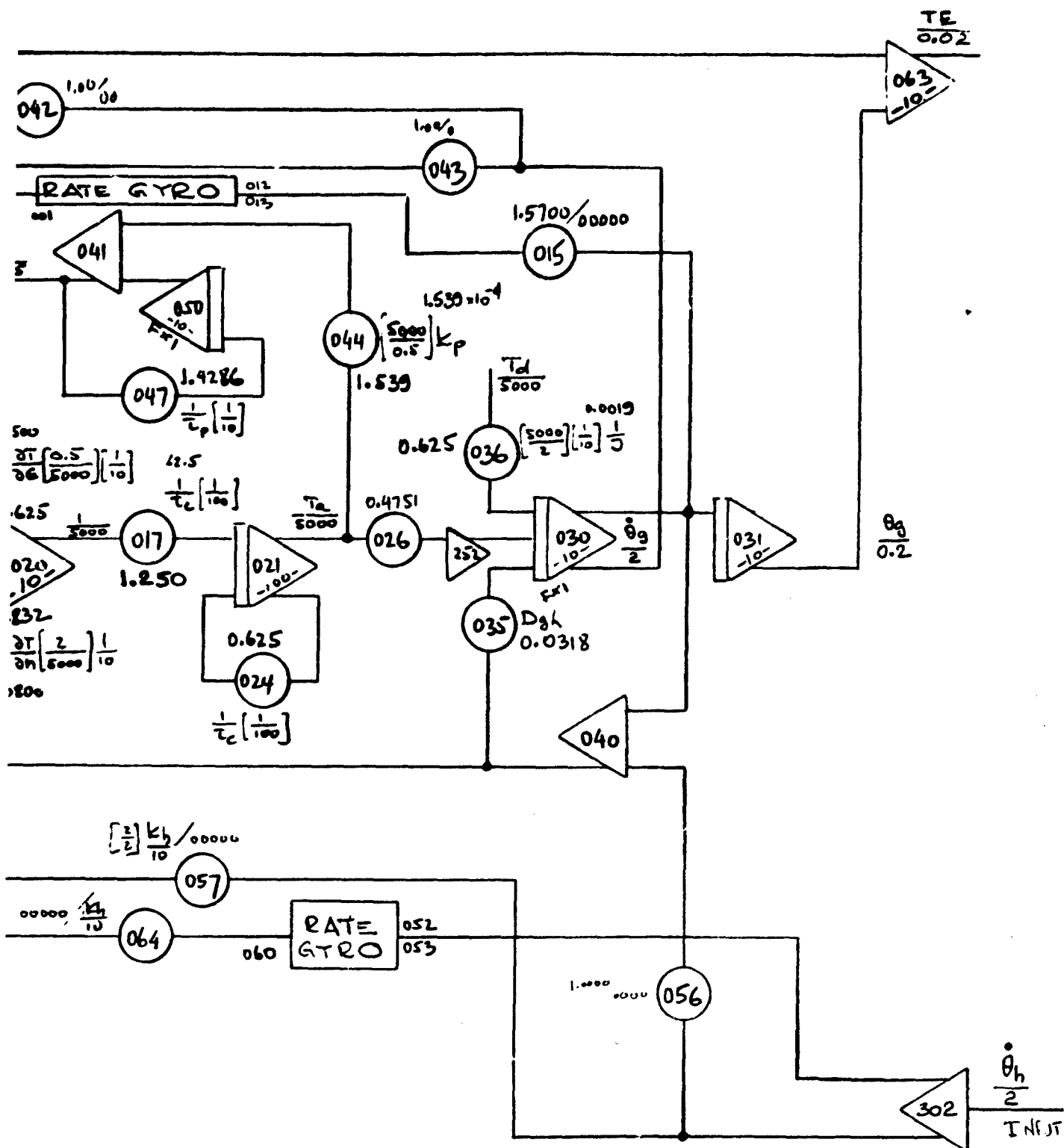


FIG. B9  
M60.10.4.74  
POSITION CONTROL - LINEAR FLOW

Figure B-9. Position Control System



tion Control System with Linear Flow Model

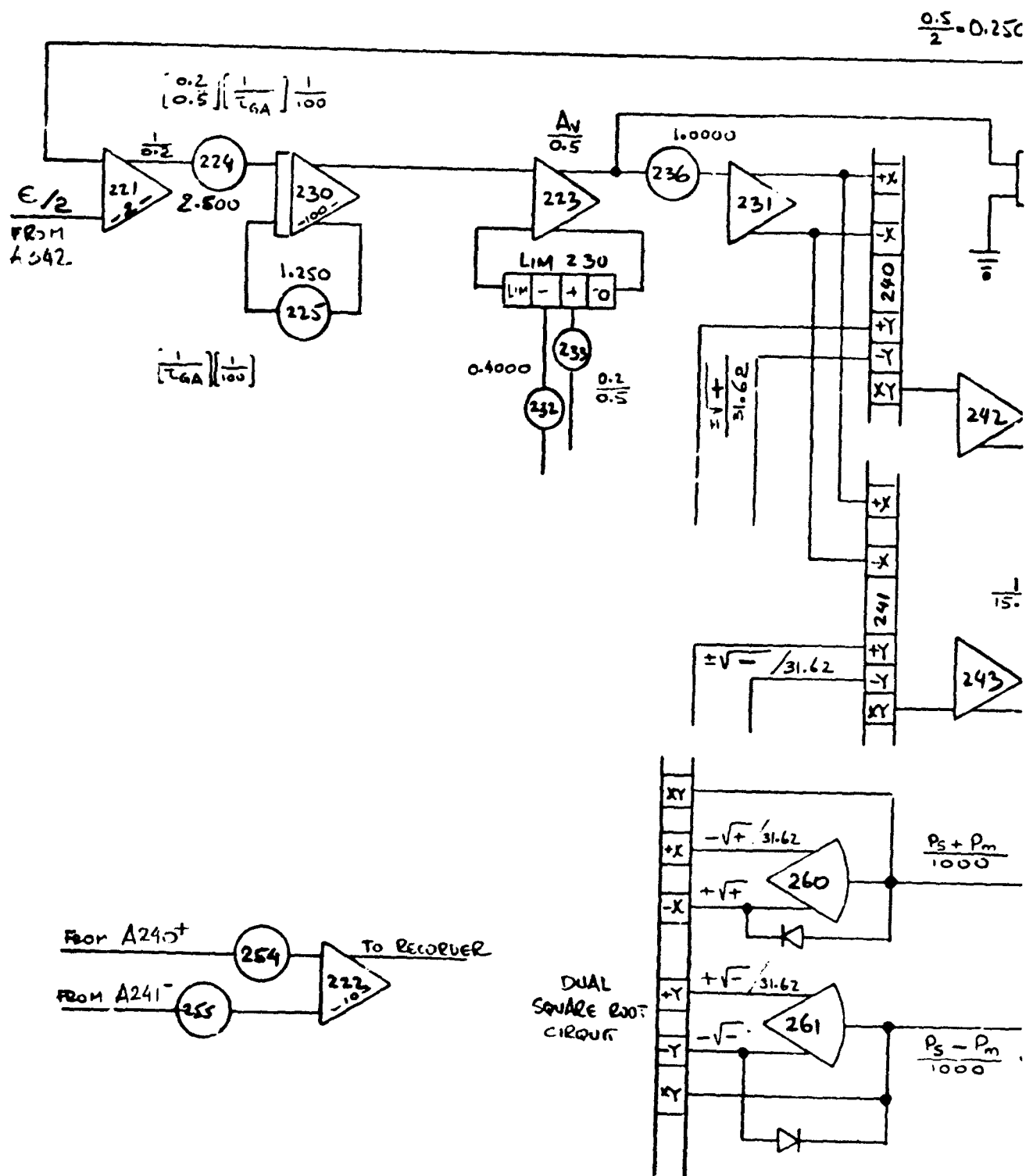
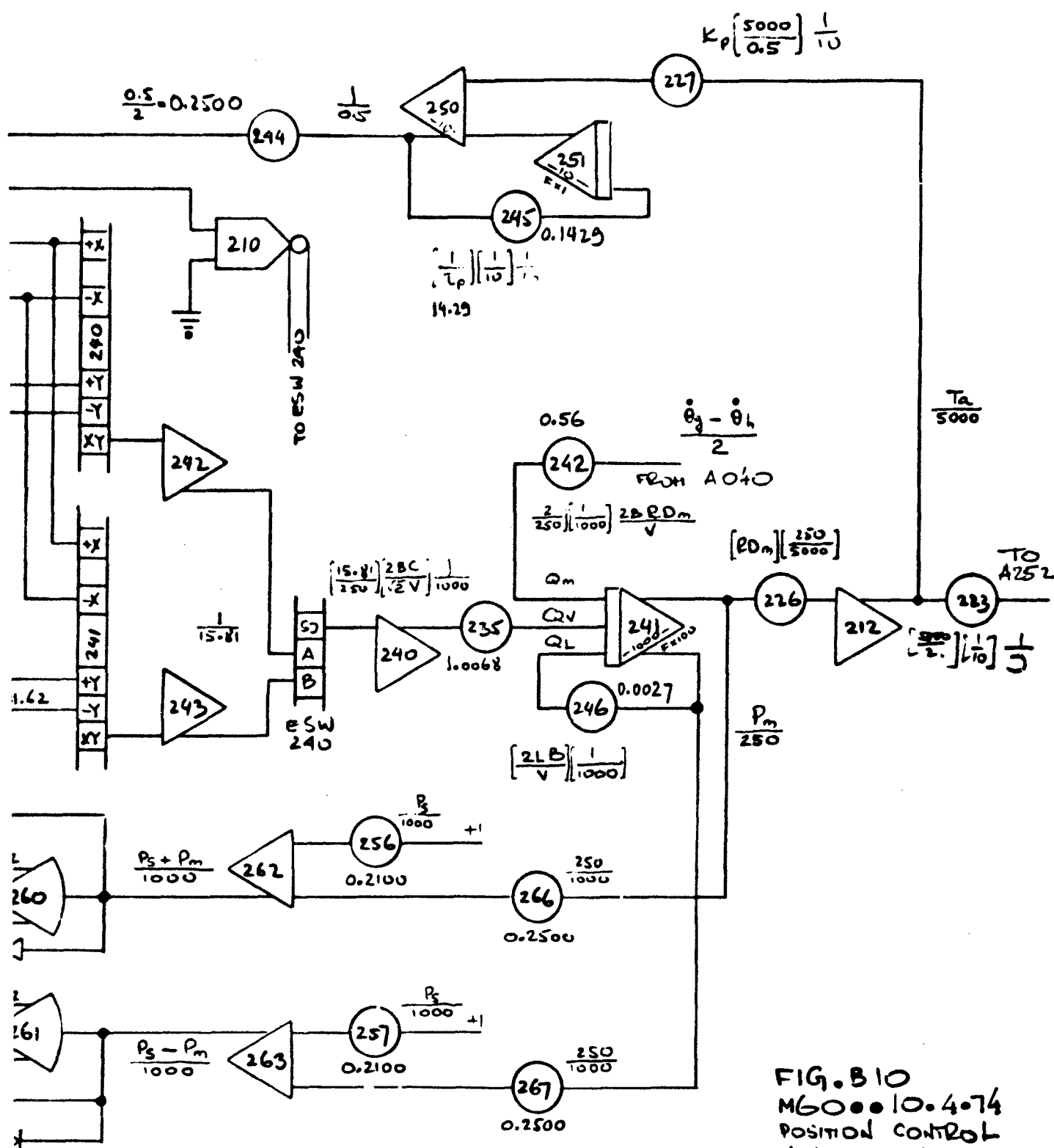
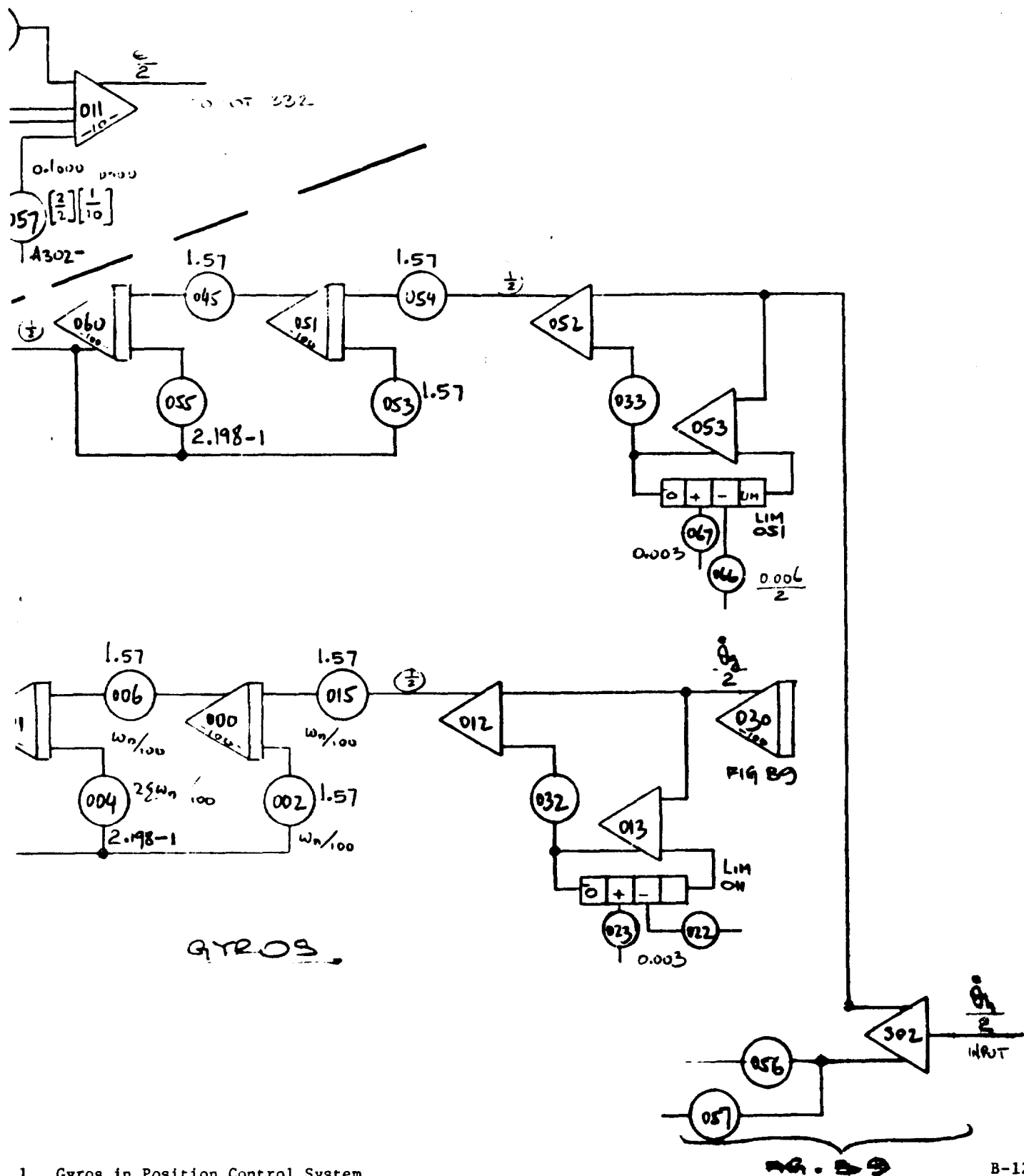


Figure B-10. Nonlinear Flow Model in 1









1. Gyros in Position Control System

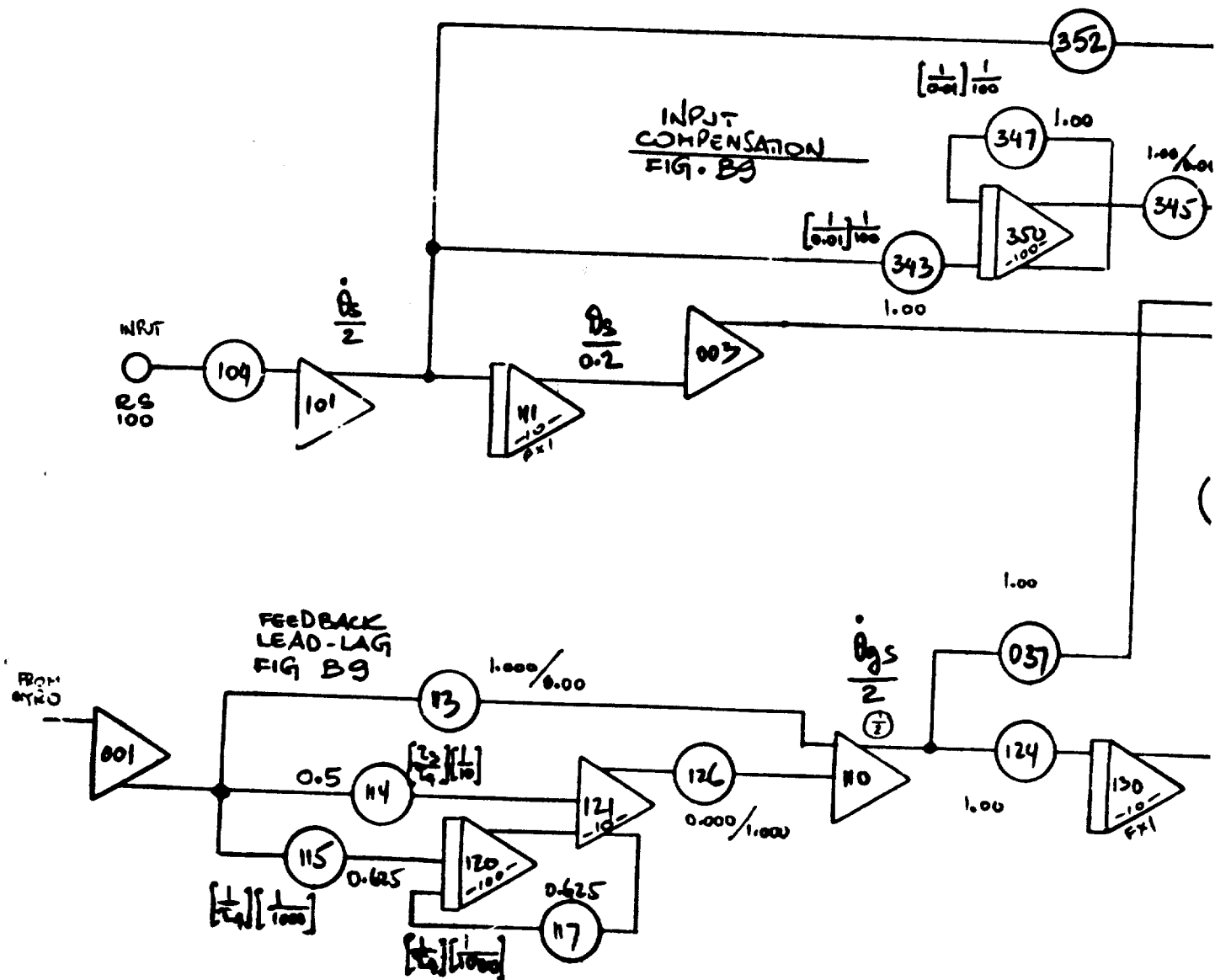


Figure B-12. Compensation Networks in

Best Available Copy  
Best Available Copy

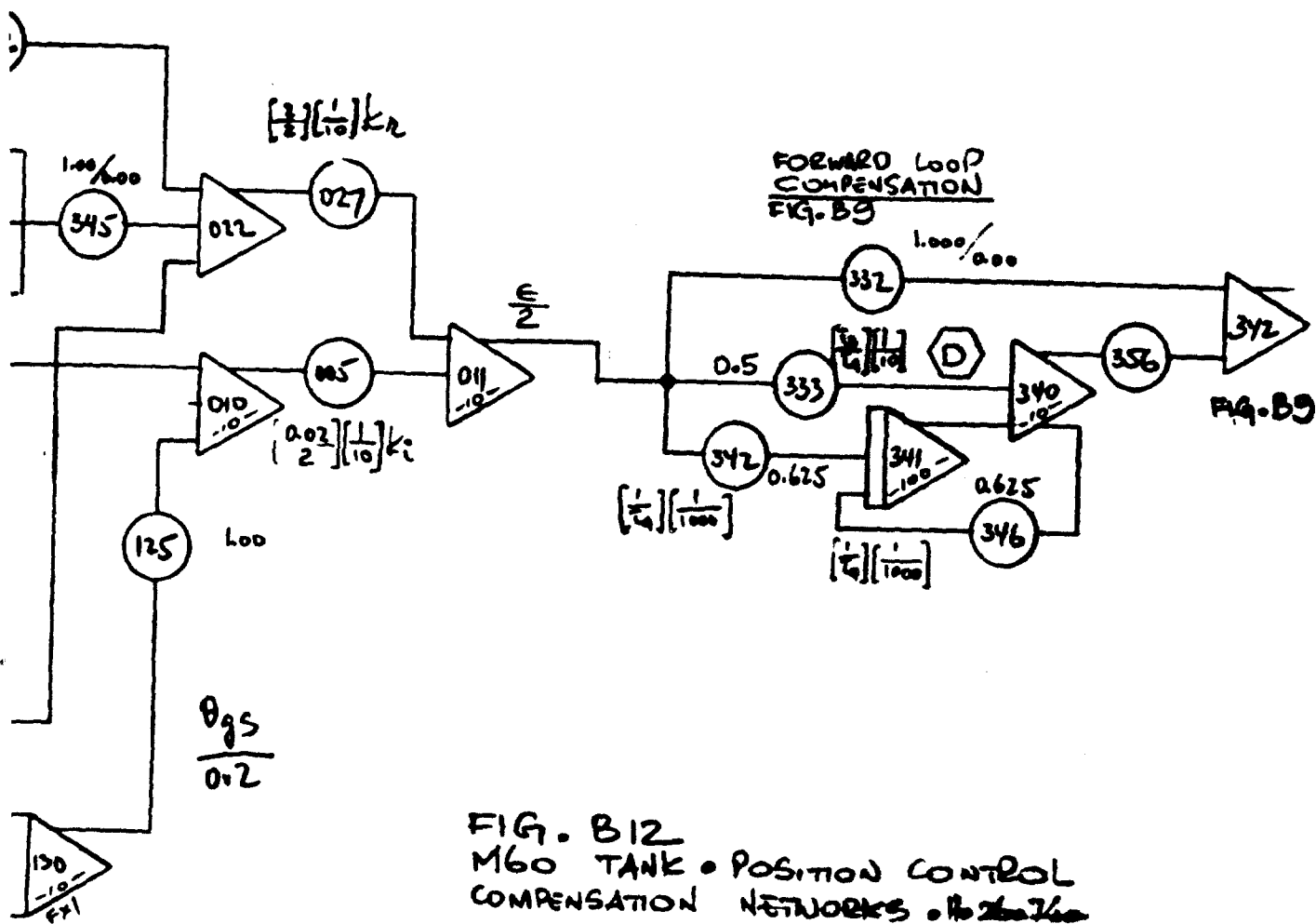


FIG. B12  
M60 TANK • POSITION CONTROL  
COMPENSATION NETWORKS • No. 2626

Networks in Position Control System